

TOSHIBA

**8 Bit Microcontroller
TLCS-870/C Series**

TMP86P820FG

Not Recommended
for New Design

TOSHIBA CORPORATION

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For a discussion of how the reliability of microcontrollers can be predicted, please refer to Section 1.3 of the chapter entitled Quality and Reliability Assurance/Handling Precautions. 030619_S

The Functional Differences on Products basis: TMP86CM29L, TMP86Cx29B, TMP86CH21 and TMP86Cx20

Products name	TMP86CM29L	TMP86C829B TMP86CH29B TMP86CM29B	TMP86CH21 TMP86CH21A	TMP86C420 TMP86C820
ROM	32 K bytes	C829: 8K bytes CH29: 16K bytes CM29: 32K bytes	16K bytes	C420: 4K bytes C820: 8K bytes
RAM	1.5K bytes	C829: 512bytes CH29: 1.5K bytes CM29: 1.5K bytes	512bytes	256bytes
I/O port	39 pins			
Minumum command execution time	0.25μsec at 16MHz			
Supply Voltage	1.8V to 3.6V at 8.0MHz/ 32.768kHz 2.7V to 3.6V at 16MHz/ 32.768kHz (Note4)	1.8V to 5.5V at 4.2MHz/32.768kHz 2.7V to 5.5V at 8.0MHz/32.768kHz 4.5V to 5.5V at 16MHz/32.768kHz		
18-bit Timer counter	1ch (ECIN input is both edge or single edge)			1ch (ECIN input is single edge)
8-bit Timer counter	4ch			2ch
Time base timer	1ch			
Watch dog timer	1ch			
UART/SIO	1ch (Note1)			N.A.
SIO	N.A			1ch
Key-on wakeup	4ch			
A/D converter	10-bit A/D: 8ch		8-bit A/D: 8ch	
LCD driver	32SEG x 4COM			
Operating Temperature	-40 to 85 °C		-40 to 85 °C (Note2)	-40 to 85 °C
Package(Body size)	LQFP64(10x10mm)	QFP64(14x14mm) LQFP64(10x10mm)		
Package (P-QFP64-1010-0.80C)	N.A	TMP86C829BFG TMP86CH29BFG TMP86CM29BFG	TMP86CH21FG	TMP86C420FG TMP86C820FG
Package (P-LQFP64-1010-0.50E)	N.A	TMP86C829BUG TMP86CH29BUG TMP86CM29BUG	TMP86CH21UG	TMP86C420UG TMP86C820UG
Package (P-LQFP64-1010-0.50D)	TMP86CM29LUG	N.A.	TMP86CH21AUG	N.A.

Note 1: UART and SIO can not use function synchronously because each function pins are shared.

Note 2: With TMP86CH21AUG the operating temperature (Topr) is -20 °C to 85 °C when the supply voltage VDD is less than 2.0V.

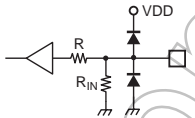
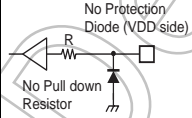
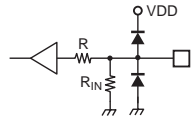
Note 3: TMP86C820/420 don't have the timer/counter-6 input/output and UART input/output.

Note 4: The electrial characteristics of TMP86CM29LUG are different from that of TMP86C829/CH29/CM29B, TMP86CH21/CH21A and TMP86C420/C820. For details, please refer to "Electrical Characteristics" in data sheet of TMP86CM29LUG.

Note 5: The operating temperature (Topr) of AD characteristics of all products (TMP86C420/C820/CH21/CH21A/C829B/CH29B/CM29B/CM29L) is -10 °C to 85 °C when the supply voltage VDD is less than 2.0V. For details, please refer to "AD Conversion Characteristics" in data sheet of each product.

Note 6: The characteristic of power supply current differs in each product. For details, please refer to "Electirical Characteristics" in data sheet of each product.

The Functional Differences on Products basis: TMP86C829B/CH29B/CM29B/PM29A/PM29B/FM29/CM29L.

Products name	TMP86C829B	TMP86CH29B	TMP86CM29B	TMP86PM29A TMP86PM29B	TMP86FM29	TMP86CM29L
ROM	8K bytes (MASK)	16K bytes (MASK)	32K bytes (MASK)	32K bytes (OTP)	32K bytes (FLASH)	32K bytes (MASK)
RAM	512 bytes	1.5K bytes				
DBR	128 bytes (Flash memory control/status registers <EEPCR, EEPSPR> are non-available.)				128 bytes (Flash memory control/status registers <EEPCR, EEPSPR> are available.)	
I/O port	39 pins					
Large current output (Nch) port	4 pins (Sink-open-drain output) 20 mA (Typ)				4 pins (Sink-open-drain output) 6 mA (Typ)	
Interrupt sources	External: 5 Internal: 14					
Timer/Counter	18bit Timer/Counter: 1ch 8bit Timer/Counter: 4ch					
UART/SIO	1ch (Note1)					
Key-on wakeup	4ch					
AD converter	10bit x 8ch (Note3)					
LCD driver	32SEG x 4COM					
Circuitry of TEST pin						
Feedback resistor in High- frequency circuit (Note4)	$R_f = 1.2 \text{ M } \Omega(\text{Typ})$				$R_f = 3 \text{ M } \Omega(\text{Typ})$	
Feedback resistor in Low- frequency circuit (Note4)	$R_f = 6 \text{ M } \Omega(\text{Typ})$				$R_f = 20 \text{ M } \Omega(\text{Typ})$	
Emulation Chip (Note2)	TMP86C929AXB					
Package	P-QFP64-1414-0.80C P-LQFP64-1010-0.50E					P-LQFP64-1010-0.50D
Operating voltage (Note 5)	1.8V to 5.5V at 4.2MHz/32.768kHz 2.7V to 5.5V at 8.0MHz/32.768kHz 4.5V to 5.5V at 16MHz/32.768kHz				1.8V to 3.6V at 8.0MHz/32.768kHz 2.7V to 3.6V at 16MHz/32.768kHz (Note 6)	

Note 1: UART and SIO can not use function synchronously because each function pins are shared.

Note 2: An emulation chip (TMP86C929AXB) can't emulate the Flash memory functions, CPU wait and serial PROM mode.

Therefore, if the software which includes Flash memory function or CPU wait is executed in TMP86C929AXB, the operation might be different from TMP86FM29/CM29L.

Note 3: The operating temperature (Topr) of AD characteristics of all products (TMP86C829B/CH29B/CM29B/PM29A/PM29B/FM29/CM29L) is -10°C to 85°C when the supply voltage VDD is less than 2.0V. For details, please refer to "AD Conversion Characteristics" in data sheet of each product.

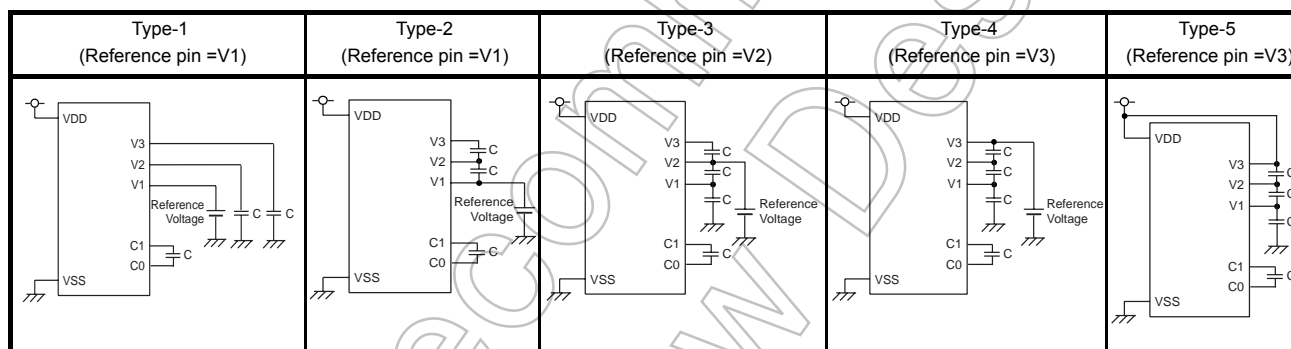
Note 4: The typical value of high and low frequency feedback resistor in TMP86FM29/CM29L are different from that of the other products. For details, please refer to "Input/Output Circuitry" in data sheet of each product.

Note 5: The characteristic of power supply current differs in each product. For details, please refer to "Electrical Characteristics" in data sheet of each product.

Note 6: The recommended operating condition of serial PROM mode in TMP86FM29 is different from MCU mode. For details, please refer to "Electrical Characteristics" in data sheet of each product.

Condition	Wait Time'	Halt/Operate	
		CPU	Peripherals
After reset release	$2^{10}/f_c$ [s]	Halt	Halt
Changing from STOP mode to NORMAL mode (at $EEPCR<MNPWDW>="1"$)	$2^{10}/f_c$ [s]	Halt	Operate
Changing from STOP mode to SLOW mode (at $EEPCR<MNPWDW>="1"$)	$2^3/f_c$ [s]	Halt	Operate
Changing from IDLE0/1/2 mode to NORMAL mode (at $EEPCR<MNPWDW>="0"$)	$2^{10}/f_c$ [s]	Halt	Operate
Changing from SLEEP0/1/2 mode to SLOW mode (at $EEPCR<MNPWDW>="0"$)	$2^3/f_c$ [s]	Halt	Operate

Note 1: TMP86FM29 has a CPU wait function which is a warming up (CPU halt) of CPU for stabilizing of power supply of Flash memory. Even though TMP86CM29L doesn't have a Flash memory, the CPU wait function is inserted to keep the compatibility with Flash product (TMP86FM29). During the CPU wait period except RESET, CPU is halted but peripheral functions are not halted. Therefore, if the interrupt occurs during the CPU wait period, the interrupt latch (IL) is set and when IMF has been set to "1", the interrupt service routine might be executed after CPU wait period. For details, please refer to "Flash Memory" in TMP86FM29 data sheet. TMP86FM29 (Flash product) should be used as non-volatile product to confirm the software of TMP86CM29L because of the above reason. And TMP86PM29A/PM29B (OTP product) should be used as non-volatile product to confirm the software of TMP86C829B/CH29B/CM29B.



Note 1: TMP86FM29/CM29L can't use LCD panel which is driven by 5V because the maximum recommended voltage is 3.6V. Therefore, the voltage level of V3 pin always should be under 3.6V.

Note 2: The operating temperature of TMP86FM29/CM29L in Type-1 and Type-2 is -10 °C to 85 °C. For details, please refer to "LCD Driver" and "Electrical Characteristics" in data sheet.

Note 3: The operating temperature of TMP86C829B/CH29B/CM29B in all Types (Type 1 to 5) is -40 °C to 85 °C. However, there is a voltage level limitation of V3 and VDD pin in each type. For details, please refer to "LCD Driver" and "Electrical Characteristics" in data sheet.

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Revision History

Date	Revision	
2006/12/21	1	First Release
2007/1/18	2	Contents Revised
2007/6/28	3	Contents Revised
2008/8/29	4	Contents Revised

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Caution in Setting the UART Noise Rejection Time

When UART is used, settings of RXDNC are limited depending on the transfer clock specified by BRG. The combination "O" is available but please do not select the combination "-".

The transfer clock generated by timer/counter interrupt is calculated by the following equation :

$$\text{Transfer clock [Hz]} = \text{Timer/counter source clock [Hz]} \div \text{TTREG set value}$$

BRG setting	Transfer clock [Hz]	RXDNC setting			
		00 (No noise rejection)	01 (Reject pulses shorter than 31/fc[s] as noise)	10 (Reject pulses shorter than 63/fc[s] as noise)	11 (Reject pulses shorter than 127/fc[s] as noise)
000	fc/13	O	O	O	-
110 (When the transfer clock generated by timer/counter interrupt is the same as the right side column)	fc/8	O	-	-	-
	fc/16	O	O	-	-
	fc/32	O	O	O	-
The setting except the above		O	O	O	O

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17. Package Dimensions

This is a technical document that describes the operating functions and electrical specifications of the 8-bit microcontroller series TLCS-870/C (LSI).

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CMOS 8-Bit Microcontroller

TMP86P820FG

The TMP86P820FG is a single-chip 8-bit high-speed and high-functionality microcomputer incorporating 8192 bytes of One-Time PROM. It is pin-compatible with the TMP86C820FG/420FG (Mask ROM version). The TMP86P820FG can realize operations equivalent to those of the TMP86C820FG/420FG by programming the on-chip PROM.

Product No.	ROM (EPROM)	RAM	Package	MaskROM MCU	Emulation Chip
TMP86P820FG	8192 bytes	256 bytes	QFP64-P-1414-0.80C	TMP86C820FG/420FG	TMP86C929AXB

1.1 Features

- 8-bit single chip microcomputer TLCS-870/C series
 - Instruction execution time :
 - 0.25 μ s (at 16 MHz)
 - 122 μ s (at 32.768 kHz)
 - 132 types & 731 basic instructions
- 15 interrupt sources (External : 5 Internal : 10)
- Input / Output ports (39 pins)
 - Large current output: 4 pins (Typ. 20mA), LED direct drive
- Prescaler
 - Time base timer
 - Divider output function
- Watchdog Timer
- 18-bit Timer/Counter : 1 ch
 - Timer Mode
 - Event Counter Mode
 - Pulse Width Measurement Mode
 - Frequency Measurement Mode

060116EBP

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• For a discussion of how the reliability of microcontrollers can be predicted, please refer to Section 1.3 of the chapter entitled Quality and Reliability Assurance/Handling Precautions. 030619_S

7. 8-bit timer counter : 2 ch
 - Timer, Event counter, Programmable divider output (PDO),
Pulse width modulation (PWM) output,
Programmable pulse generation (PPG) modes
8. 8-bit SIO: 1 ch
9. 8-bit successive approximation type AD converter (with sample hold)
 - Analog inputs: 8ch
10. Key-on wakeup : 4 ch
11. LCD driver/controller
 - Built-in voltage booster for LCD driver With display memory
 - LCD direct drive capability (MAX 32 seg × 4 com)
 - 1/4, 1/3, 1/2 duties or static drive are programmably selectable
12. Clock operation
 - Single clock mode
 - Dual clock mode
13. Low power consumption operation
 - STOP mode: Oscillation stops. (Battery/Capacitor back-up.)
 - SLOW1 mode: Low power consumption operation using low-frequency clock. (High-frequency clock stop.)
 - SLOW2 mode: Low power consumption operation using low-frequency clock. (High-frequency clock oscillate.)
 - IDLE0 mode: CPU stops, and only the Time-Based-Timer (TBT) on peripherals operate using high frequency clock. Release by falling edge of the source clock which is set by TBTCCR<TBTCK>.
 - IDLE1 mode: CPU stops and peripherals operate using high frequency clock. Release by interrupt. (CPU restarts).
 - IDLE2 mode: CPU stops and peripherals operate using high and low frequency clock. Release by interrupt. (CPU restarts).
 - SLEEP0 mode: CPU stops, and only the Time-Based-Timer (TBT) on peripherals operate using low frequency clock. Release by falling edge of the source clock which is set by TBTCCR<TBTCK>.
 - SLEEP1 mode: CPU stops, and peripherals operate using low frequency clock. Release by interrupt. (CPU restarts).
 - SLEEP2 mode: CPU stops and peripherals operate using high and low frequency clock. Release by interrupt.
14. Wide operation voltage:
 - 4.5 V to 5.5 V at 16MHz / 32.768 kHz
 - 2.7 V to 5.5 V at 8 MHz / 32.768 kHz
 - 1.8 V to 5.5 V at 4.2MHz / 32.768 kHz

1.2 Pin Assignment

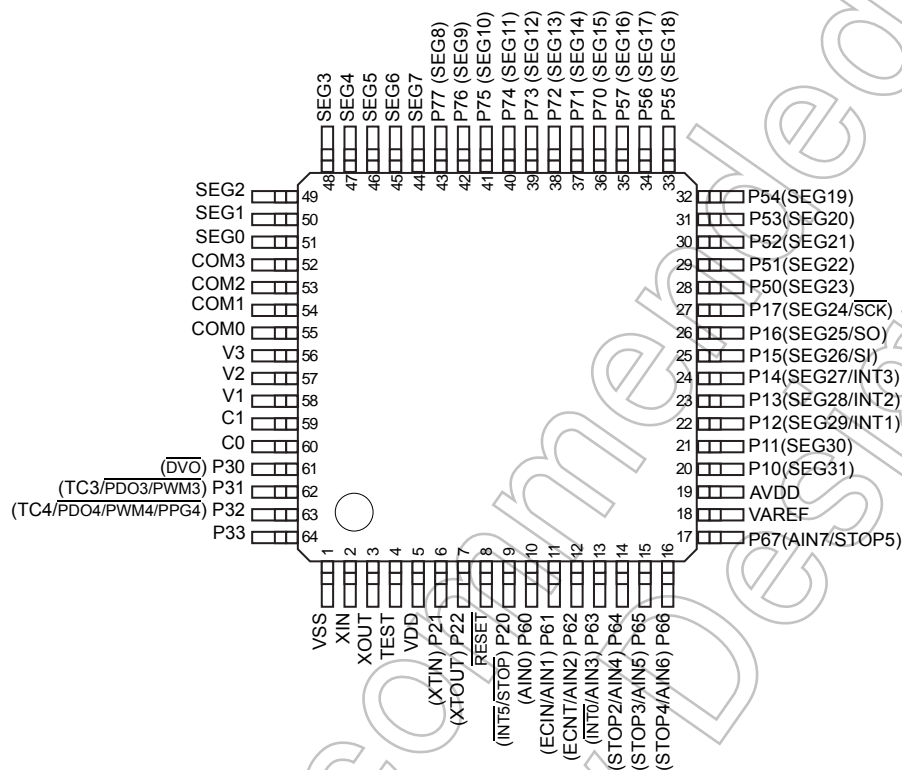


Figure 1-1 Pin Assignment

1.3 Block Diagram

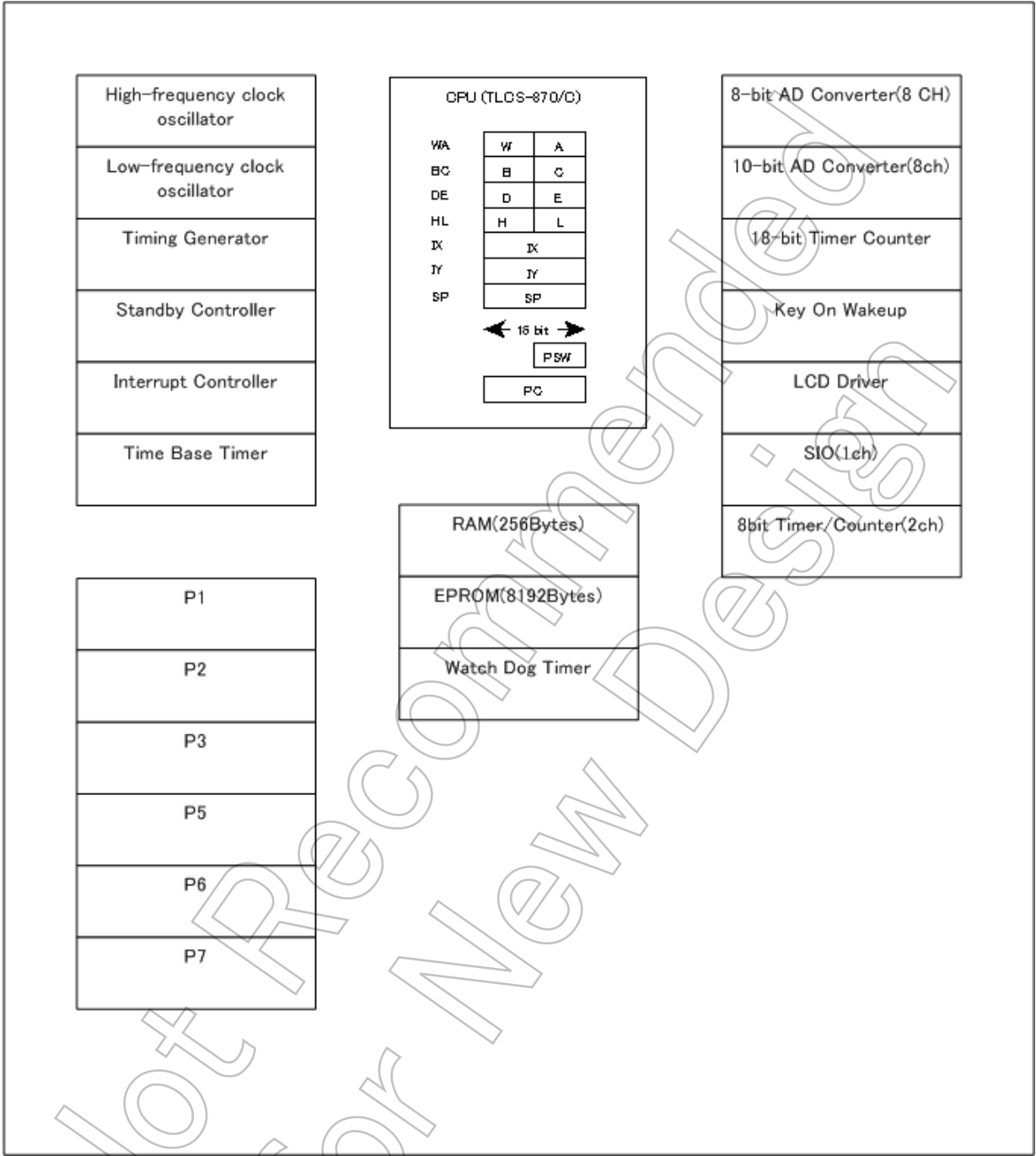


Figure 1-2 Block Diagram

1.4 Pin Names and Functions

Table 1-1 Pin Names and Functions(1/3)

Pin Name	Pin Number	Input/Output	Functions
P17 SEG24 SCK	27	IO O IO	PORT17 LCD segment output 24 Serial Clock I/O
P16 SEG25 SO	26	IO O O	PORT16 LCD segment output 25 Serial Data Output
P15 SEG26 SI	25	IO O I	PORT15 LCD segment output 26 Serial Data Input
P14 SEG27 INT3	24	IO O I	PORT14 LCD segment output 27 External interrupt 3 input
P13 SEG28 INT2	23	IO O I	PORT13 LCD segment output 28 External interrupt 2 input
P12 SEG29 INT1	22	IO O I	PORT12 LCD segment output 29 External interrupt 1 input
P11 SEG30	21	IO O	PORT11 LCD segment output 30
P10 SEG31	20	IO O	PORT10 LCD segment output 31
P22 XTOUT	7	IO O	PORT22 Resonator connecting pins(32.768kHz) for inputting external clock
P21 XTIN	6	IO I	PORT21 Resonator connecting pins(32.768kHz) for inputting external clock
P20 STOP INT5	9	IO I I	PORT20 STOP mode release signal input External interrupt 5 input
P33	64	IO	PORT33
P32 PDO4/PWM4/PPG4 TC4	63	IO O I	PORT32 PDO4/PWM4/PPG4 output TC4 input
P31 PDO3/PWM3 TC3	62	IO O I	PORT31 PDO3/PWM3 output TC3 input
P30 DVO	61	IO O	PORT30 Divider Output
P57 SEG16	35	IO O	PORT57 LCD segment output 16
P56 SEG17	34	IO O	PORT56 LCD segment output 17
P55 SEG18	33	IO O	PORT55 LCD segment output 18

Table 1-1 Pin Names and Functions(2/3)

Pin Name	Pin Number	Input/Output	Functions
P54 SEG19	32	IO O	PORT54 LCD segment output 19
P53 SEG20	31	IO O	PORT53 LCD segment output 20
P52 SEG21	30	IO O	PORT52 LCD segment output 21
P51 SEG22	29	IO O	PORT51 LCD segment output 22
P50 SEG23	28	IO O	PORT50 LCD segment output 23
P67 AIN7 STOP5	17	IO I I	PORT67 Analog Input7 STOP5 input
P66 AIN6 STOP4	16	IO I I	PORT66 Analog Input6 STOP4 input
P65 AIN5 STOP3	15	IO I I	PORT65 Analog Input5 STOP3 input
P64 AIN4 STOP2	14	IO I I	PORT64 Analog Input4 STOP2 input
P63 AIN3 INT0	13	IO I I	PORT63 Analog Input3 External interrupt 0 input
P62 AIN2 ECNT	12	IO I I	PORT62 Analog Input2 ECNT input
P61 AIN1 ECIN	11	IO I I	PORT61 Analog Input1 ECIN input
P60 AIN0	10	IO I	PORT60 Analog Input0
P77 SEG8	43	IO O	PORT77 LCD segment output 8
P76 SEG9	42	IO O	PORT76 LCD segment output 9
P75 SEG10	41	IO O	PORT75 LCD segment output 10
P74 SEG11	40	IO O	PORT74 LCD segment output 11
P73 SEG12	39	IO O	PORT73 LCD segment output 12
P72 SEG13	38	IO O	PORT72 LCD segment output 13
P71 SEG14	37	IO O	PORT71 LCD segment output 14
P70 SEG15	36	IO O	PORT70 LCD segment output 15

Table 1-1 Pin Names and Functions(3/3)

Pin Name	Pin Number	Input/Output	Functions
SEG7	44	O	LCD segment output 7
SEG6	45	O	LCD segment output 6
SEG5	46	O	LCD segment output 5
SEG4	47	O	LCD segment output 4
SEG3	48	O	LCD segment output 3
SEG2	49	O	LCD segment output 2
SEG1	50	O	LCD segment output 1
SEG0	51	O	LCD segment output 0
COM3	52	O	LCD common output 3
COM2	53	O	LCD common output 2
COM1	54	O	LCD common output 1
COM0	55	O	LCD common output 0
V3	56	I	LCD voltage booster pin
V2	57	I	LCD voltage booster pin
V1	58	I	LCD voltage booster pin
C1	59	I	LCD voltage booster pin
C0	60	I	LCD voltage booster pin
XIN	2	I	Resonator connecting pins for high-frequency clock
XOUT	3	O	Resonator connecting pins for high-frequency clock
RESET	8	IO	Reset signal
TEST	4	I	Test pin for out-going test. Normally, be fixed to low.
VAREF	18	I	Analog Base Voltage Input Pin for A/D Conversion
AVDD	19	I	Analog Power Supply
VDD	5	I	Power Supply
VSS	1	I	0(GND)

Not Recommended
for New Design

2. Operational Description

2.1 CPU Core Functions

The CPU core consists of a CPU, a system clock controller, and an interrupt controller.

This section provides a description of the CPU core, the program memory, the data memory, and the reset circuit.

2.1.1 Memory Address Map

The TMP86P820FG memory is composed OTP, RAM, DBR(Data buffer register) and SFR(Special function register). They are all mapped in 64-Kbyte address space. Figure 2-1 shows the TMP86P820FG memory address map.

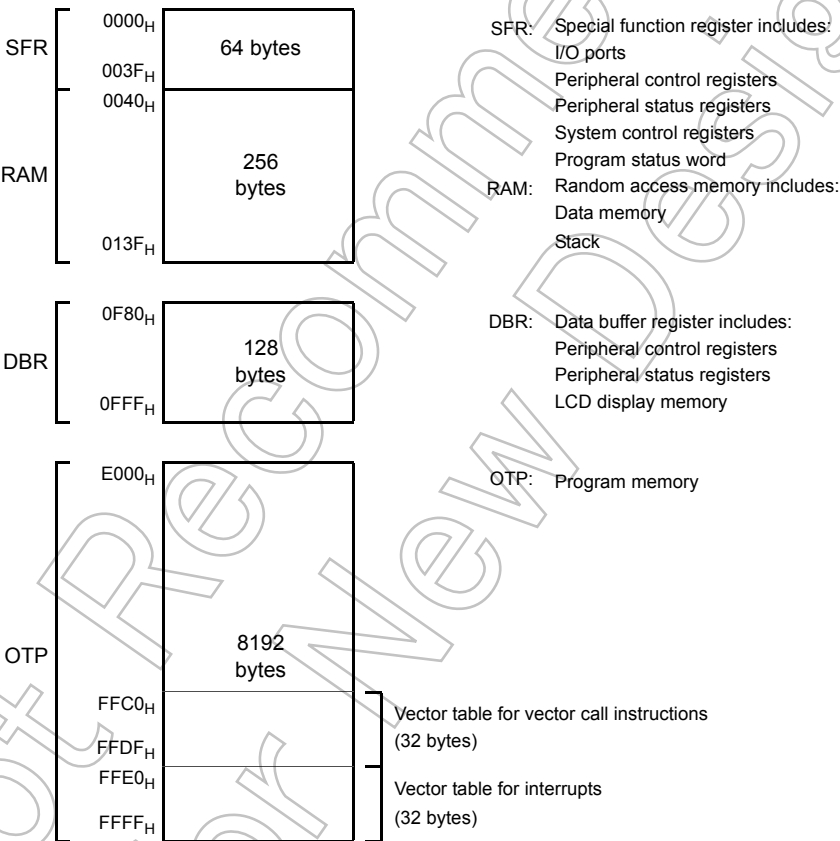


Figure 2-1 Memory Address Map

2.1.2 Program Memory (OTP)

The TMP86P820FG has a 8192 bytes (Address E000H to FFFFH) of program memory (OTP).

2.1.3 Data Memory (RAM)

The TMP86P820FG has 256bytes (Address 0040H to 013FH) of internal RAM. The first 192 bytes (0040H to 00FFH) of the internal RAM are located in the direct area; instructions with shorten operations are available against such an area.

The data memory contents become unstable when the power supply is turned on; therefore, the data memory should be initialized by an initialization routine.

Example :Clears RAM to “00H”. (TMP86P820FG)

```
LD      HL, 0040H      ; Start address setup
LD      A, H           ; Initial value (00H) setup
LD      BC, 00FFH
SRAMCLR: LD      (HL), A
INC     HL
DEC     BC
JRS     F, SRAMCLR
```

2.2 System Clock Controller

The system clock controller consists of a clock generator, a timing generator, and a standby controller.

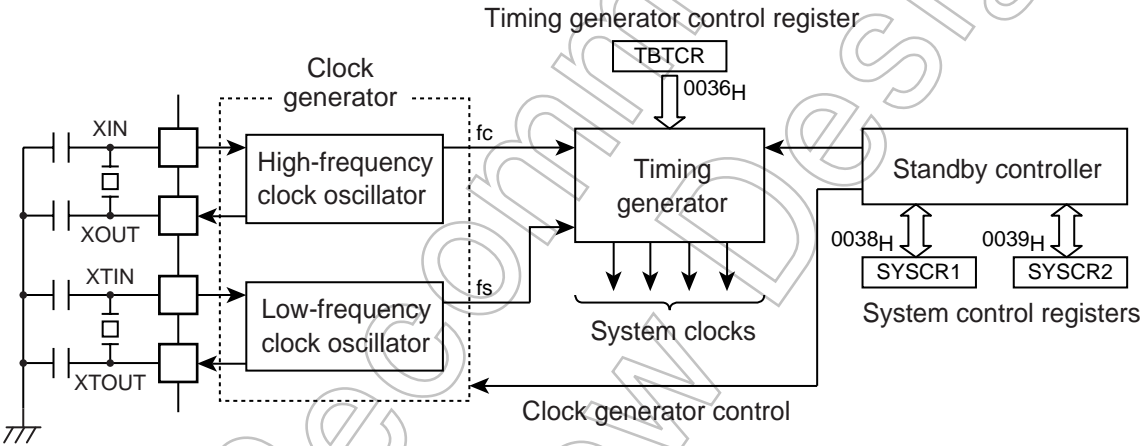


Figure 2-2 System Clock Control

2.2.1 Clock Generator

The clock generator generates the basic clock which provides the system clocks supplied to the CPU core and peripheral hardware. It contains two oscillation circuits: One for the high-frequency clock and one for the low-frequency clock. Power consumption can be reduced by switching of the standby controller to low-power operation based on the low-frequency clock.

The high-frequency (fc) clock and low-frequency (fs) clock can easily be obtained by connecting a resonator between the XIN/XOUT and XTIN/XTOUT pins respectively. Clock input from an external oscillator is also possible. In this case, external clock is applied to XIN/XTIN pin with XOUT/XTOUT pin not connected.

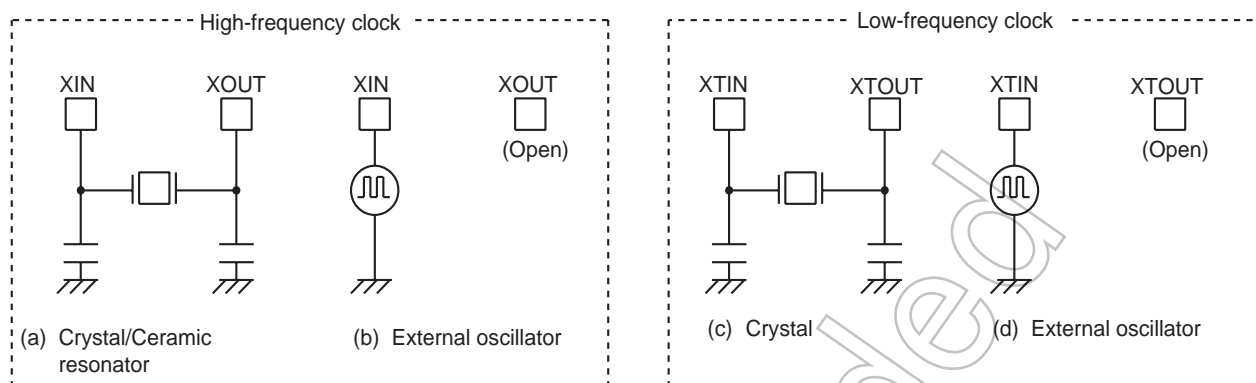


Figure 2-3 Examples of Resonator Connection

Note: The function to monitor the basic clock directly at external is not provided for hardware, however, with disabling all interrupts and watchdog timers, the oscillation frequency can be adjusted by monitoring the pulse which the fixed frequency is outputted to the port by the program.
The system to require the adjustment of the oscillation frequency should create the program for the adjustment in advance.

2.2.2 Timing Generator

The timing generator generates the various system clocks supplied to the CPU core and peripheral hardware from the basic clock (fc or fs). The timing generator provides the following functions.

1. Generation of main system clock
2. Generation of divider output (\overline{DVO}) pulses
3. Generation of source clocks for time base timer
4. Generation of source clocks for watchdog timer
5. Generation of internal source clocks for timer/counters
6. Generation of warm-up clocks for releasing STOP mode
7. LCD

2.2.2.1 Configuration of timing generator

The timing generator consists of a 2-stage prescaler, a 21-stage divider, a main system clock generator, and machine cycle counters.

An input clock to the 7th stage of the divider depends on the operating mode, SYSCR2<SYSCK> and TBTCR<DV7CK>, that is shown in Figure 2-4. As reset and STOP mode started/canceled, the prescaler and the divider are cleared to “0”.

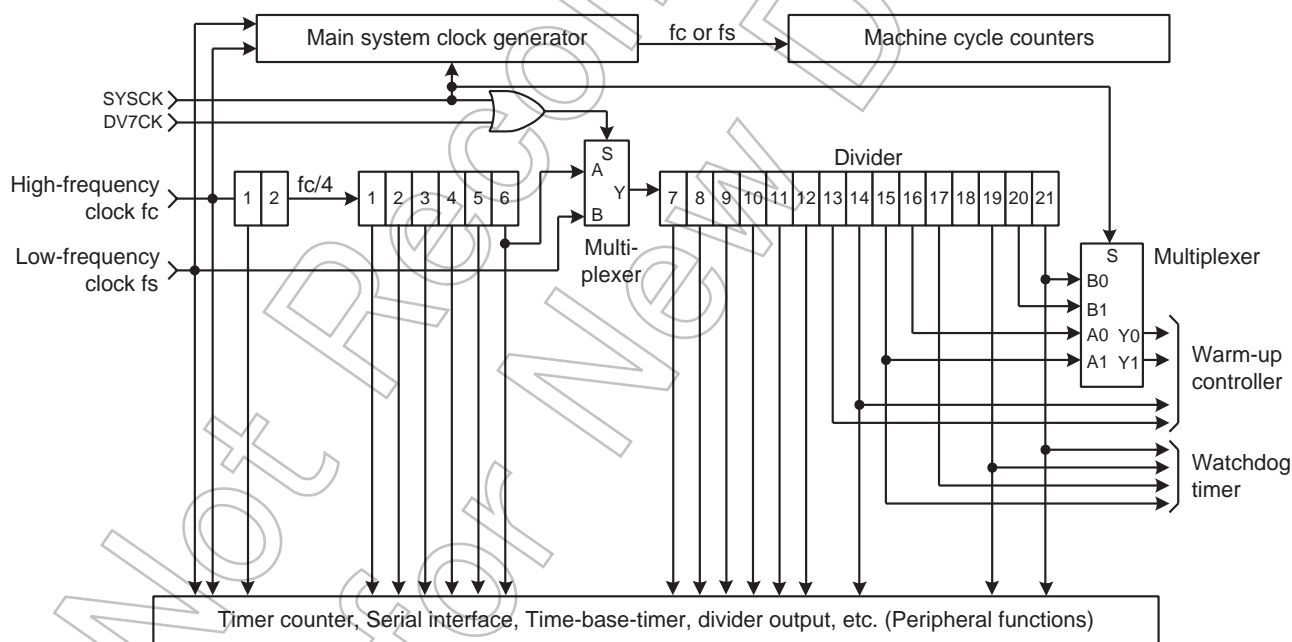


Figure 2-4 Configuration of Timing Generator

Timing Generator Control Register

TBTCR (0036H)	7	6	5	4	3	2	1	0	
	(DVOEN)	(DVOCK)	DV7CK	(TBTEN)	(TBTCK)				(Initial value: 0000 0000)
	DV7CK	Selection of input to the 7th stage of the divider			0: $fc/2^8$ [Hz] 1: fs				R/W

- Note 1: In single clock mode, do not set DV7CK to “1”.
- Note 2: Do not set “1” on DV7CK while the low-frequency clock is not operated stably.
- Note 3: fc: High-frequency clock [Hz], fs: Low-frequency clock [Hz], *: Don't care
- Note 4: In SLOW1/2 and SLEEP1/2 modes, the DV7CK setting is ineffective, and fs is input to the 7th stage of the divider.
- Note 5: When STOP mode is entered from NORMAL1/2 mode, the DV7CK setting is ineffective during the warm-up period after release of STOP mode, and the 6th stage of the divider is input to the 7th stage during this period.

2.2.2.2 Machine cycle

Instruction execution and peripheral hardware operation are synchronized with the main system clock.

The minimum instruction execution unit is called an “machine cycle”. There are a total of 10 different types of instructions for the TLCS-870/C Series: Ranging from 1-cycle instructions which require one machine cycle for execution to 10-cycle instructions which require 10 machine cycles for execution. A machine cycle consists of 4 states (S0 to S3), and each state consists of one main system clock.

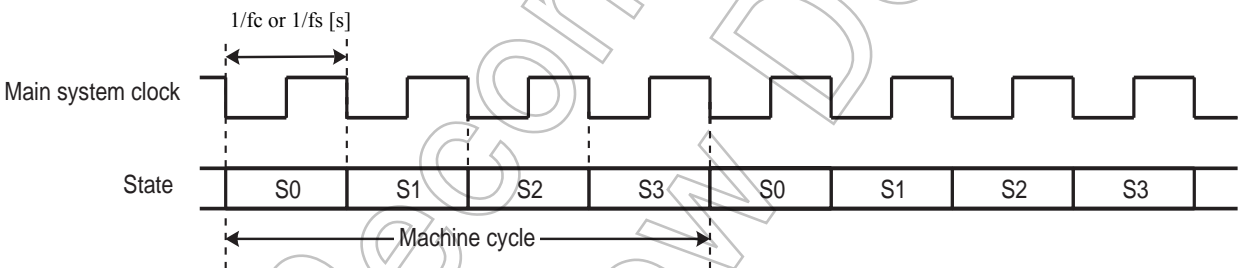


Figure 2-5 Machine Cycle

2.2.3 Operation Mode Control Circuit

The operation mode control circuit starts and stops the oscillation circuits for the high-frequency and low-frequency clocks, and switches the main system clock. There are three operating modes: Single clock mode, dual clock mode and STOP mode. These modes are controlled by the system control registers (SYSCR1 and SYSCR2). Figure 2-6 shows the operating mode transition diagram.

2.2.3.1 Single-clock mode

Only the oscillation circuit for the high-frequency clock is used, and P21 (XTIN) and P22 (XTOUT) pins are used as input/output ports. The main-system clock is obtained from the high-frequency clock. In the single-clock mode, the machine cycle time is $4/fc$ [s].

(1) NORMAL1 mode

In this mode, both the CPU core and on-chip peripherals operate using the high-frequency clock. The TMP86P820FG is placed in this mode after reset.

(2) IDLE1 mode

In this mode, the internal oscillation circuit remains active. The CPU and the watchdog timer are halted; however on-chip peripherals remain active (Operate using the high-frequency clock).

IDLE1 mode is started by SYSCR2<IDLE> = "1", and IDLE1 mode is released to NORMAL1 mode by an interrupt request from the on-chip peripherals or external interrupt inputs. When the IMF (Interrupt master enable flag) is "1" (Interrupt enable), the execution will resume with the acceptance of the interrupt, and the operation will return to normal after the interrupt service is completed. When the IMF is "0" (Interrupt disable), the execution will resume with the instruction which follows the IDLE1 mode start instruction.

(3) IDLE0 mode

In this mode, all the circuit, except oscillator and the timer-base-timer, stops operation.

This mode is enabled by SYSCR2<TGHALT> = "1".

When IDLE0 mode starts, the CPU stops and the timing generator stops feeding the clock to the peripheral circuits other than TBT. Then, upon detecting the falling edge of the source clock selected with TBTCCR<TBTCK>, the timing generator starts feeding the clock to all peripheral circuits.

When returned from IDLE0 mode, the CPU restarts operating, entering NORMAL1 mode back again. IDLE0 mode is entered and returned regardless of how TBTCCR<TBTEN> is set. When IMF = "1", EF6 (TBT interrupt individual enable flag) = "1", and TBTCCR<TBTEN> = "1", interrupt processing is performed. When IDLE0 mode is entered while TBTCCR<TBTEN> = "1", the INTTBT interrupt latch is set after returning to NORMAL1 mode.

2.2.3.2 Dual-clock mode

Both the high-frequency and low-frequency oscillation circuits are used in this mode. P21 (XTIN) and P22 (XTOUT) pins cannot be used as input/output ports. The main system clock is obtained from the high-frequency clock in NORMAL2 and IDLE2 modes, and is obtained from the low-frequency clock in SLOW and SLEEP modes. The machine cycle time is $4/f_c$ [s] in the NORMAL2 and IDLE2 modes, and $4/f_s$ [s] ($122\ \mu\text{s}$ at $f_s = 32.768\ \text{kHz}$) in the SLOW and SLEEP modes.

The TLCS-870/C is placed in the signal-clock mode during reset. To use the dual-clock mode, the low-frequency oscillator should be turned on at the start of a program.

(1) NORMAL2 mode

In this mode, the CPU core operates with the high-frequency clock. On-chip peripherals operate using the high-frequency clock and/or low-frequency clock.

(2) SLOW2 mode

In this mode, the CPU core operates with the low-frequency clock, while both the high-frequency clock and the low-frequency clock are operated. As the SYSCR2<SYSCK> becomes "1", the hardware changes into SLOW2 mode. As the SYSCR2<SYSCK> becomes "0", the hardware changes into NORMAL2 mode. As the SYSCR2<XEN> becomes "0", the hardware changes into SLOW1 mode. Do not clear SYSCR2<XTEN> to "0" during SLOW2 mode.

(3) SLOW1 mode

This mode can be used to reduce power-consumption by turning off oscillation of the high-frequency clock. The CPU core and on-chip peripherals operate using the low-frequency clock.

Switching back and forth between SLOW1 and SLOW2 modes are performed by SYSCR2<XEN>. In SLOW1 and SLEEP modes, the input clock to the 1st stage of the divider is stopped; output from the 1st to 6th stages is also stopped.

(4) IDLE2 mode

In this mode, the internal oscillation circuit remain active. The CPU and the watchdog timer are halted; however, on-chip peripherals remain active (Operate using the high-frequency clock and/or the low-frequency clock). Starting and releasing of IDLE2 mode are the same as for IDLE1 mode, except that operation returns to NORMAL2 mode.

(5) SLEEP1 mode

In this mode, the internal oscillation circuit of the low-frequency clock remains active. The CPU, the watchdog timer, and the internal oscillation circuit of the high-frequency clock are halted; however, on-chip peripherals remain active (Operate using the low-frequency clock). Starting and releasing of SLEEP mode are the same as for IDLE1 mode, except that operation returns to SLOW1 mode. In SLOW1 and SLEEP1 modes, the input clock to the 1st stage of the divider is stopped; output from the 1st to 6th stages is also stopped.

(6) SLEEP2 mode

The SLEEP2 mode is the idle mode corresponding to the SLOW2 mode. The status under the SLEEP2 mode is same as that under the SLEEP1 mode, except for the oscillation circuit of the high-frequency clock.

(7) SLEEP0 mode

In this mode, all the circuit, except oscillator and the timer-base-timer, stops operation. This mode is enabled by setting "1" on bit SYSCR2<TGHALT>.

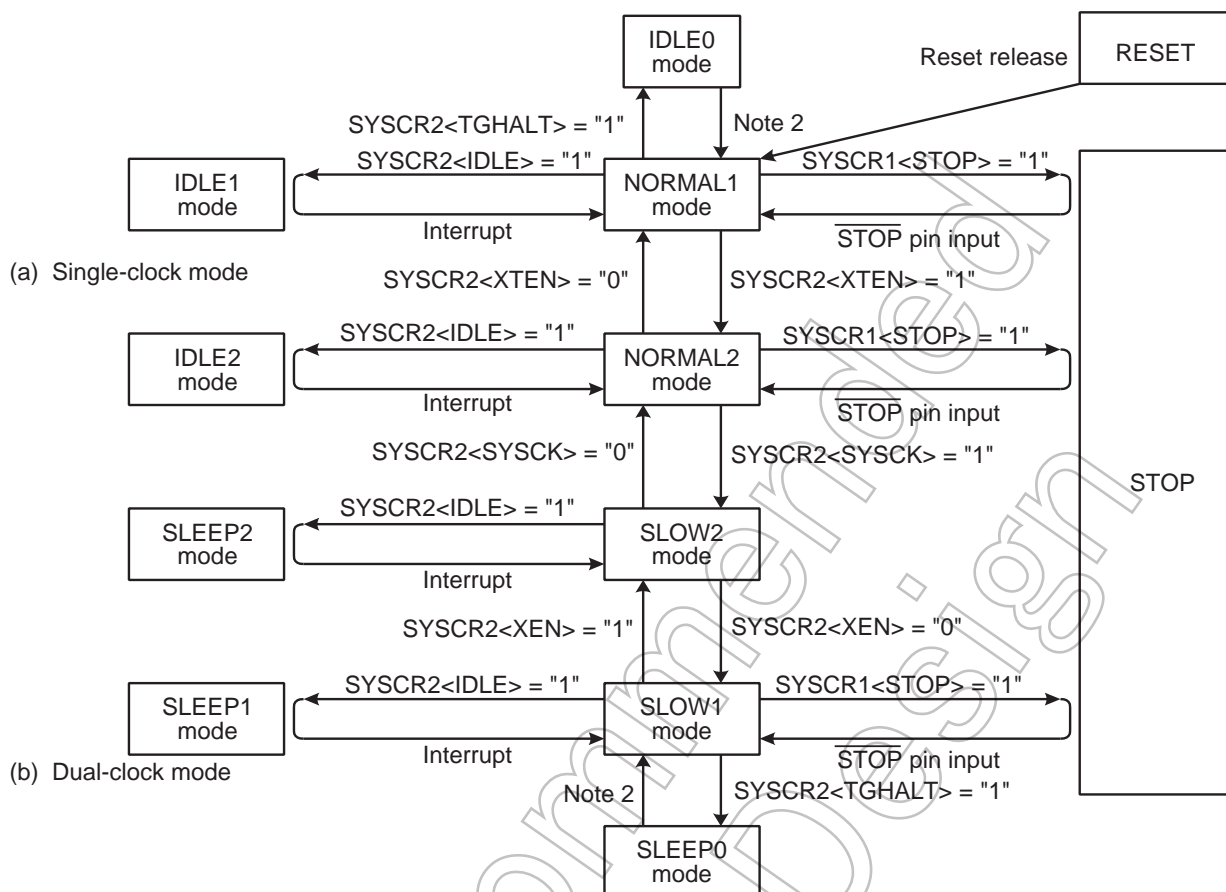
When SLEEP0 mode starts, the CPU stops and the timing generator stops feeding the clock to the peripheral circuits other than TBT. Then, upon detecting the falling edge of the source clock selected with TBTCR<TBTCK>, the timing generator starts feeding the clock to all peripheral circuits.

When returned from SLEEP0 mode, the CPU restarts operating, entering SLOW1 mode back again. SLEEP0 mode is entered and returned regardless of how TBTCR<TBTEN> is set. When IMF = "1", EF6 (TBT interrupt individual enable flag) = "1", and TBTCR<TBTEN> = "1", interrupt processing is performed. When SLEEP0 mode is entered while TBTCR<TBTEN> = "1", the INTTBT interrupt latch is set after returning to SLOW1 mode.

2.2.3.3 STOP mode

In this mode, the internal oscillation circuit is turned off, causing all system operations to be halted. The internal status immediately prior to the halt is held with a lowest power consumption during STOP mode.

STOP mode is started by the system control register 1 (SYSCR1), and STOP mode is released by a inputting (Either level-sensitive or edge-sensitive can be programmably selected) to the STOP pin. After the warm-up period is completed, the execution resumes with the instruction which follows the STOP mode start instruction.



Note 1: NORMAL1 and NORMAL2 modes are generically called NORMAL; SLOW1 and SLOW2 are called SLOW; IDLE0, IDLE1 and IDLE2 are called IDLE; SLEEP0, SLEEP1 and SLEEP2 are called SLEEP.

Note 2: The mode is released by falling edge of TBTCR<TBTCk> setting.

Figure 2-6 Operating Mode Transition Diagram

Table 2-1 Operating Mode and Conditions

Operating Mode		Oscillator		CPU Core	TBT	Other Peripherals	Machine Cycle Time
		High Frequency	Low Frequency				
Single clock	RESET	Oscillation	Stop	Reset	Reset	Reset	4/fc [s]
	NORMAL1			Operate	Operate	Operate	
	IDLE1			Halt		Halt	
	IDLE0						
	STOP	Stop	Halt	Halt	—		
Dual clock	NORMAL2	Oscillation	Oscillation	Operate with high frequency	Operate	Operate	4/fc [s]
	IDLE2			Halt			
	SLOW2			Operate with low frequency			
	SLEEP2			Halt			
	SLOW1	Stop	Stop	Operate with low frequency	Halt	Halt	4/fs [s]
	SLEEP1						
	SLEEP0						
	STOP			Halt			—

System Control Register 1

SYSCR1	7	6	5	4	3	2	1	0	
(0038H)	STOP	RELM	RETM	OUTEN	WUT				(Initial value: 0000 00**)

STOP	STOP mode start	0: CPU core and peripherals remain active 1: CPU core and peripherals are halted (Start STOP mode)			R/W
RELM	Release method for STOP mode	0: Edge-sensitive release 1: Level-sensitive release			R/W
RETM	Operating mode after STOP mode	0: Return to NORMAL 1/2 mode 1: Return to SLOW1 mode			R/W
OUTEN	Port output during STOP mode	0: High impedance 1: Output kept			R/W
WUT	Warm-up time at releasing STOP mode		Return to NORMAL mode	Return to SLOW mode	R/W
		00	$3 \times 2^{16}/f_c$	$3 \times 2^{13}/f_s$	
		01	$2^{16}/f_c$	$2^{13}/f_s$	
		10	$3 \times 2^{14}/f_c$	$3 \times 2^6/f_s$	
		11	$2^{14}/f_c$	$2^6/f_s$	

Note 1: Always set RETM to "0" when transiting from NORMAL mode to STOP mode. Always set RETM to "1" when transiting from SLOW mode to STOP mode.

Note 2: When STOP mode is released with $\overline{\text{RESET}}$ pin input, a return is made to NORMAL1 regardless of the RETM contents.

Note 3: fc: High-frequency clock [Hz], fs: Low-frequency clock [Hz], *: Don't care

Note 4: Bits 1 and 0 in SYSCR1 are read as undefined data when a read instruction is executed.

Note 5: As the hardware becomes STOP mode under OUTEN = "0", input value is fixed to "0"; therefore it may cause external interrupt request on account of falling edge.

Note 6: When the key-on wakeup is used, RELM should be set to "1".

Note 7: Port P20 is used as $\overline{\text{STOP}}$ pin. Therefore, when stop mode is started, OUTEN does not affect to P20, and P20 becomes High-Z mode.

Note 8: The warmig-up time should be set correctly for using oscillator.

System Control Register 2

SYSCR2	7	6	5	4	3	2	1	0	
(0039H)	XEN	XTEN	SYSCK	IDLE	TGHALT				(Initial value: 1000 *0**)

XEN	High-frequency oscillator control	0: Turn off oscillation 1: Turn on oscillation	R/W
XTEN	Low-frequency oscillator control	0: Turn off oscillation 1: Turn on oscillation	
SYSCK	Main system clock select (Write)/main system clock monitor (Read)	0: High-frequency clock (NORMAL1/NORMAL2/IDLE1/IDLE2) 1: Low-frequency clock (SLOW1/SLOW2/SLEEP1/SLEEP2)	
IDLE	CPU and watchdog timer control (IDLE1/2 and SLEEP1/2 modes)	0: CPU and watchdog timer remain active 1: CPU and watchdog timer are stopped (Start IDLE1/2 and SLEEP1/2 modes)	R/W
TGHALT	TG control (IDLE0 and SLEEP0 modes)	0: Feeding clock to all peripherals from TG 1: Stop feeding clock to peripherals except TBT from TG. (Start IDLE0 and SLEEP0 modes)	

Note 1: A reset is applied if both XEN and XTEN are cleared to "0", XEN is cleared to "0" when SYSCK = "0", or XTEN is cleared to "0" when SYSCK = "1".

Note 2: *: Don't care, TG: Timing generator, *: Don't care

Note 3: Bits 3, 1 and 0 in SYSCR2 are always read as undefined value.

Note 4: Do not set IDLE and TGHALT to "1" simultaneously.

Note 5: Because returning from IDLE0/SLEEP0 to NORMAL1/SLOW1 is executed by the asynchronous internal clock, the period of IDLE0/SLEEP0 mode might be shorter than the period setting by $\text{TBTCL} < \text{TBTCK} >$.

Note 6: When IDLE1/2 or SLEEP1/2 mode is released, IDLE is automatically cleared to "0".

Note 7: When IDLE0 or SLEEP0 mode is released, TGHALT is automatically cleared to "0".

Note 8: Before setting TGHALT to "1", be sure to stop peripherals. If peripherals are not stopped, the interrupt latch of peripherals may be set after IDLE0 or SLEEP0 mode is released.

2.2.4 Operating Mode Control

2.2.4.1 STOP mode

STOP mode is controlled by the system control register 1, the $\overline{\text{STOP}}$ pin input and key-on wakeup input (STOP5 to STOP2) which is controlled by the STOP mode release control register (STOPCR).

The $\overline{\text{STOP}}$ pin is also used both as a port P20 and an $\overline{\text{INT5}}$ (external interrupt input 5) pin. STOP mode is started by setting SYSCR1<STOP> to “1”. During STOP mode, the following status is maintained.

1. Oscillations are turned off, and all internal operations are halted.
2. The data memory, registers, the program status word and port output latches are all held in the status in effect before STOP mode was entered.
3. The prescaler and the divider of the timing generator are cleared to “0”.
4. The program counter holds the address 2 ahead of the instruction (e.g., [SET (SYSCR1).7]) which started STOP mode.

STOP mode includes a level-sensitive mode and an edge-sensitive mode, either of which can be selected with the SYSCR1<RELM>. Do not use any key-on wakeup input (STOP5 to STOP2) for releasing STOP mode in edge-sensitive mode.

Note 1: The STOP mode can be released by either the STOP or key-on wakeup pin (STOP5 to STOP2). However, because the STOP pin is different from the key-on wakeup and can not inhibit the release input, the STOP pin must be used for releasing STOP mode.

Note 2: During STOP period (from start of STOP mode to end of warm up), due to changes in the external interrupt pin signal, interrupt latches may be set to “1” and interrupts may be accepted immediately after STOP mode is released. Before starting STOP mode, therefore, disable interrupts. Also, before enabling interrupts after STOP mode is released, clear unnecessary interrupt latches.

(1) Level-sensitive release mode (RELM = “1”)

In this mode, STOP mode is released by setting the $\overline{\text{STOP}}$ pin high or setting the STOP5 to STOP2 pin input which is enabled by STOPCR. This mode is used for capacitor backup when the main power supply is cut off and long term battery backup.

Even if an instruction for starting STOP mode is executed while $\overline{\text{STOP}}$ pin input is high or STOP5 to STOP2 input is low, STOP mode does not start but instead the warm-up sequence starts immediately. Thus, to start STOP mode in the level-sensitive release mode, it is necessary for the program to first confirm that the $\overline{\text{STOP}}$ pin input is low or STOP5 to STOP2 input is high. The following two methods can be used for confirmation.

1. Testing a port.
2. Using an external interrupt input $\overline{\text{INT5}}$ ($\overline{\text{INT5}}$ is a falling edge-sensitive input).

Example 1 :Starting STOP mode from NORMAL mode by testing a port P20.

	LD	(SYSCR1). 01010000B	; Sets up the level-sensitive release mode
SSTOPH:	TEST	(P2PRD). 0	; Wait until the $\overline{\text{STOP}}$ pin input goes low level
	JRS	F, SSTOPH	
	DI		; IMF ← 0
	SET	(SYSCR1). 7	; Starts STOP mode

Example 2 :Starting STOP mode from NORMAL mode with an INT5 interrupt.

```
PINT5:      TEST      (P2PRD). 0      ; To reject noise, STOP mode does not start if
           JRS        F, SINT5        port P20 is at high
           LD         (SYSCR1), 01010000B ; Sets up the level-sensitive release mode.
           DI          ; IMF ← 0
           SET        (SYSCR1). 7      ; Starts STOP mode
SINT5:      RETI
```

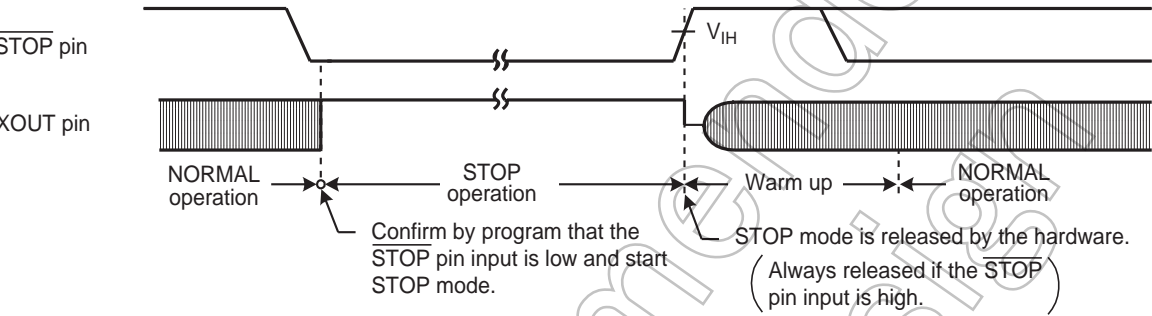


Figure 2-7 Level-sensitive Release Mode

Note 1: Even if the $\overline{\text{STOP}}$ pin input is low after warm-up start, the STOP mode is not restarted.

Note 2: In this case of changing to the level-sensitive mode from the edge-sensitive mode, the release mode is not switched until a rising edge of the $\overline{\text{STOP}}$ pin input is detected.

(2) Edge-sensitive release mode (RELM = “0”)

In this mode, STOP mode is released by a rising edge of the $\overline{\text{STOP}}$ pin input. This is used in applications where a relatively short program is executed repeatedly at periodic intervals. This periodic signal (for example, a clock from a low-power consumption oscillator) is input to the $\overline{\text{STOP}}$ pin. In the edge-sensitive release mode, STOP mode is started even when the $\overline{\text{STOP}}$ pin input is high level. Do not use any STOP5 to STOP2 pin input for releasing STOP mode in edge-sensitive release mode.

Example :Starting STOP mode from NORMAL mode

```
DI          ; IMF ← 0
LD          (SYSCR1), 10010000B ; Starts after specified to the edge-sensitive release mode
```

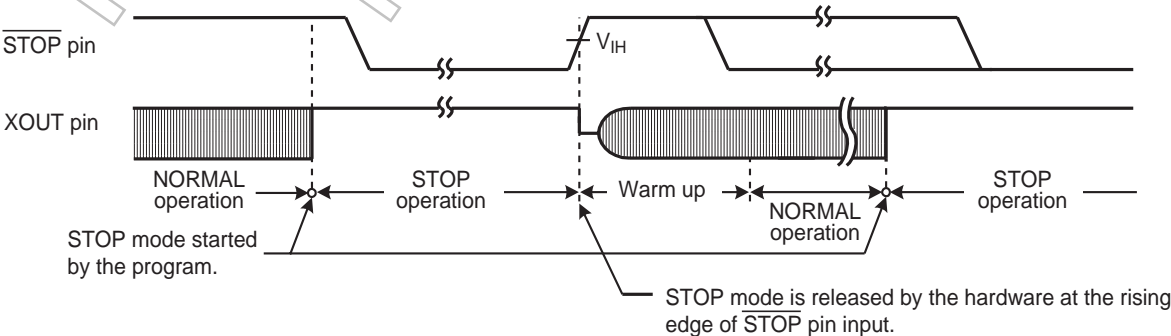


Figure 2-8 Edge-sensitive Release Mode

STOP mode is released by the following sequence.

1. In the dual-clock mode, when returning to NORMAL2, both the high-frequency and low-frequency clock oscillators are turned on; when returning to SLOW1 mode, only the low-frequency clock oscillator is turned on. In the single-clock mode, only the high-frequency clock oscillator is turned on.
2. A warm-up period is inserted to allow oscillation time to stabilize. During warm up, all internal operations remain halted. Four different warm-up times can be selected with the SYSCR1<WUT> in accordance with the resonator characteristics.
3. When the warm-up time has elapsed, normal operation resumes with the instruction following the STOP mode start instruction.

Note 1: When the STOP mode is released, the start is made after the prescaler and the divider of the timing generator are cleared to "0".

Note 2: STOP mode can also be released by inputting low level on the RESET pin, which immediately performs the normal reset operation.

Note 3: When STOP mode is released with a low hold voltage, the following cautions must be observed. The power supply voltage must be at the operating voltage level before releasing STOP mode. The RESET pin input must also be "H" level, rising together with the power supply voltage. In this case, if an external time constant circuit has been connected, the RESET pin input voltage will increase at a slower pace than the power supply voltage. At this time, there is a danger that a reset may occur if input voltage level of the RESET pin drops below the non-inverting high-level input voltage (Hysteresis input).

Table 2-2 Warm-up Time Example (at $f_c = 16.0$ MHz, $f_s = 32.768$ kHz)

WUT	Warm-up Time [ms]	
	Return to NORMAL Mode	Return to SLOW Mode
00	12.288	750
01	4.096	250
10	3.072	5.85
11	1.024	1.95

Note 1: The warm-up time is obtained by dividing the basic clock by the divider. Therefore, the warm-up time may include a certain amount of error if there is any fluctuation of the oscillation frequency when STOP mode is released. Thus, the warm-up time must be considered as an approximate value.

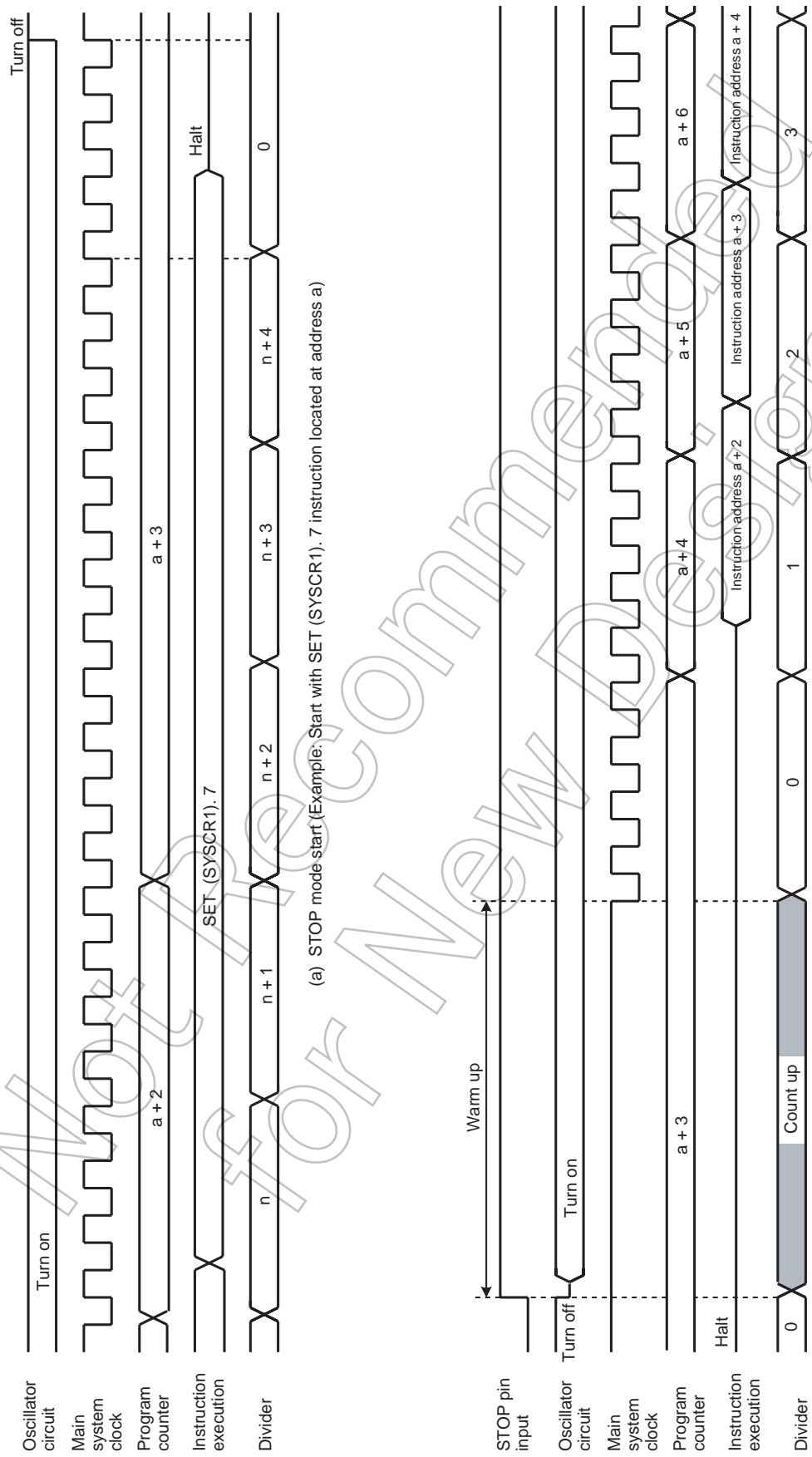


Figure 2-9 STOP Mode Start/Release

2.2.4.2 IDLE1/2 mode and SLEEP1/2 mode

IDLE1/2 and SLEEP1/2 modes are controlled by the system control register 2 (SYSCR2) and maskable interrupts. The following status is maintained during these modes.

1. Operation of the CPU and watchdog timer (WDT) is halted. On-chip peripherals continue to operate.
2. The data memory, CPU registers, program status word and port output latches are all held in the status in effect before these modes were entered.
3. The program counter holds the address 2 ahead of the instruction which starts these modes.

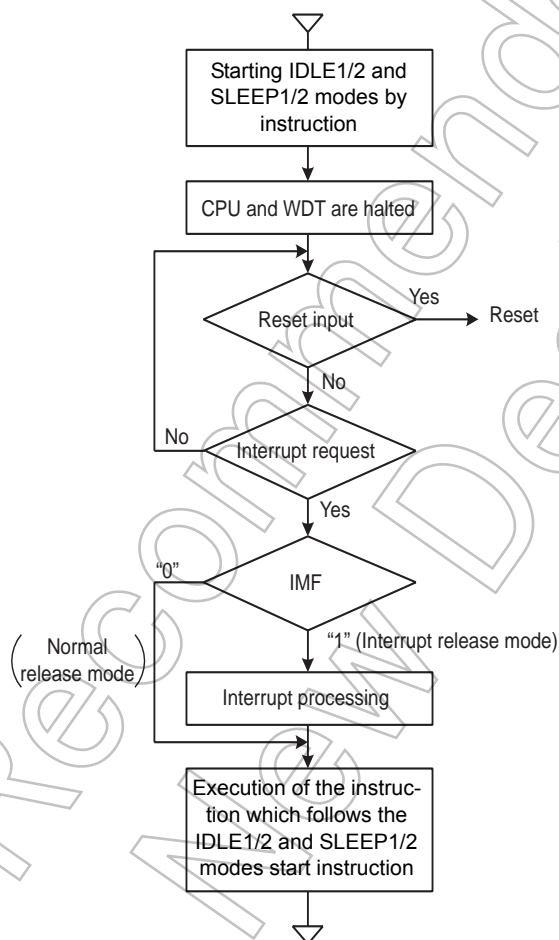


Figure 2-10 IDLE1/2 and SLEEP1/2 Modes

- Start the IDLE1/2 and SLEEP1/2 modes

After IMF is set to "0", set the individual interrupt enable flag (EF) which releases IDLE1/2 and SLEEP1/2 modes. To start IDLE1/2 and SLEEP1/2 modes, set SYSCR2<IDLE> to "1".

- Release the IDLE1/2 and SLEEP1/2 modes

IDLE1/2 and SLEEP1/2 modes include a normal release mode and an interrupt release mode. These modes are selected by interrupt master enable flag (IMF). After releasing IDLE1/2 and SLEEP1/2 modes, the SYSCR2<IDLE> is automatically cleared to "0" and the operation mode is returned to the mode preceding IDLE1/2 and SLEEP1/2 modes.

IDLE1/2 and SLEEP1/2 modes can also be released by inputting low level on the $\overline{\text{RESET}}$ pin. After releasing reset, the operation mode is started from NORMAL1 mode.

(1) Normal release mode (IMF = "0")

IDLE1/2 and SLEEP1/2 modes are released by any interrupt source enabled by the individual interrupt enable flag (EF). After the interrupt is generated, the program operation is resumed from the instruction following the IDLE1/2 and SLEEP1/2 modes start instruction. Normally, the interrupt latches (IL) of the interrupt source used for releasing must be cleared to "0" by load instructions.

(2) Interrupt release mode (IMF = "1")

IDLE1/2 and SLEEP1/2 modes are released by any interrupt source enabled with the individual interrupt enable flag (EF) and the interrupt processing is started. After the interrupt is processed, the program operation is resumed from the instruction following the instruction, which starts IDLE1/2 and SLEEP1/2 modes.

Note: When a watchdog timer interrupts is generated immediately before IDLE1/2 and SLEEP1/2 modes are started, the watchdog timer interrupt will be processed but IDLE1/2 and SLEEP1/2 modes will not be started.

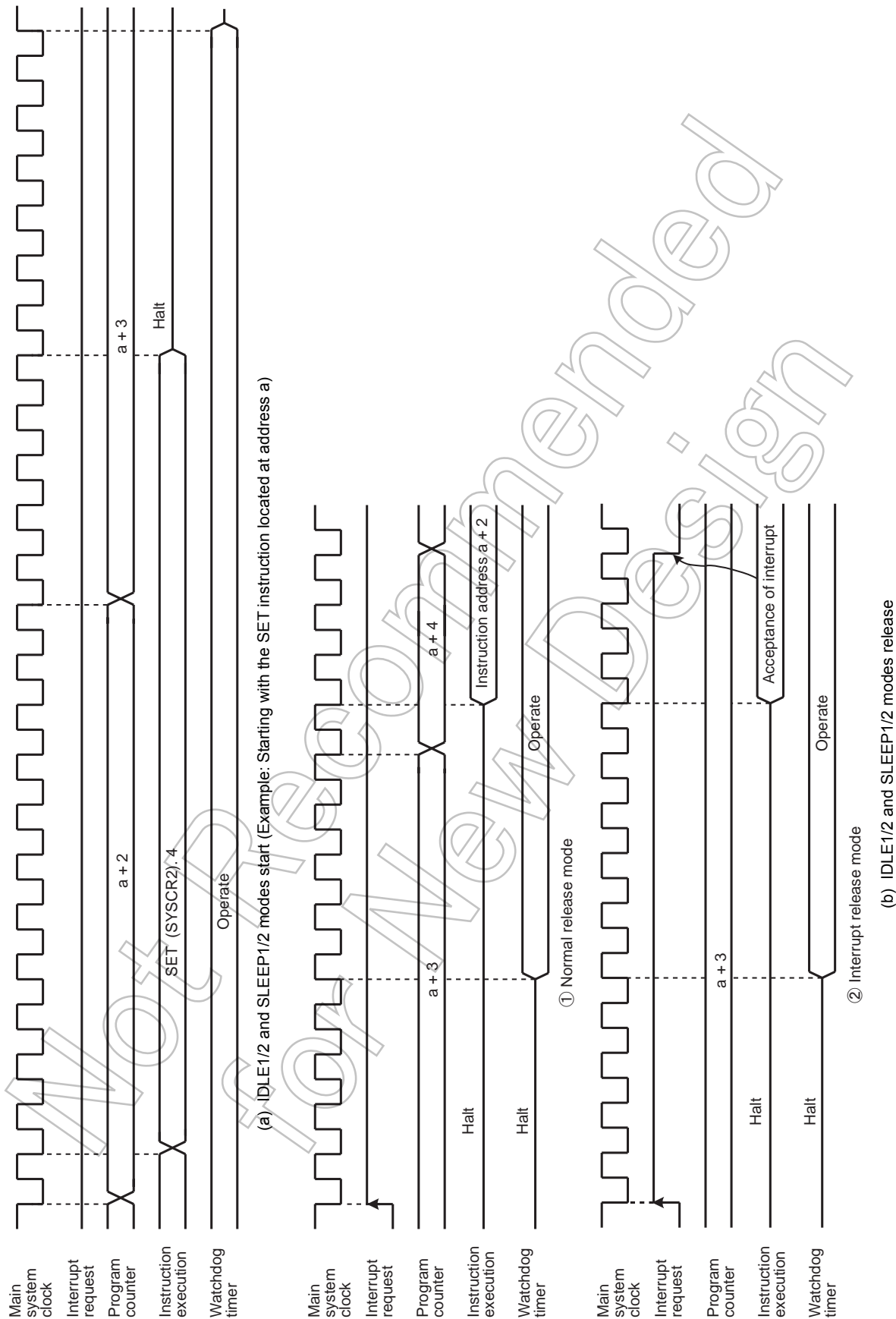


Figure 2-11 IDLE1/2 and SLEEP1/2 Modes Start/Release

2.2.4.3 IDLE0 and SLEEP0 modes (IDLE0, SLEEP0)

IDLE0 and SLEEP0 modes are controlled by the system control register 2 (SYSCR2) and the time base timer control register (TBTCCR). The following status is maintained during IDLE0 and SLEEP0 modes.

1. Timing generator stops feeding clock to peripherals except TBT.
2. The data memory, CPU registers, program status word and port output latches are all held in the status in effect before IDLE0 and SLEEP0 modes were entered.
3. The program counter holds the address 2 ahead of the instruction which starts IDLE0 and SLEEP0 modes.

Note: Before starting IDLE0 or SLEEP0 mode, be sure to stop (Disable) peripherals.

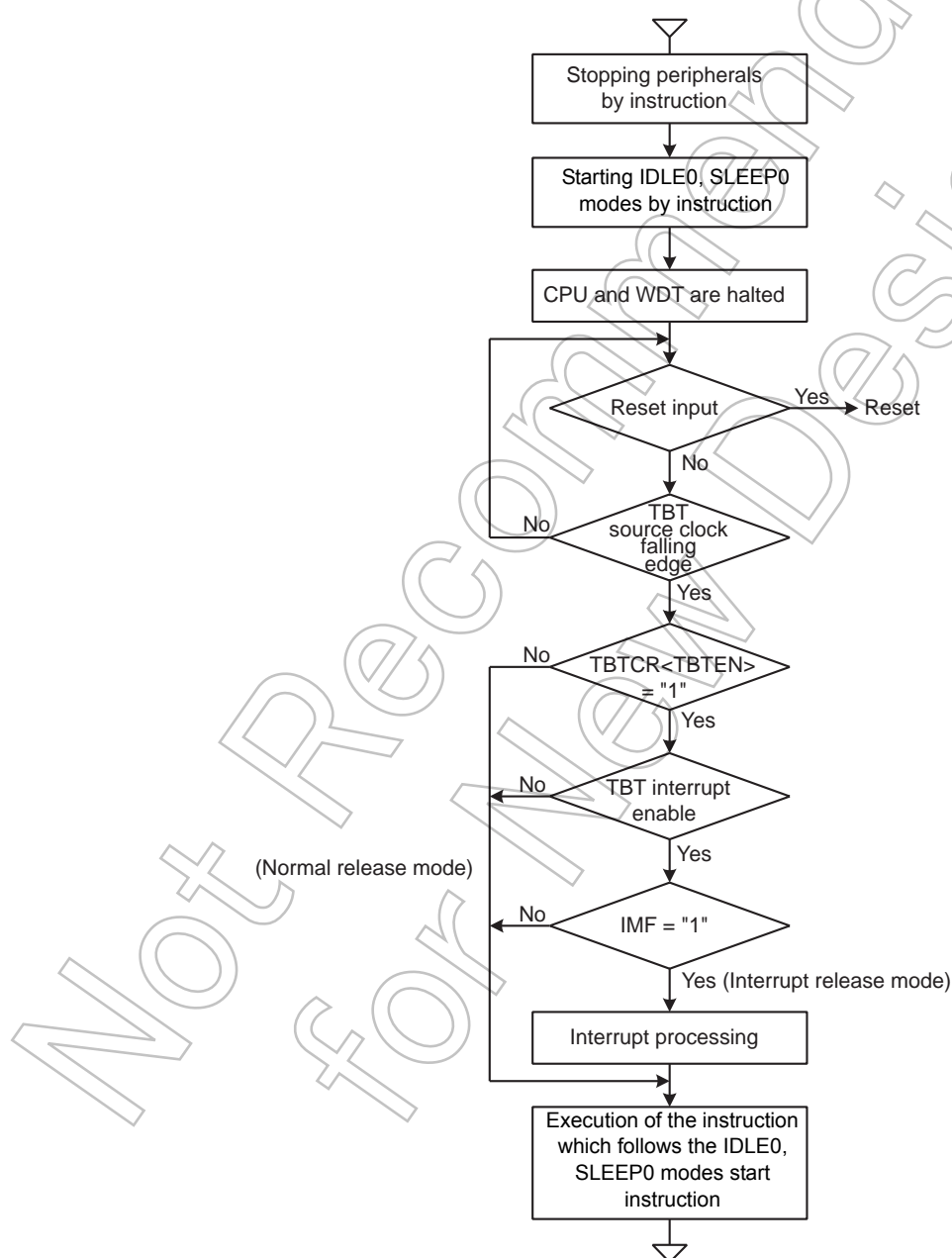


Figure 2-12 IDLE0 and SLEEP0 Modes

- Start the IDLE0 and SLEEP0 modes

Stop (Disable) peripherals such as a timer counter.

To start IDLE0 and SLEEP0 modes, set SYSCR2<TGHALT> to “1”.

- Release the IDLE0 and SLEEP0 modes

IDLE0 and SLEEP0 modes include a normal release mode and an interrupt release mode.

These modes are selected by interrupt master flag (IMF), the individual interrupt enable flag of TBT and TBTCR<TBTEN>.

After releasing IDLE0 and SLEEP0 modes, the SYSCR2<TGHALT> is automatically cleared to “0” and the operation mode is returned to the mode preceding IDLE0 and SLEEP0 modes. Before starting the IDLE0 or SLEEP0 mode, when the TBTCR<TBTEN> is set to “1”, INTTBT interrupt latch is set to “1”.

IDLE0 and SLEEP0 modes can also be released by inputting low level on the $\overline{\text{RESET}}$ pin. After releasing reset, the operation mode is started from NORMAL1 mode.

Note: IDLE0 and SLEEP0 modes start/release without reference to TBTCR<TBTEN> setting.

- (1) Normal release mode (IMF•EF6•TBTCR<TBTEN> = “0”)

IDLE0 and SLEEP0 modes are released by the source clock falling edge, which is setting by the TBTCR<TBTCK>. After the falling edge is detected, the program operation is resumed from the instruction following the IDLE0 and SLEEP0 modes start instruction. Before starting the IDLE0 or SLEEP0 mode, when the TBTCR<TBTEN> is set to “1”, INTTBT interrupt latch is set to “1”.

- (2) Interrupt release mode (IMF•EF6•TBTCR<TBTEN> = “1”)

IDLE0 and SLEEP0 modes are released by the source clock falling edge, which is setting by the TBTCR<TBTCK> and INTTBT interrupt processing is started.

Note 1: Because returning from IDLE0, SLEEP0 to NORMAL1, SLOW1 is executed by the asynchronous internal clock, the period of IDLE0, SLEEP0 mode might be the shorter than the period setting by TBTCR<TBTCK>.

Note 2: When a watchdog timer interrupt is generated immediately before IDLE0/SLEEP0 mode is started, the watchdog timer interrupt will be processed but IDLE0/SLEEP0 mode will not be started.

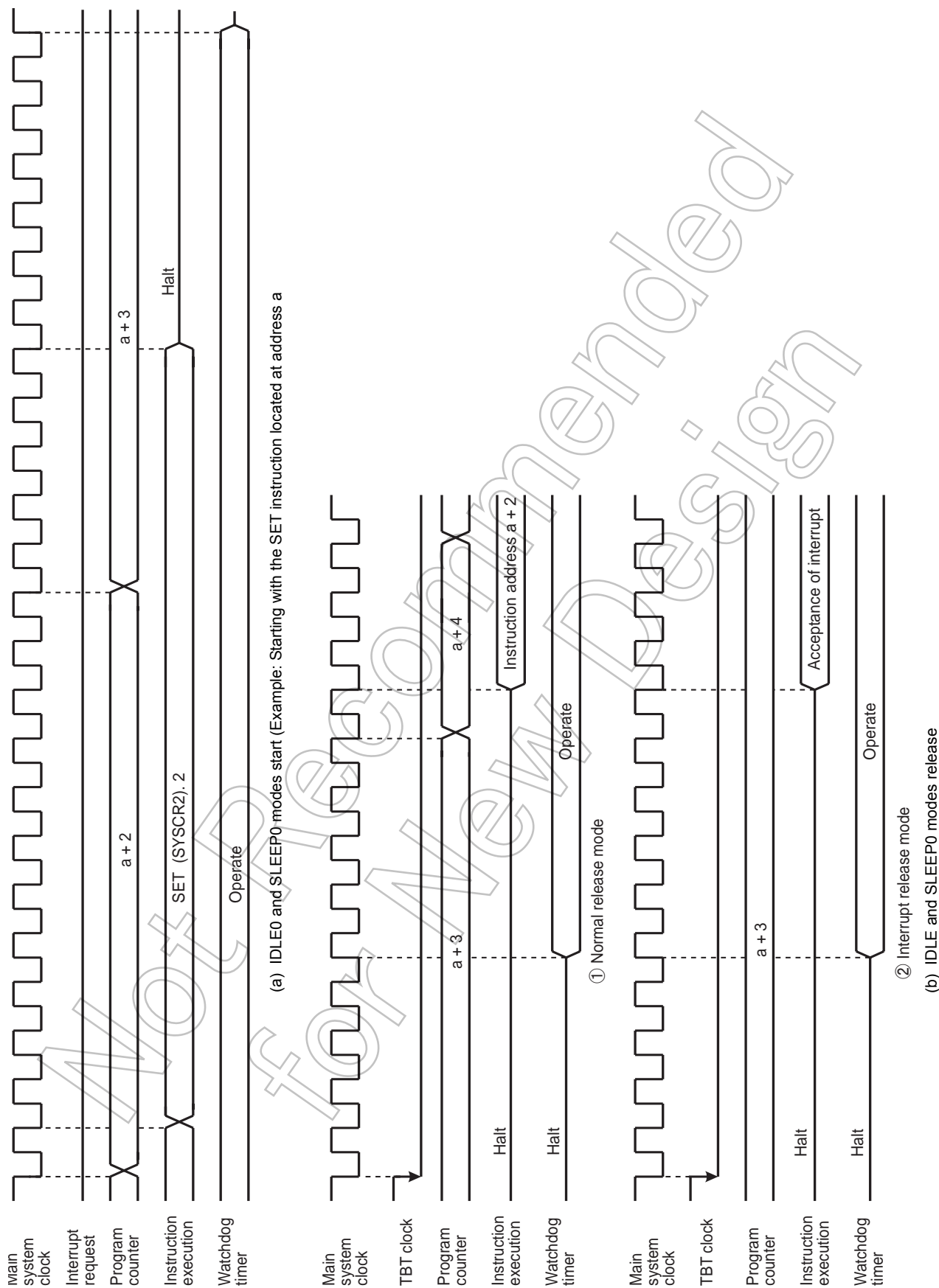


Figure 2-13 IDLE0 and SLEEP0 Modes Start/Release

2.2.4.4 SLOW mode

SLOW mode is controlled by the system control register 2 (SYSCR2).

The following is the methods to switch the mode with the warm-up counter.

(1) Switching from NORMAL2 mode to SLOW1 mode

First, set SYSCR2<SYSCK> to switch the main system clock to the low-frequency clock for SLOW2 mode. Next, clear SYSCR2<XEN> to turn off high-frequency oscillation.

Note: The high-frequency clock can be continued oscillation in order to return to NORMAL2 mode from SLOW mode quickly. Always turn off oscillation of high-frequency clock when switching from SLOW mode to stop mode.

Example 1 :Switching from NORMAL2 mode to SLOW1 mode.

```
SET      (SYSCR2). 5      ; SYSCR2<SYSCK> ← 1
                          ; (Switches the main system clock to the low-frequency
                          ; clock for SLOW2)

CLR      (SYSCR2). 7      ; SYSCR2<XEN> ← 0
                          ; (Turns off high-frequency oscillation)
```

Example 2 :Switching to the SLOW1 mode after low-frequency clock has stabilized.

```
SET      (SYSCR2). 6      ; SYSCR2<XTEN> ← 1

LD       (TC3CR), 43H     ; Sets mode for TC4, 3 (16-bit mode, fs for source)

LD       (TC4CR), 05H     ; Sets warming-up counter mode

LDW      (TTREG3), 8000H   ; Sets warm-up time (Depend on oscillator accompanied)

DI       ; IMF ← 0

SET      (EIRH). 3        ; Enables INTTC4

EI       ; IMF ← 1

SET      (TC4CR). 3       ; Starts TC4, 3

:

PINTTC4: CLR      (TC4CR). 3 ; Stops TC4, 3

SET      (SYSCR2). 5      ; SYSCR2<SYSCK> ← 1
                          ; (Switches the main system clock to the low-frequency clock)

CLR      (SYSCR2). 7      ; SYSCR2<XEN> ← 0
                          ; (Turns off high-frequency oscillation)

RETI

:

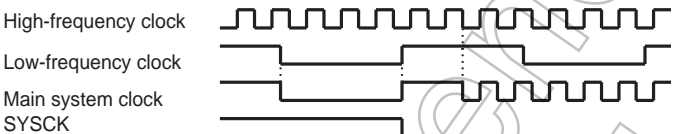
VINTTC4: DW      PINTTC4   ; INTTC4 vector table
```

(2) Switching from SLOW1 mode to NORMAL2 mode

First, set SYSCR2<XEN> to turn on the high-frequency oscillation. When time for stabilization (Warm up) has been taken by the timer/counter (TC4,TC3), clear SYSCR2<SYSCK> to switch the main system clock to the high-frequency clock.

SLOW mode can also be released by inputting low level on the $\overline{\text{RESET}}$ pin. After releasing reset, the operation mode is started from NORMAL1 mode.

Note: After SYSCK is cleared to "0", executing the instructions is continued by the low-frequency clock for the period synchronized with low-frequency and high-frequency clocks.



Example :Switching from the SLOW1 mode to the NORMAL2 mode (fc = 16 MHz, warm-up time is 4.0 ms).

```
SET      (SYSCR2). 7      ; SYSCR2<XEN> ← 1 (Starts high-frequency oscillation)

LD       (TC3CR), 63H      ; Sets mode for TC4, 3 (16-bit mode, fc for source)

LD       (TC4CR), 05H      ; Sets warming-up counter mode

LD       (TTREG4), 0F8H    ; Sets warm-up time

DI       ; IMF ← 0

SET      (EIRH). 3        ; Enables INTTC4

EI       ; IMF ← 1

SET      (TC4CR). 3        ; Starts TC4, 3
:

PINTTC4: CLR      (TC4CR). 3 ; Stops TC4, 3

          CLR      (SYSCR2). 5 ; SYSCR2<SYSCK> ← 0
                               ; (Switches the main system clock to the high-frequency clock)

          RETI

:

VINTTC4: DW      PINTTC4    ; INTTC4 vector table
```

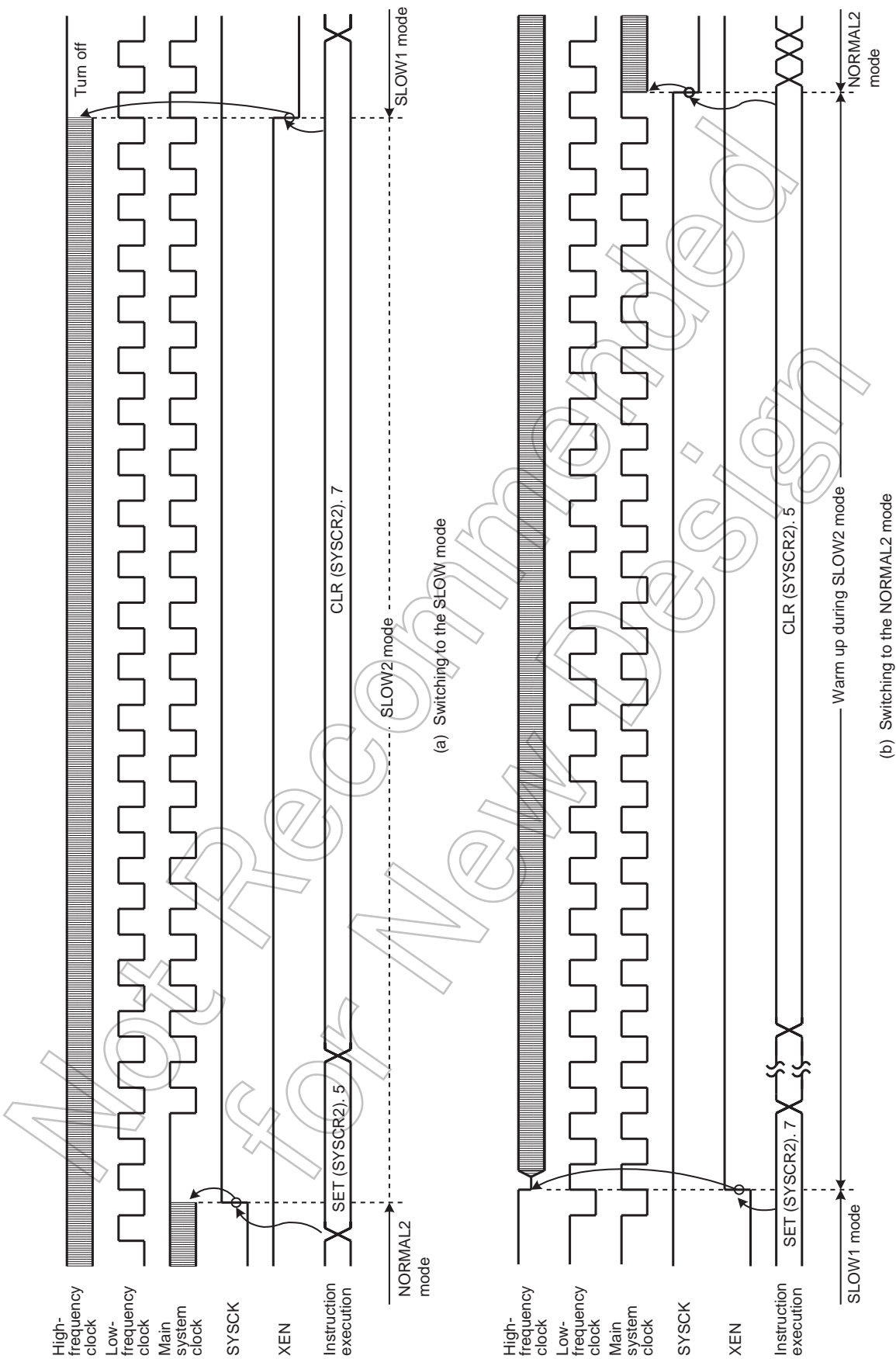


Figure 2-14 Switching between the NORMAL2 and SLOW Modes

2.3 Reset Circuit

The TMP86P820FG has four types of reset generation procedures: An external reset input, an address trap reset, a watchdog timer reset and a system clock reset. Of these reset, the address trap reset, the watchdog timer and the system clock reset are a malfunction reset. When the malfunction reset request is detected, reset occurs during the maximum $24/f_c[s]$ (The $\overline{\text{RESET}}$ pin outputs "L" level).

The malfunction reset circuit such as watchdog timer reset, address trap reset and system clock reset is not initialized when power is turned on. Therefore, reset may occur during maximum $24/f_c[s]$ ($1.5\mu s$ at 16.0 MHz) when power is turned on. $\overline{\text{RESET}}$ pin outputs "L" level during maximum $24/f_c[s]$ ($1.5\mu s$ at 16.0MHz).

Table 2-3 shows on-chip hardware initialization by reset action.

Table 2-3 Initializing Internal Status by Reset Action

On-chip Hardware	Initial Value	On-chip Hardware	Initial Value
Program counter (PC)	(FFFEH)	Prescaler and divider of timing generator	0
Stack pointer (SP)	Not initialized		
General-purpose registers (W, A, B, C, D, E, H, L, IX, IY)	Not initialized		
Jump status flag (JF)	Not initialized	Watchdog timer	Enable
Zero flag (ZF)	Not initialized	Output latches of I/O ports	Refer to I/O port circuitry
Carry flag (CF)	Not initialized		
Half carry flag (HF)	Not initialized		
Sign flag (SF)	Not initialized		
Overflow flag (VF)	Not initialized		
Interrupt master enable flag (IMF)	0		
Interrupt individual enable flags (EF)	0	Control registers	Refer to each of control register
Interrupt latches (IL)	0		
		LCD data buffer	Not initialized
		RAM	Not initialized

2.3.1 External Reset Input

The $\overline{\text{RESET}}$ pin contains a Schmitt trigger (Hysteresis) with an internal pull-up resistor.

When the $\overline{\text{RESET}}$ pin is held at "L" level for at least 3 machine cycles ($12/f_c[s]$) with the power supply voltage within the operating voltage range and oscillation stable, a reset is applied and the internal state is initialized.

When the $\overline{\text{RESET}}$ pin input goes high, the reset operation is released and the program execution starts at the vector address stored at addresses FFFEh to FFFFh.

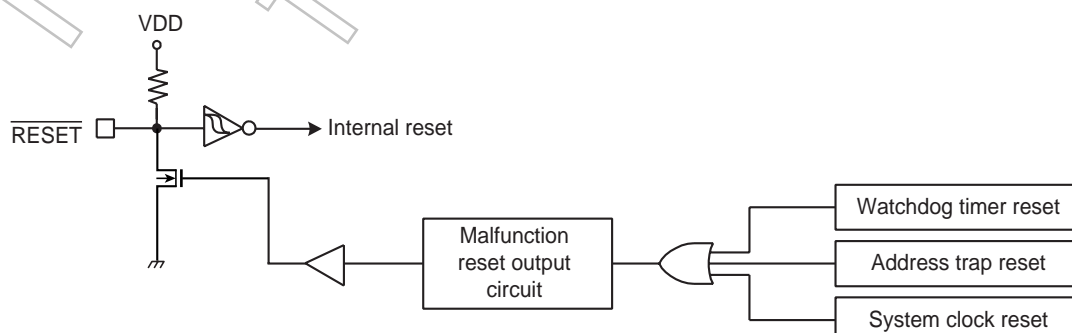
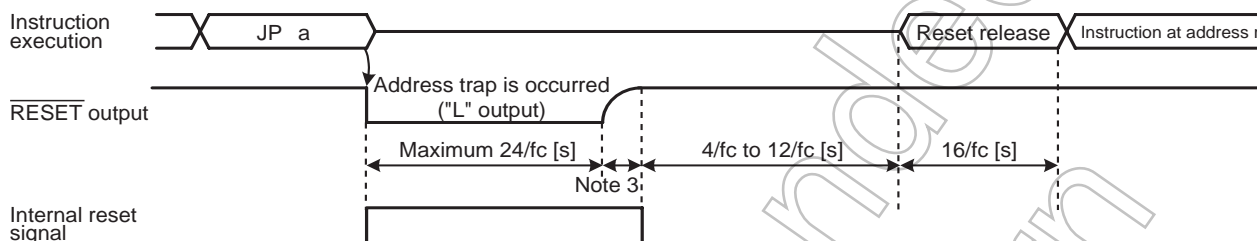


Figure 2-15 Reset Circuit

2.3.2 Address trap reset

If the CPU should start looping for some cause such as noise and an attempt be made to fetch an instruction from the on-chip RAM (when $WDTCR1<ATAS>$ is set to "1"), DBR or the SFR area, address trap reset will be generated. The reset time is maximum $24/f_c[s]$ ($1.5\mu s$ at 16.0 MHz). Then, the \overline{RESET} pin outputs "L" level during maximum $24/f_c[s]$.

Note: The operating mode under address trapped is alternative of reset or interrupt. The address trap area is alternative.



Note 1: Address "a" is in the SFR, DBR or on-chip RAM ($WDTCR1<ATAS> = "1"$) space.

Note 2: During reset release, reset vector "r" is read out, and an instruction at address "r" is fetched and decoded.

Note 3: Varies on account of external condition: voltage or capacitance

Figure 2-16 Address Trap Reset

2.3.3 Watchdog timer reset

Refer to Section "Watchdog Timer".

2.3.4 System clock reset

If the condition as follows is detected, the system clock reset occurs automatically to prevent dead lock of the CPU. (The oscillation is continued without stopping.)

- In case of clearing $SYSCR2<XEN>$ and $SYSCR2<XTEN>$ simultaneously to "0".
- In case of clearing $SYSCR2<XEN>$ to "0", when the $SYSCR2<SYSCK>$ is "0".
- In case of clearing $SYSCR2<XTEN>$ to "0", when the $SYSCR2<SYSCK>$ is "1".

The reset time is maximum $24/f_c$ ($1.5\mu s$ at 16.0 MHz). Then, the \overline{RESET} pin outputs "L" level during maximum $24/f_c[s]$ ($1.5\mu s$ at 16.0MHz).

Not Recommended
for New Design

Not Recommended
for New Design

3. Interrupt Control Circuit

The TMP86P820FG has a total of 15 interrupt sources excluding reset, of which 1 source levels are multiplexed. Interrupts can be nested with priorities. Four of the internal interrupt sources are non-maskable while the rest are maskable.

Interrupt sources are provided with interrupt latches (IL), which hold interrupt requests, and independent vectors. The interrupt latch is set to "1" by the generation of its interrupt request which requests the CPU to accept its interrupts. Interrupts are enabled or disabled by software using the interrupt master enable flag (IMF) and interrupt enable flag (EF). If more than one interrupts are generated simultaneously, interrupts are accepted in order which is dominated by hardware. However, there are no prioritized interrupt factors among non-maskable interrupts.

Interrupt Factors		Enable Condition	Interrupt Latch	Vector Address	Priority
Internal/External	(Reset)	Non-maskable	–	FFFE	1
Internal	INTSWI (Software interrupt)	Non-maskable	–	FFFC	2
Internal	INTUNDEF (Executed the undefined instruction interrupt)	Non-maskable	–	FFFC	2
Internal	INTATRAP (Address trap interrupt)	Non-maskable	IL2	FFFA	2
Internal	INTWDT (Watchdog timer interrupt)	Non-maskable	IL3	FFF8	2
External	INT0	IMF•EF4 = 1, INT0EN = 1	IL4	FFF6	5
External	INT1	IMF•EF5 = 1	IL5	FFF4	6
Internal	INTTBT	IMF•EF6 = 1	IL6	FFF2	7
External	INT2	IMF•EF7 = 1	IL7	FFF0	8
Internal	INTTC1	IMF•EF8 = 1	IL8	FFEE	9
Internal	INTSIO	IMF•EF9 = 1	IL9	FFEC	10
-	Reserved	IMF•EF10 = 1	IL10	FFEA	11
Internal	INTTC4	IMF•EF11 = 1	IL11	FFE8	12
-	Reserved	IMF•EF12 = 1	IL12	FFE6	13
Internal	INTADC	IMF•EF13 = 1	IL13	FFE4	14
External	INT3	IMF•EF14 = 1, IL14ER = 0	IL14	FFE2	15
Internal	INTTC3	IMF•EF14 = 1, IL14ER = 1			
External	INT5	IMF•EF15 = 1	IL15	FFE0	16

Note 1: The INTSEL register is used to select the interrupt source to be enabled for each multiplexed source level (see 3.3 Interrupt Source Selector (INTSEL)).

Note 2: To use the address trap interrupt (INTATRAP), clear WDTCR1<ATOUT> to "0" (It is set for the "reset request" after reset is cancelled). For details, see "Address Trap".

Note 3: To use the watchdog timer interrupt (INTWDT), clear WDTCR1<WDTOUT> to "0" (It is set for the "Reset request" after reset is released). For details, see "Watchdog Timer".

3.1 Interrupt latches (IL15 to IL2)

An interrupt latch is provided for each interrupt source, except for a software interrupt and an executed the undefined instruction interrupt. When interrupt request is generated, the latch is set to "1", and the CPU is requested to accept the interrupt if its interrupt is enabled. The interrupt latch is cleared to "0" immediately after accepting interrupt. All interrupt latches are initialized to "0" during reset.

The interrupt latches are located on address 003CH and 003DH in SFR area. Each latch can be cleared to "0" individually by instruction. However, IL2 and IL3 should not be cleared to "0" by software. For clearing the interrupt latch, load instruction should be used and then IL2 and IL3 should be set to "1". If the read-modify-write instructions such as bit manipulation or operation instructions are used, interrupt request would be cleared inadequately if interrupt is requested while such instructions are executed.

Interrupt latches are not set to "1" by an instruction.

Since interrupt latches can be read, the status for interrupt requests can be monitored by software.

Note: In main program, before manipulating the interrupt enable flag (EF) or the interrupt latch (IL), be sure to clear IMF to "0" (Disable interrupt by DI instruction). Then set IMF newly again as required after operating on the EF or IL (Enable interrupt by EI instruction)
In interrupt service routine, because the IMF becomes "0" automatically, clearing IMF need not execute normally on interrupt service routine. However, if using multiple interrupt on interrupt service routine, manipulating EF or IL should be executed before setting IMF="1".

Example 1 :Clears interrupt latches

```
DI                                ; IMF ← 0
LDW      (ILL), 1110100000111111B ; IL12, IL10 to IL6 ← 0
EI                                ; IMF ← 1
```

Example 2 :Reads interrupt latchess

```
LD      WA, (ILL)                ; W ← ILH, A ← ILL
```

Example 3 :Tests interrupt latches

```
TEST      (ILL). 7                ; if IL7 = 1 then jump
JR        F, SSET
```

3.2 Interrupt enable register (EIR)

The interrupt enable register (EIR) enables and disables the acceptance of interrupts, except for the non-maskable interrupts (Software interrupt, undefined instruction interrupt, address trap interrupt and watchdog interrupt). Non-maskable interrupt is accepted regardless of the contents of the EIR.

The EIR consists of an interrupt master enable flag (IMF) and the individual interrupt enable flags (EF). These registers are located on address 003AH and 003BH in SFR area, and they can be read and written by an instructions (Including read-modify-write instructions such as bit manipulation or operation instructions).

3.2.1 Interrupt master enable flag (IMF)

The interrupt enable register (IMF) enables and disables the acceptance of the whole maskable interrupt. While IMF = "0", all maskable interrupts are not accepted regardless of the status on each individual interrupt enable flag (EF). By setting IMF to "1", the interrupt becomes acceptable if the individuals are enabled. When an interrupt is accepted, IMF is cleared to "0" after the latest status on IMF is stacked. Thus the maskable interrupts which follow are disabled. By executing return interrupt instruction [RETI/RETN], the stacked data, which was the status before interrupt acceptance, is loaded on IMF again.

The IMF is located on bit0 in EIRL (Address: 003AH in SFR), and can be read and written by an instruction. The IMF is normally set and cleared by [EI] and [DI] instruction respectively. During reset, the IMF is initialized to "0".

3.2.2 Individual interrupt enable flags (EF15 to EF4)

Each of these flags enables and disables the acceptance of its maskable interrupt. Setting the corresponding bit of an individual interrupt enable flag to "1" enables acceptance of its interrupt, and setting the bit to "0" disables acceptance. During reset, all the individual interrupt enable flags (EF15 to EF4) are initialized to "0" and all maskable interrupts are not accepted until they are set to "1".

Note: In main program, before manipulating the interrupt enable flag (EF) or the interrupt latch (IL), be sure to clear IMF to "0" (Disable interrupt by DI instruction). Then set IMF newly again as required after operating on the EF or IL (Enable interrupt by EI instruction)
In interrupt service routine, because the IMF becomes "0" automatically, clearing IMF need not execute normally on interrupt service routine. However, if using multiple interrupt on interrupt service routine, manipulating EF or IL should be executed before setting IMF="1".

Example 1 :Enables interrupts individually and sets IMF

```
DI                                     ; IMF ← 0

LDW      (EIRL), 1110100010100000B   ; EF15 to EF13, EF11, EF7, EF5 ← 1
:                                     Note: IMF should not be set.
:
EI                                     ; IMF ← 1
```

Example 2 :C compiler description example

```
unsigned int _io (3AH) EIRL;          /* 3AH shows EIRL address */
_Di();
EIRL = 10100000B;
:
_Ei();
```

Interrupt Latches

(Initial value: 00000000 000000**)

ILH,ILL (003DH, 003CH)	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
	IL15	IL14	IL13	IL12	IL11	IL10	IL9	IL8	IL7	IL6	IL5	IL4	IL3	IL2		

ILH (003DH)								ILL (003CH)							
IL15 to IL2		Interrupt latches		at RD 0: No interrupt request 1: Interrupt request				at WR 0: Clears the interrupt request 1: (Interrupt latch is not set.)				R/W			

- Note 1: To clear any one of bits IL7 to IL4, be sure to write "1" into IL2 and IL3.
- Note 2: In main program, before manipulating the interrupt enable flag (EF) or the interrupt latch (IL), be sure to clear IMF to "0" (Disable interrupt by DI instruction). Then set IMF newly again as required after operating on the EF or IL (Enable interrupt by EI instruction)
In interrupt service routine, because the IMF becomes "0" automatically, clearing IMF need not execute normally on interrupt service routine. However, if using multiple interrupt on interrupt service routine, manipulating EF or IL should be executed before setting IMF="1".
- Note 3: Do not clear IL with read-modify-write instructions such as bit operations.

Interrupt Enable Registers

(Initial value: 00000000 0000***)

EIRH,EIRL (003BH, 003AH)	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
	EF15	EF14	EF13	EF12	EF11	EF10	EF9	EF8	EF7	EF6	EF5	EF4				IMF

EIRH (003BH) EIRL (003AH)

EF15 to EF4	Individual-interrupt enable flag (Specified for each bit)	0: Disables the acceptance of each maskable interrupt. 1: Enables the acceptance of each maskable interrupt.	R/W
IMF	Interrupt-master enable flag	0: Disables the acceptance of all maskable interrupts 1: Enables the acceptance of all maskable interrupts	

- Note 1: *: Don't care
- Note 2: Do not set IMF and the interrupt enable flag (EF15 to EF4) to "1" at the same time.
- Note 3: In main program, before manipulating the interrupt enable flag (EF) or the interrupt latch (IL), be sure to clear IMF to "0" (Disable interrupt by DI instruction). Then set IMF newly again as required after operating on the EF or IL (Enable interrupt by EI instruction)
In interrupt service routine, because the IMF becomes "0" automatically, clearing IMF need not execute normally on interrupt service routine. However, if using multiple interrupt on interrupt service routine, manipulating EF or IL should be executed before setting IMF="1".

3.3 Interrupt Source Selector (INTSEL)

Each interrupt source that shares the interrupt source level with another interrupt source is allowed to enable the interrupt latch only when it is selected in the INTSEL register. The interrupt controller does not hold interrupt requests corresponding to interrupt sources that are not selected in the INTSEL register. Therefore, the INTSEL register must be set appropriately before interrupt requests are generated.

The following interrupt sources share their interrupt source level; the source is selected on the register INTSEL.

- 1. INT3 and INTTC3 share the interrupt source level whose priority is 15.

Interrupt source selector

INTSEL (003EH)	7	6	5	4	3	2	1	0	
	-	-	-	-	-	-	IL14ER	-	(Initial value: **** **0*)

IL14ER	Selects INT3 or INTTC3	0: INT3 1: INTTC3	R/W
--------	------------------------	----------------------	-----

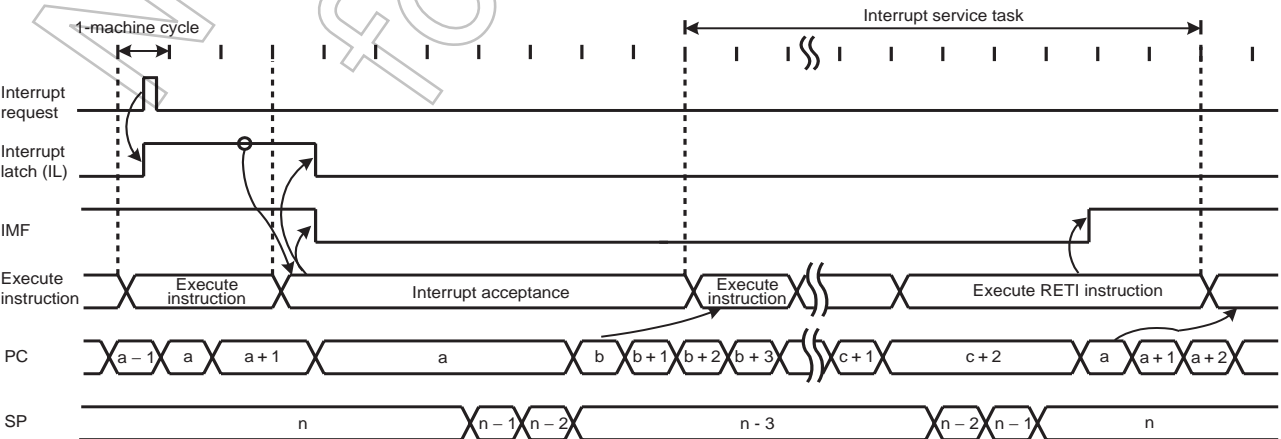
3.4 Interrupt Sequence

An interrupt request, which raised interrupt latch, is held, until interrupt is accepted or interrupt latch is cleared to “0” by resetting or an instruction. Interrupt acceptance sequence requires 8 machine cycles (2 μ s @16 MHz) after the completion of the current instruction. The interrupt service task terminates upon execution of an interrupt return instruction [RETI] (for maskable interrupts) or [RETN] (for non-maskable interrupts). Figure 3-1 shows the timing chart of interrupt acceptance processing.

3.4.1 Interrupt acceptance processing is packaged as follows.

- a. The interrupt master enable flag (IMF) is cleared to “0” in order to disable the acceptance of any following interrupt.
- b. The interrupt latch (IL) for the interrupt source accepted is cleared to “0”.
- c. The contents of the program counter (PC) and the program status word, including the interrupt master enable flag (IMF), are saved (Pushed) on the stack in sequence of PSW + IMF, PCH, PCL. Meanwhile, the stack pointer (SP) is decremented by 3.
- d. The entry address (Interrupt vector) of the corresponding interrupt service program, loaded on the vector table, is transferred to the program counter.
- e. The instruction stored at the entry address of the interrupt service program is executed.

Note: When the contents of PSW are saved on the stack, the contents of IMF are also saved.



Note 1: a: Return address entry address, b: Entry address, c: Address which RETI instruction is stored
Note 2: On condition that interrupt is enabled, it takes 38/fc [s] or 38/fs [s] at maximum (If the interrupt latch is set at the first machine cycle on 10 cycle instruction) to start interrupt acceptance processing since its interrupt latch is set.

Figure 3-1 Timing Chart of Interrupt Acceptance/Return Interrupt Instruction

Example: Correspondence between vector table address for INTTBT and the entry address of the interrupt service program

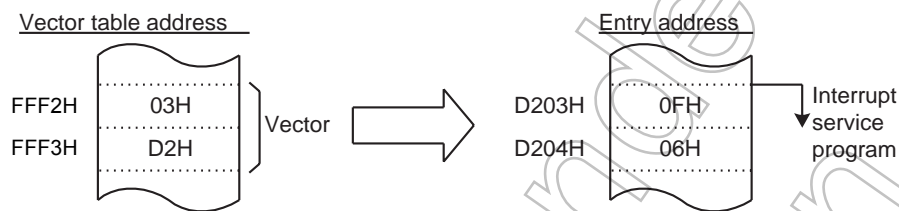


Figure 3-2 Vector table address,Entry address

A maskable interrupt is not accepted until the IMF is set to “1” even if the maskable interrupt higher than the level of current servicing interrupt is requested.

In order to utilize nested interrupt service, the IMF is set to “1” in the interrupt service program. In this case, acceptable interrupt sources are selectively enabled by the individual interrupt enable flags.

To avoid overloaded nesting, clear the individual interrupt enable flag whose interrupt is currently serviced, before setting IMF to “1”. As for non-maskable interrupt, keep interrupt service shorten compared with length between interrupt requests; otherwise the status cannot be recovered as non-maskable interrupt would simply nested.

3.4.2 Saving/restoring general-purpose registers

During interrupt acceptance processing, the program counter (PC) and the program status word (PSW, includes IMF) are automatically saved on the stack, but the accumulator and others are not. These registers are saved by software if necessary. When multiple interrupt services are nested, it is also necessary to avoid using the same data memory area for saving registers. The following methods are used to save/restore the general-purpose registers.

3.4.2.1 Using PUSH and POP instructions

If only a specific register is saved or interrupts of the same source are nested, general-purpose registers can be saved/restored using the PUSH/POP instructions.

Example :Save/store register using PUSH and POP instructions

```
PINTxx:      PUSH      WA           ; Save WA register
              (interrupt processing)
              POP       WA          ; Restore WA register
              RETI                ; RETURN
```

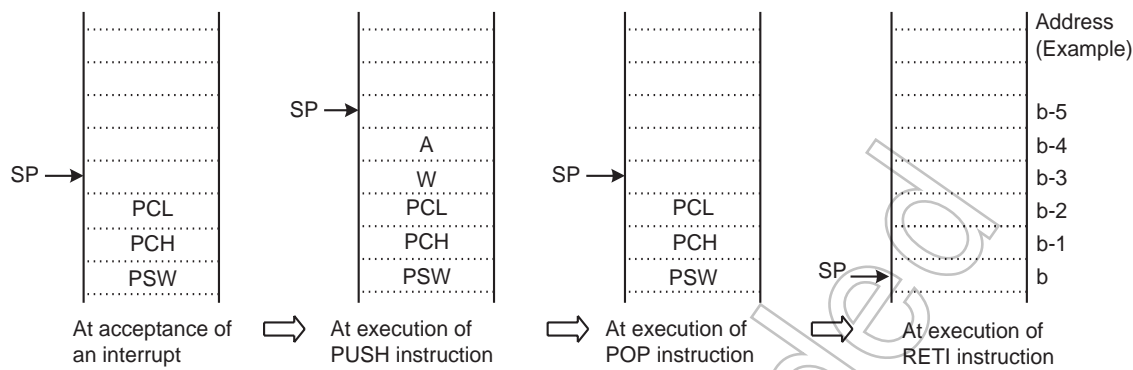


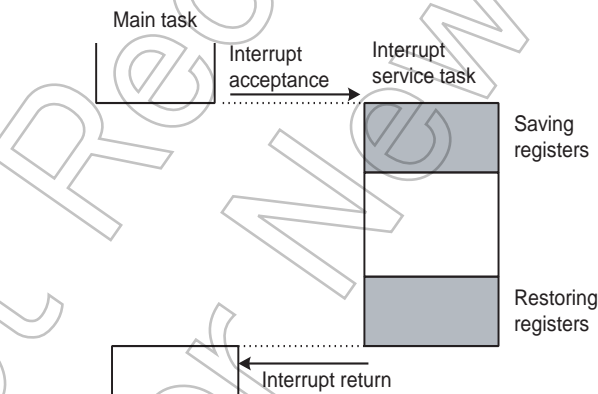
Figure 3-3 Save/store register using PUSH and POP instructions

3.4.2.2 Using data transfer instructions

To save only a specific register without nested interrupts, data transfer instructions are available.

Example :Save/store register using data transfer instructions

```
PINTxx:    LD      (GSAVA), A      ; Save A register
           (interrupt processing)
           LD      A, (GSAVA)     ; Restore A register
           RETI                   ; RETURN
```



Saving/Restoring general-purpose registers using PUSH/POP data transfer instruction

Figure 3-4 Saving/Restoring General-purpose Registers under Interrupt Processing

3.4.3 Interrupt return

Interrupt return instructions [RETI]/[RETN] perform as follows.

[RETI]/[RETN] Interrupt Return
1. Program counter (PC) and program status word (PSW, includes IMF) are restored from the stack.
2. Stack pointer (SP) is incremented by 3.

As for address trap interrupt (INTATRAP), it is required to alter stacked data for program counter (PC) to restarting address, during interrupt service program.

Note: If [RETN] is executed with the above data unaltered, the program returns to the address trap area and INTATRAP occurs again. When interrupt acceptance processing has completed, stacked data for PCL and PCH are located on address (SP + 1) and (SP + 2) respectively.

Example 1 :Returning from address trap interrupt (INTATRAP) service program

```
PINTxx:  POP      WA          ; Recover SP by 2
          LD       WA, Return Address ;
          PUSH     WA          ; Alter stacked data
          (interrupt processing)
          RETN                ; RETURN
```

Example 2 :Restarting without returning interrupt
(In this case, PSW (Includes IMF) before interrupt acceptance is discarded.)

```
PINTxx:  INC      SP          ; Recover SP by 3
          INC      SP
          INC      SP
          (interrupt processing)
          LD       EIRL, data    ; Set IMF to "1" or clear it to "0"
          JP       Restart Address ; Jump into restarting address
```

Interrupt requests are sampled during the final cycle of the instruction being executed. Thus, the next interrupt can be accepted immediately after the interrupt return instruction is executed.

Note 1: It is recommended that stack pointer be return to rate before INTATRAP (Increment 3 times), if return interrupt instruction [RETN] is not utilized during interrupt service program under INTATRAP (such as Example 2).

Note 2: When the interrupt processing time is longer than the interrupt request generation time, the interrupt service task is performed but not the main task.

3.5 Software Interrupt (INTSW)

Executing the SWI instruction generates a software interrupt and immediately starts interrupt processing (INTSW is highest prioritized interrupt).

Use the SWI instruction only for detection of the address error or for debugging.

3.5.1 Address error detection

FFH is read if for some cause such as noise the CPU attempts to fetch an instruction from a non-existent memory address during single chip mode. Code FFH is the SWI instruction, so a software interrupt is generated and an address error is detected. The address error detection range can be further expanded by writing FFH to unused areas of the program memory. Address trap reset is generated in case that an instruction is fetched from RAM, DBR or SFR areas.

3.5.2 Debugging

Debugging efficiency can be increased by placing the SWI instruction at the software break point setting address.

3.6 Undefined Instruction Interrupt (INTUNDEF)

Taking code which is not defined as authorized instruction for instruction causes INTUNDEF. INTUNDEF is generated when the CPU fetches such a code and tries to execute it. INTUNDEF is accepted even if non-maskable interrupt is in process. Contemporary process is broken and INTUNDEF interrupt process starts, soon after it is requested.

Note: The undefined instruction interrupt (INTUNDEF) forces CPU to jump into vector address, as software interrupt (SWI) does.

3.7 Address Trap Interrupt (INTATRAP)

Fetching instruction from unauthorized area for instructions (Address trapped area) causes reset output or address trap interrupt (INTATRAP). INTATRAP is accepted even if non-maskable interrupt is in process. Contemporary process is broken and INTATRAP interrupt process starts, soon after it is requested.

Note: The operating mode under address trapped, whether to be reset output or interrupt processing, is selected on watchdog timer control register (WDTCR).

3.8 External Interrupts

The TMP86P820FG has 5 external interrupt inputs. These inputs are equipped with digital noise reject circuits (Pulse inputs of less than a certain time are eliminated as noise).

Edge selection is also possible with INT1 to INT3. The $\overline{\text{INT0}}$ /P63 pin can be configured as either an external interrupt input pin or an input/output port, and is configured as an input port during reset.

Edge selection, noise reject control and $\overline{\text{INT0}}$ /P63 pin function selection are performed by the external interrupt control register (EINTCR).

Source	Pin	Enable Conditions	Release Edge	Digital Noise Reject
INT0	$\overline{\text{INT0}}$	IMF • EF4 • INT0EN=1	Falling edge	Pulses of less than 2/fc [s] are eliminated as noise. Pulses of 7/fc [s] or more are considered to be signals. In the SLOW or the SLEEP mode, pulses of less than 1/fs [s] are eliminated as noise. Pulses of 3.5/fs [s] or more are considered to be signals.
INT1	INT1	IMF • EF5 = 1	Falling edge or Rising edge	Pulses of less than 15/fc or 63/fc [s] are eliminated as noise. Pulses of 49/fc or 193/fc [s] or more are considered to be signals. In the SLOW or the SLEEP mode, pulses of less than 1/fs [s] are eliminated as noise. Pulses of 3.5/fs [s] or more are considered to be signals.
INT2	INT2	IMF • EF7 = 1	Falling edge or Rising edge	Pulses of less than 7/fc [s] are eliminated as noise. Pulses of 25/fc [s] or more are considered to be signals. In the SLOW or the SLEEP mode, pulses of less than 1/fs [s] are eliminated as noise. Pulses of 3.5/fs [s] or more are considered to be signals.
INT3	INT3	IMF • EF14 = 1 and IL14ER=0	Falling edge or Rising edge	Pulses of less than 7/fc [s] are eliminated as noise. Pulses of 25/fc [s] or more are considered to be signals. In the SLOW or the SLEEP mode, pulses of less than 1/fs [s] are eliminated as noise. Pulses of 3.5/fs [s] or more are considered to be signals.
INT5	$\overline{\text{INT5}}$	IMF • EF15 = 1	Falling edge	Pulses of less than 2/fc [s] are eliminated as noise. Pulses of 7/fc [s] or more are considered to be signals. In the SLOW or the SLEEP mode, pulses of less than 1/fs [s] are eliminated as noise. Pulses of 3.5/fs [s] or more are considered to be signals.

Note 1: In NORMAL1/2 or IDLE1/2 mode, if a signal with no noise is input on an external interrupt pin, it takes a maximum of "signal establishment time + 6/fs[s]" from the input signal's edge to set the interrupt latch.

Note 2: When INT0EN = "0", IL4 is not set even if a falling edge is detected on the $\overline{\text{INT0}}$ pin input.

Note 3: When a pin with more than one function is used as an output and a change occurs in data or input/output status, an interrupt request signal is generated in a pseudo manner. In this case, it is necessary to perform appropriate processing such as disabling the interrupt enable flag.

Not Recommended
for New Design

External Interrupt Control Register

EINTCR	7	6	5	4	3	2	1	0	
(0037H)	INT1NC	INT0EN	-	-	INT3ES	INT2ES	INT1ES		(Initial value: 00** 000*)

INT1NC	Noise reject time select	0: Pulses of less than 63/fc [s] are eliminated as noise 1: Pulses of less than 15/fc [s] are eliminated as noise	R/W
INT0EN	P63/ $\overline{\text{INT0}}$ pin configuration	0: P63 input/output port 1: $\overline{\text{INT0}}$ pin (Port P63 should be set to an input mode)	R/W
INT3 ES	INT3 edge select	0: Rising edge 1: Falling edge	R/W
INT2 ES	INT2 edge select	0: Rising edge 1: Falling edge	R/W
INT1 ES	INT1 edge select	0: Rising edge 1: Falling edge	R/W

- Note 1: fc: High-frequency clock [Hz], *: Don't care
- Note 2: When the system clock frequency is switched between high and low or when the external interrupt control register (EINTCR) is overwritten, the noise canceller may not operate normally. It is recommended that external interrupts are disabled using the interrupt enable register (EIR).
- Note 3: The maximum time from modifying INT1NC until a noise reject time is changed is $2^6/\text{fc}$.

Not Recommended
for New Design

4. Special Function Register (SFR)

The TMP86P820FG adopts the memory mapped I/O system, and all peripheral control and data transfers are performed through the special function register (SFR) or the data buffer register (DBR). The SFR is mapped on address 0000H to 003FH, DBR is mapped on address 0F80H to 0FFFH.

This chapter shows the arrangement of the special function register (SFR) and data buffer register (DBR) for TMP86P820FG.

4.1 SFR

Address	Read	Write
0000H	Reserved	
0001H	P1DR	
0002H	P2DR	
0003H	P3DR	
0004H	P3OUTCR	
0005H	P5DR	
0006H	P6DR	
0007H	P7DR	
0008H	P1PRD	-
0009H	P2PRD	-
000AH	P3PRD	-
000BH	P5PRD	-
000CH	P6CR	
000DH	P7PRD	-
000EH	ADCCR1	
000FH	ADCCR2	
0010H	TREG1AL	
0011H	TREG1AM	
0012H	TREG1AH	
0013H	TREG1B	
0014H	TC1CR1	
0015H	TC1CR2	
0016H	TC1SR	-
0017H	Reserved	
0018H	TC3CR	
0019H	TC4CR	
001AH	Reserved	
001BH	Reserved	
001CH	TTREG3	
001DH	TTREG4	
001EH	Reserved	
001FH	Reserved	
0020H	ADCDR1	-
0021H	ADCDR2	-
0022H	Reserved	
0023H	Reserved	
0024H	Reserved	
0025H	Reserved	

Address	Read	Write
0026H		Reserved
0027H		Reserved
0028H		LDCR
0029H		P1LCR
002AH		P5LCR
002BH		P7LCR
002CH		PWREG3
002DH		PWREG4
002EH		Reserved
002FH		Reserved
0030H		Reserved
0031H		Reserved
0032H		Reserved
0033H		Reserved
0034H	-	WDTCR1
0035H	-	WDTCR2
0036H		TBTCR
0037H		EINTCR
0038H		SYSCR1
0039H		SYSCR2
003AH		EIRL
003BH		EIRH
003CH		ILL
003DH		ILH
003EH		INTSEL
003FH		PSW

Note 1: Do not access reserved areas by the program.

Note 2: - ; Cannot be accessed.

Note 3: Write-only registers and interrupt latches cannot use the read-modify-write instructions (Bit manipulation instructions such as SET, CLR, etc. and logical operation instructions such as AND, OR, etc.).

4.2 DBR

Address	Read	Write
0F80H		SEG1/0
0F81H		SEG3/2
0F82H		SEG5/4
0F83H		SEG7/6
0F84H		SEG9/8
0F85H		SEG11/10
0F86H		SEG13/12
0F87H		SEG15/14
0F88H		SEG17/16
0F89H		SEG19/18
0F8AH		SEG21/20
0F8BH		SEG23/22
0F8CH		SEG25/24
0F8DH		SEG27/26
0F8EH		SEG29/28
0F8FH		SEG31/30
0F90H		SIOBR0
0F91H		SIOBR1
0F92H		SIOBR2
0F93H		SIOBR3
0F94H		SIOBR4
0F95H		SIOBR5
0F96H		SIOBR6
0F97H		SIOBR7
0F98H	-	SIOCR1
0F99H	SIOSR	SIOCR2
0F9AH	-	STOPCR
0F9BH		Reserved
0F9CH		Reserved
0F9DH		Reserved
0F9EH		Reserved
0F9FH		Reserved

Address	Read	Write
0FA0H		Reserved
::		::
0FBFH		Reserved

Address	Read	Write
0FC0H		Reserved
::		::
0FDFH		Reserved

Address	Read	Write
0FE0H		Reserved
::		::
0FFFH		Reserved

Note 1: Do not access reserved areas by the program.

Note 2: – ; Cannot be accessed.

Note 3: Write-only registers and interrupt latches cannot use the read-modify-write instructions (Bit manipulation instructions such as SET, CLR, etc. and logical operation instructions such as AND, OR, etc.).

Not Recommended
for New Design

5. I/O Ports

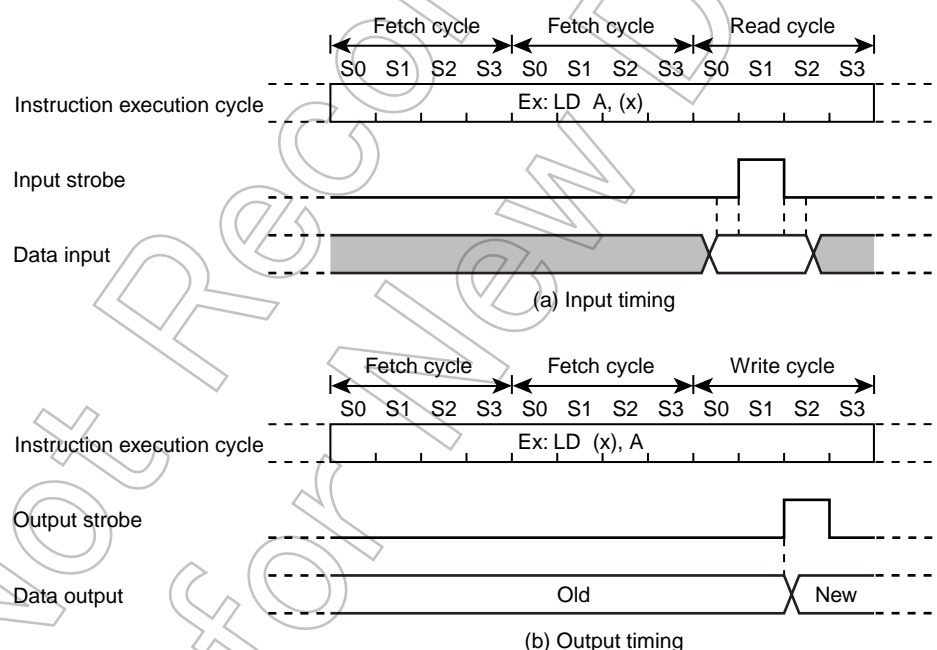
The TMP86P820FG have 6 parallel input/output ports (39 pins) as follows.

	Primary Function	Secondary Functions
Port P1	8-bit I/O port	External interrupt input, serial interface input/output, and segment output.
Port P2	3-bit I/O port	Low-frequency resonator connections, external interrupt input, STOP mode release signal input.
Port P3	4-bit I/O port	Timer/counter input/output and divider output.
Port P5	8-bit I/O port	Segment output.
Port P6	8-bit I/O port	Analog input, external interrupt input, timer/counter input and STOP mode release signal input.
Port P7	8-bit I/O port	Segment output.

Each output port contains a latch, which holds the output data. All input ports do not have latches, so the external input data should be externally held until the input data is read from outside or reading should be performed several times before processing. Figure 5-1 shows input/output timing examples.

External data is read from an I/O port in the S1 state of the read cycle during execution of the read instruction. This timing cannot be recognized from outside, so that transient input such as chattering must be processed by the program.

Output data changes in the S2 state of the write cycle during execution of the instruction which writes to an I/O port.



Note: The positions of the read and write cycles may vary, depending on the instruction.

Figure 5-1 Input/Output Timing (Example)

5.2 Port P2 (P22 to P20)

Port P2 is a 3-bit input/output port.

It is also used as an external interrupt, a STOP mode release signal input, and low-frequency crystal oscillator connection pins. When used as an input port or a secondary function pins, respective output latch (P2DR) should be set to “1”.

During reset, the output latch is initialized to “1”.

A low-frequency crystal oscillator (32.768 kHz) is connected to pins P21 (XTIN) and P22 (XTOUT) in the dual-clock mode. In the single-clock mode, pins P21 and P22 can be used as normal input/output ports.

It is recommended that pin P20 should be used as an external interrupt input, a STOP mode release signal input, or an input port. If it is used as an output port, the interrupt latch is set on the falling edge of the output pulse.

P2 port output latch (P2DR) and P2 port terminal input (P2PRD) are located on their respective address.

When read the output latch data, the P2DR register should be read and when read the terminal input data, the P2PRD register should be read. If a read instruction is executed for port P2, read data of bits 7 to 3 are unstable.

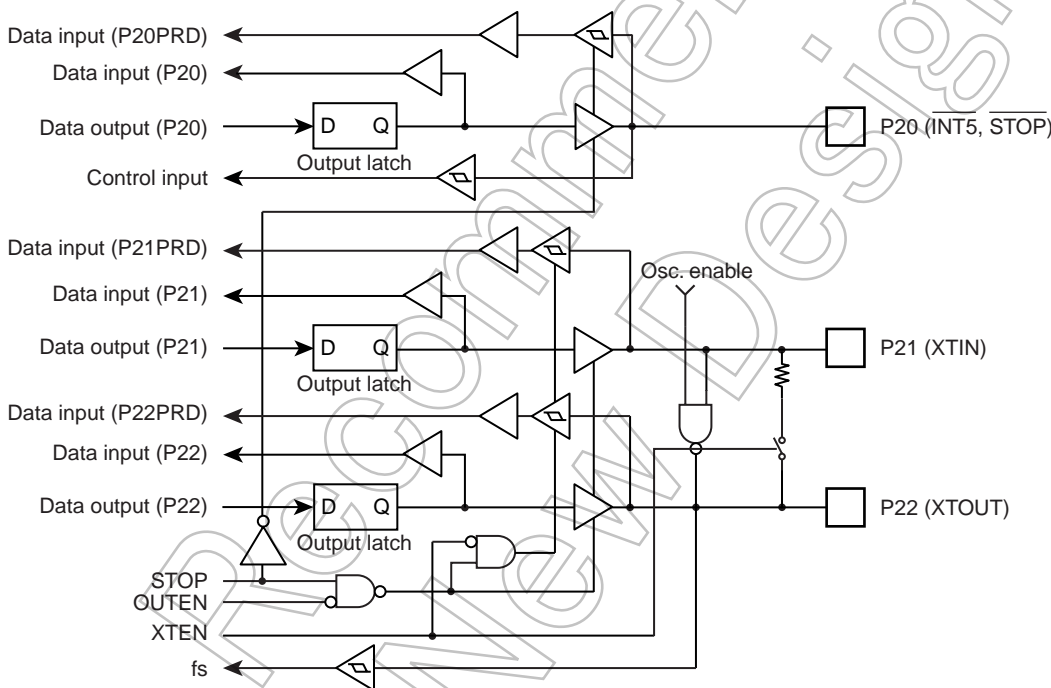


Figure 5-3 Port 2

	7	6	5	4	3	2	1	0	
P2DR (0002H) R/W						P22 XTOUT	P21 XTIN	P20 INT5 STOP	(Initial value: **** *111)
	7	6	5	4	3	2	1	0	
P2PRD (0009H) Read only						P22	P21	P20	

Note: Port P20 is used as $\overline{\text{STOP}}$ pin. Therefore, when stop mode is started, OUTEN does not affect to P20, and P20 becomes High-Z mode.

5.4 Port P5 (P57 to P50)

Port P5 is an 8-bit input/output port which is also used as a segment pins of LCD.

When used as input port, the respective output latch (P5DR) should be set to “1”.

During reset, the output latch is initialized to “1”.

When used as a segment pins of LCD, the respective bit of P5LCR should be set to “1”. When used as an output port, the respective P5LCR bit should be set to “0”.

P5 port output latch (P5DR) and P5 port terminal input (P5PRD) are located on their respective address.

When read the output latch data, the P5DR register should be read and when read the terminal input data, the P5PRD register should be read. If the terminal input data which is configured as LCD segment output is read, unstable data is read.

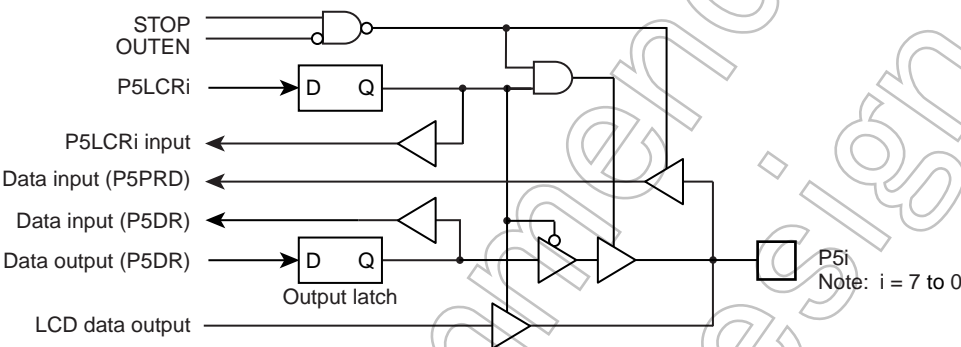


Figure 5-5 Port 5

P5DR (0005H) R/W	7	6	5	4	3	2	1	0	(Initial value: 1111 1111)
	P57 SEG16	P56 SEG17	P55 SEG18	P54 SEG19	P53 SEG20	P52 SEG21	P51 SEG22	P50 SEG23	
P5LCR (002AH)	7	6	5	4	3	2	1	0	(Initial value: 0000 0000)
P5LCR	Port P5/segment output control (set for each bit individually)						0: P5 input/output port 1: LCD segment output		R/W
P5PRD (000BH) Read only	7	6	5	4	3	2	1	0	
	P57	P56	P55	P54	P53	P52	P51	P50	

Read-modify-write instruction writes the all data of 8-bit after data is read and modified. Because a bit setting Input mode read data of terminal, the output latch is changed by these instruction. So P6 port can not input data.

5.6 Port P7 (P77 to P70)

Port P7 is an 8-bit input/output port which is also used as a segment pins of LCD.

When used as input port, the respective output latch (P7DR) should be set to “1”.

During reset, the output latch is initialized to “1”.

When used as a segment pins of LCD, the respective bit of P7LCR should be set to “1” and its corresponding P7LCR bit should be set to “0”. When used as an output port, the respective P7LCR bit should be set to “0”.

P7 port output latch (P7DR) and P7 port terminal input (P7PRD) are located on their respective address.

When read the output latch data, the P7DR register should be read and when read the terminal input data, the P7PRD register should be read. If the terminal input data which is configured as LCD segment output is read, unstable data is read.

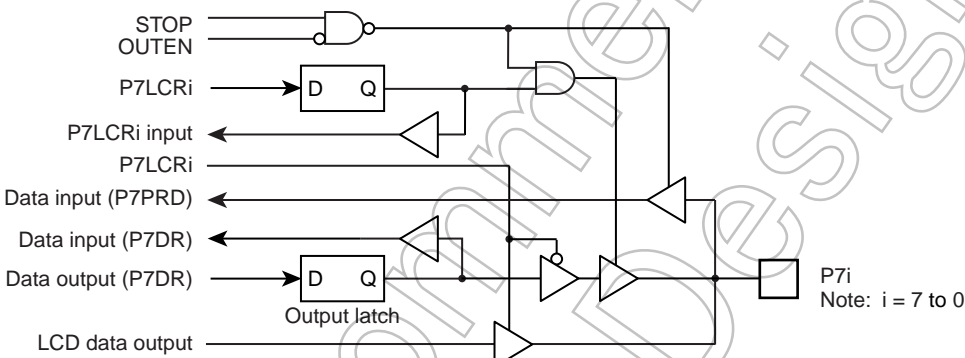


Figure 5-7 Port 7

P7DR (0007H) R/W	7	6	5	4	3	2	1	0	
	P77 SEG8	P76 SEG9	P75 SEG10	P74 SEG11	P73 SEG12	P72 SEG13	P71 SEG14	P70 SEG15	(Initial value: 1111 1111)
P7LCR (002BH)	7	6	5	4	3	2	1	0	
									(Initial value: 0000 0000)
P7LCR	Port P7/segment output control (set for each bit individually)							0: P7 input/output port 1: Segment output	R/W
P7PRD (000DH) Read only	7	6	5	4	3	2	1	0	
	P77	P76	P75	P74	P73	P72	P71	P70	

Not Recommended
for New Design

6. Time Base Timer (TBT)

The time base timer generates time base for key scanning, dynamic displaying, etc. It also provides a time base timer interrupt (INTTBT).

6.1 Time Base Timer

6.1.1 Configuration

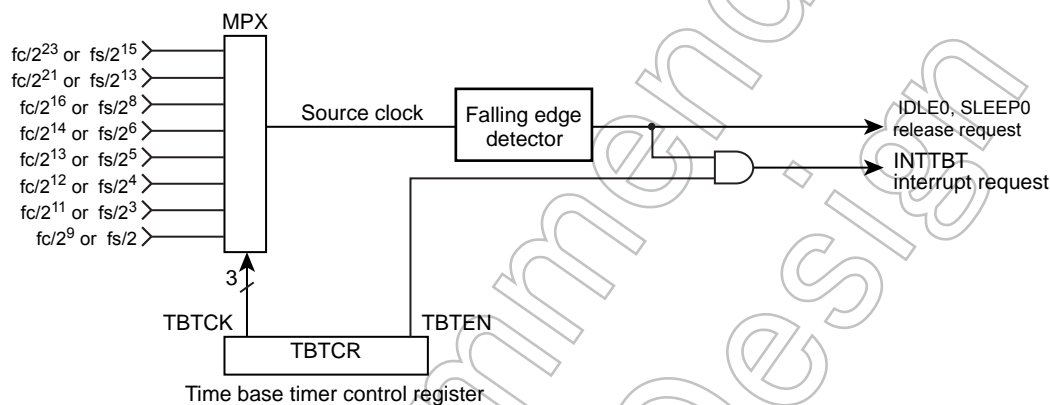


Figure 6-1 Time Base Timer configuration

6.1.2 Control

Time Base Timer is controlled by Time Base Timer control register (TBTCR).

Time Base Timer Control Register

	7	6	5	4	3	2	1	0	
TBTCR (0036H)	(DVOEN)	(DVOCK)	(DV7CK)	TBTEN	TBTC				(Initial Value: 0000 0000)

TBTEN	Time Base Timer enable / disable	0: Disable 1: Enable				
TBTC	Time Base Timer interrupt Frequency select : [Hz]		NORMAL 1/2, IDLE 1/2 Mode		SLOW 1/2 SLEEP 1/2 Mode	R/W
			DV7CK = 0	DV7CK = 1		
		000	$f_c/2^{23}$	$f_s/2^{15}$	$f_s/2^{15}$	
		001	$f_c/2^{21}$	$f_s/2^{13}$	$f_s/2^{13}$	
		010	$f_c/2^{16}$	$f_s/2^8$	—	
		011	$f_c/2^{14}$	$f_s/2^6$	—	
		100	$f_c/2^{13}$	$f_s/2^5$	—	
		101	$f_c/2^{12}$	$f_s/2^4$	—	
		110	$f_c/2^{11}$	$f_s/2^3$	—	
		111	$f_c/2^9$	$f_s/2$	—	

Note 1: fc; High-frequency clock [Hz], fs; Low-frequency clock [Hz], *, Don't care

Note 2: The interrupt frequency (TBTCK) must be selected with the time base timer disabled (TBTEN="0"). (The interrupt frequency must not be changed with the disable from the enable state.) Both frequency selection and enabling can be performed simultaneously.

Example :Set the time base timer frequency to $f_c/2^{16}$ [Hz] and enable an INTTBT interrupt.

```
LD      (TBTCK) , 00000010B      ; TBTCK ← 010
LD      (TBTCK) , 00001010B      ; TBTEN ← 1
DI                               ; IMF ← 0
SET     (EIRL) . 6
```

Table 6-1 Time Base Timer Interrupt Frequency (Example : $f_c = 16.0$ MHz, $f_s = 32.768$ kHz)

TBTCK	Time Base Timer Interrupt Frequency [Hz]		
	NORMAL1/2, IDLE1/2 Mode	NORMAL1/2, IDLE1/2 Mode	SLOW1/2, SLEEP1/2 Mode
	DV7CK = 0	DV7CK = 1	
000	1.91	1	1
001	7.63	4	4
010	244.14	128	—
011	976.56	512	—
100	1953.13	1024	—
101	3906.25	2048	—
110	7812.5	4096	—
111	31250	16384	—

6.1.3 Function

An INTTBT (Time Base Timer Interrupt) is generated on the first falling edge of source clock (The divider output of the timing generator which is selected by TBTCK.) after time base timer has been enabled.

The divider is not cleared by the program; therefore, only the first interrupt may be generated ahead of the set interrupt period (Figure 6-2).

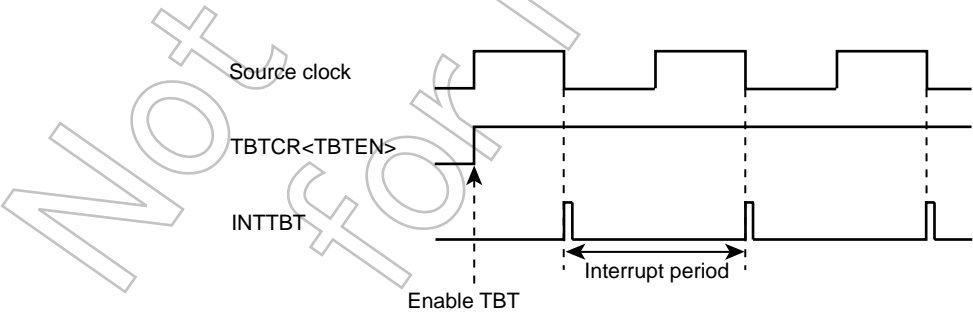


Figure 6-2 Time Base Timer Interrupt

6.2 Divider Output ($\overline{\text{DVO}}$)

Approximately 50% duty pulse can be output using the divider output circuit, which is useful for piezoelectric buzzer drive. Divider output is from $\overline{\text{DVO}}$ pin.

6.2.1 Configuration

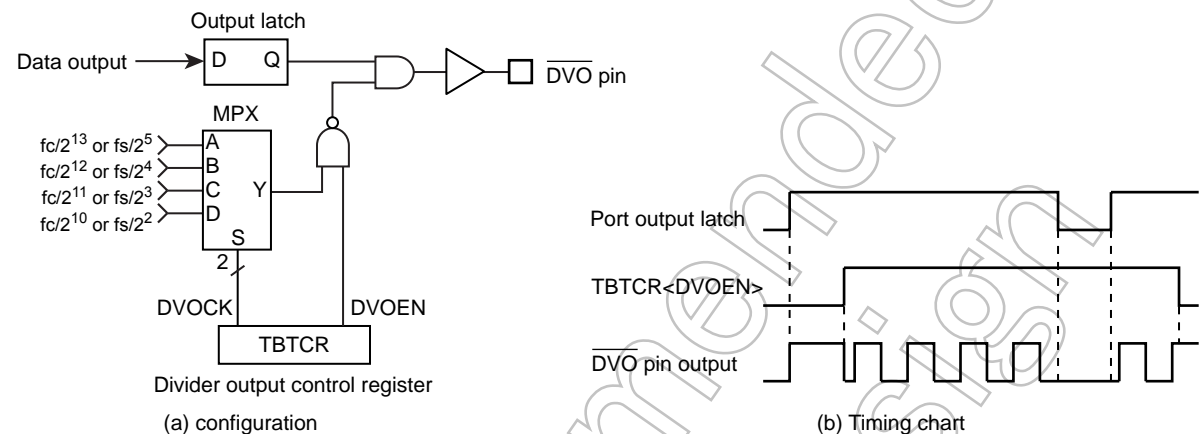


Figure 6-3 Divider Output

6.2.2 Control

The Divider Output is controlled by the Time Base Timer Control Register.

Time Base Timer Control Register

	7	6	5	4	3	2	1	0	
TBTCR (0036H)	DVOEN	DVOCK	(DV7CK)	(TBTEN)	(TBTCK)				(Initial value: 0000 0000)

DVOEN	Divider output enable / disable	0: Disable 1: Enable			R/W
DVOCK	Divider Output ($\overline{\text{DVO}}$) frequency selection: [Hz]	NORMAL 1/2, IDLE 1/2 Mode			R/W
		DV7CK = 0		DV7CK = 1	
		00	$fc/2^{13}$	$fs/2^5$	$fs/2^5$
		01	$fc/2^{12}$	$fs/2^4$	$fs/2^4$
		10	$fc/2^{11}$	$fs/2^3$	$fs/2^3$
		11	$fc/2^{10}$	$fs/2^2$	$fs/2^2$

Note: Selection of divider output frequency (DVOCK) must be made while divider output is disabled (DVOEN="0"). Also, in other words, when changing the state of the divider output frequency from enabled (DVOEN="1") to disable(DVOEN="0"), do not change the setting of the divider output frequency.

Example :1.95 kHz pulse output (fc = 16.0 MHz)

```
LD      (TBTCR) , 00000000B      ; DVOCK ← "00"  
LD      (TBTCR) , 10000000B      ; DVOEN ← "1"
```

Table 6-2 Divider Output Frequency (Example : fc = 16.0 MHz, fs = 32.768 kHz)

DVOCK	Divider Output Frequency [Hz]		
	NORMAL1/2, IDLE1/2 Mode		SLOW1/2, SLEEP1/2 Mode
	DV7CK = 0	DV7CK = 1	
00	1.953 k	1.024 k	1.024 k
01	3.906 k	2.048 k	2.048 k
10	7.813 k	4.096 k	4.096 k
11	15.625 k	8.192 k	8.192 k

7. Watchdog Timer (WDT)

The watchdog timer is a fail-safe system to detect rapidly the CPU malfunctions such as endless loops due to spurious noises or the deadlock conditions, and return the CPU to a system recovery routine.

The watchdog timer signal for detecting malfunctions can be programmed only once as “reset request” or “interrupt request”. Upon the reset release, this signal is initialized to “reset request”.

When the watchdog timer is not used to detect malfunctions, it can be used as the timer to provide a periodic interrupt.

Note: Care must be taken in system design since the watchdog timer functions are not be operated completely due to effect of disturbing noise.

7.1 Watchdog Timer Configuration

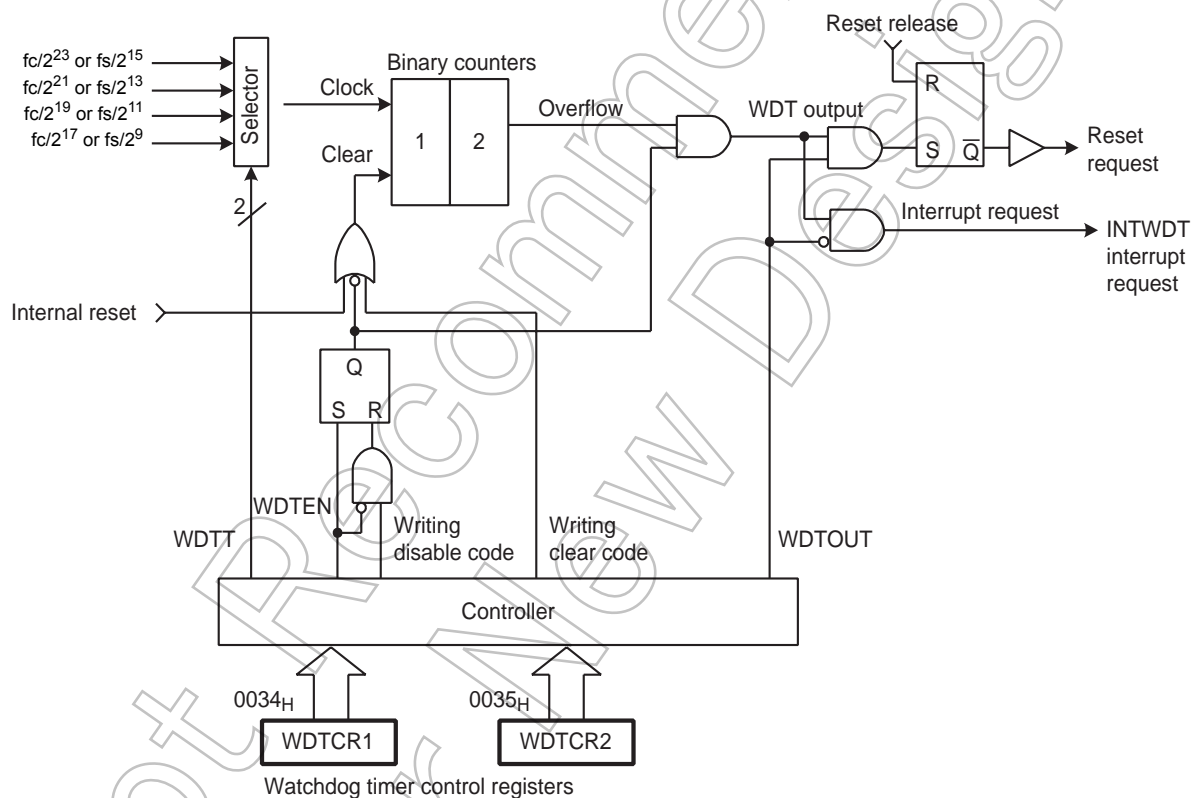


Figure 7-1 Watchdog Timer Configuration

7.2 Watchdog Timer Control

The watchdog timer is controlled by the watchdog timer control registers (WDTCR1 and WDTCR2). The watchdog timer is automatically enabled after the reset release.

7.2.1 Malfunction Detection Methods Using the Watchdog Timer

The CPU malfunction is detected, as shown below.

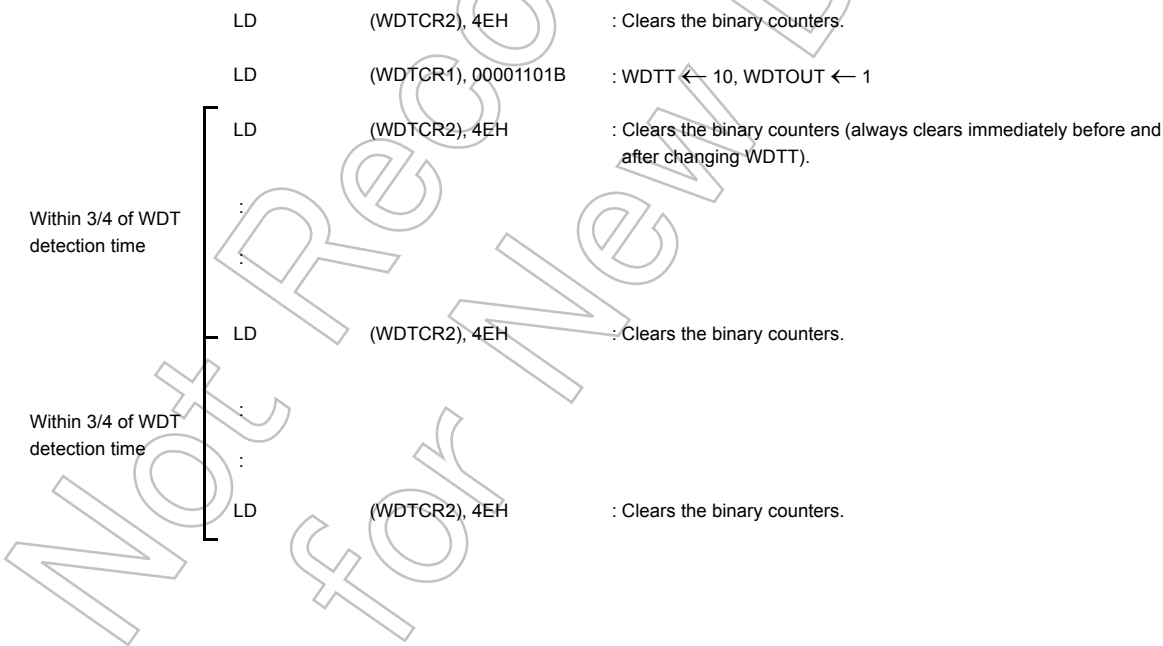
1. Set the detection time, select the output, and clear the binary counter.
2. Clear the binary counter repeatedly within the specified detection time.

If the CPU malfunctions such as endless loops or the deadlock conditions occur for some reason, the watchdog timer output is activated by the binary-counter overflow unless the binary counters are cleared. When WDTCR1<WDTOUT> is set to “1” at this time, the reset request is generated and the RESET pin outputs a low-level signal, then internal hardware is initialized. When WDTCR1<WDTOUT> is set to “0”, a watchdog timer interrupt (INTWDT) is generated.

The watchdog timer temporarily stops counting in the STOP mode including the warm-up or IDLE/SLEEP mode, and automatically restarts (continues counting) when the STOP/IDLE/SLEEP mode is inactivated.

Note: The watchdog timer consists of an internal divider and a two-stage binary counter. When the clear code 4EH is written, only the binary counter is cleared, but not the internal divider. The minimum binary-counter overflow time, that depends on the timing at which the clear code (4EH) is written to the WDTCR2 register, may be 3/4 of the time set in WDTCR1<WDTT>. Therefore, write the clear code using a cycle shorter than 3/4 of the time set to WDTCR1<WDTT>.

Example :Setting the watchdog timer detection time to $2^{21}/f_c$ [s], and resetting the CPU malfunction detection



Watchdog Timer Control Register 1

WDTCR1 (0034H)	7	6	5	4	3	2	1	0	
			(ATAS)	(ATOUT)	WDTEN	WDTT	WDTOUT		(Initial value: **11 1001)

WDTEN	Watchdog timer enable/disable	0: Disable (Writing the disable code to WDTCR2 is required.) 1: Enable				Write only
WDTT	Watchdog timer detection time [s]		NORMAL1/2 mode		SLOW1/2 mode	Write only
			DV7CK = 0	DV7CK = 1		
		00	2 ²⁵ /fc	2 ¹⁷ /fs	2 ¹⁷ /fs	
		01	2 ²³ /fc	2 ¹⁵ /fs	2 ¹⁵ fs	
		10	2 ²¹ fc	2 ¹³ /fs	2 ¹³ fs	
		11	2 ¹⁹ /fc	2 ¹¹ /fs	2 ¹¹ /fs	
WDTOUT	Watchdog timer output select	0: Interrupt request 1: Reset request				Write only

Note 1: After clearing WDTOUT to "0", the program cannot set it to "1".

Note 2: fc: High-frequency clock [Hz], fs: Low-frequency clock [Hz], *: Don't care

Note 3: WDTCR1 is a write-only register and must not be used with any of read-modify-write instructions. If WDTCR1 is read, a don't care is read.

Note 4: To activate the STOP mode, disable the watchdog timer or clear the counter immediately before entering the STOP mode. After clearing the counter, clear the counter again immediately after the STOP mode is inactivated.

Note 5: To clear WDTEN, set the register in accordance with the procedures shown in "7.2.3 Watchdog Timer Disable".

Watchdog Timer Control Register 2

WDTCR2 (0035H)	7	6	5	4	3	2	1	0	
									(Initial value: **** *)

WDTCR2	Write Watchdog timer control code	4EH: Clear the watchdog timer binary counter (Clear code) B1H: Disable the watchdog timer (Disable code) D2H: Enable assigning address trap area Others: Invalid	Write only
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Note 1: The disable code is valid only when WDTCR1<WDTEN> = 0.

Note 2: *: Don't care

Note 3: The binary counter of the watchdog timer must not be cleared by the interrupt task.

Note 4: Write the clear code 4EH using a cycle shorter than 3/4 of the time set in WDTCR1<WDTT>.

7.2.2 Watchdog Timer Enable

Setting WDTCR1<WDTEN> to "1" enables the watchdog timer. Since WDTCR1<WDTEN> is initialized to "1" during reset, the watchdog timer is enabled automatically after the reset release.

7.2.3 Watchdog Timer Disable

To disable the watchdog timer, set the register in accordance with the following procedures. Setting the register in other procedures causes a malfunction of the microcontroller.

- 1. Set the interrupt master flag (IMF) to “0”.
- 2. Set WDTCR2 to the clear code (4EH).
- 3. Set WDTCR1<WDTEN> to “0”.
- 4. Set WDTCR2 to the disable code (B1H).

Note: While the watchdog timer is disabled, the binary counters of the watchdog timer are cleared.

Example :Disabling the watchdog timer

DI

: IMF ← 0

LD

(WDTCR2), 04EH

: Clears the binary counter

LDW

(WDTCR1), 0B101H

: WDTEN ← 0, WDTCR2 ← Disable code

Table 7-1 Watchdog Timer Detection Time (Example: fc = 16.0 MHz, fs = 32.768 kHz)

WDTT	Watchdog Timer Detection Time[s]		
	NORMAL1/2 mode		SLOW mode
	DV7CK = 0	DV7CK = 1	
00	2.097	4	4
01	524.288 m	1	1
10	131.072 m	250 m	250 m
11	32.768 m	62.5 m	62.5 m

7.2.4 Watchdog Timer Interrupt (INTWDT)

When WDTCR1<WDTOUT> is cleared to “0”, a watchdog timer interrupt request (INTWDT) is generated by the binary-counter overflow.

A watchdog timer interrupt is the non-maskable interrupt which can be accepted regardless of the interrupt master flag (IMF).

When a watchdog timer interrupt is generated while the other interrupt including a watchdog timer interrupt is already accepted, the new watchdog timer interrupt is processed immediately and the previous interrupt is held pending. Therefore, if watchdog timer interrupts are generated continuously without execution of the RETN instruction, too many levels of nesting may cause a malfunction of the microcontroller.

To generate a watchdog timer interrupt, set the stack pointer before setting WDTCR1<WDTOUT>.

Example :Setting watchdog timer interrupt

LD

SP, 013FH

: Sets the stack pointer

LD

(WDTCR1), 00001000B

: WDTOUT ← 0

7.2.5 Watchdog Timer Reset

When a binary-counter overflow occurs while $WDTCR1<WDTOUT>$ is set to "1", a watchdog timer reset request is generated. When a watchdog timer reset request is generated, the \overline{RESET} pin outputs a low-level signal and the internal hardware is reset. The reset time is maximum $24/f_c$ [s] ($1.5\ \mu\text{s}$ @ $f_c = 16.0\ \text{MHz}$).

Note: When a watchdog timer reset is generated in the SLOW1 mode, the reset time is maximum $24/f_c$ (high-frequency clock) since the high-frequency clock oscillator is restarted. However, when crystals have inaccuracies upon start of the high-frequency clock oscillator, the reset time should be considered as an approximate value because it has slight errors.

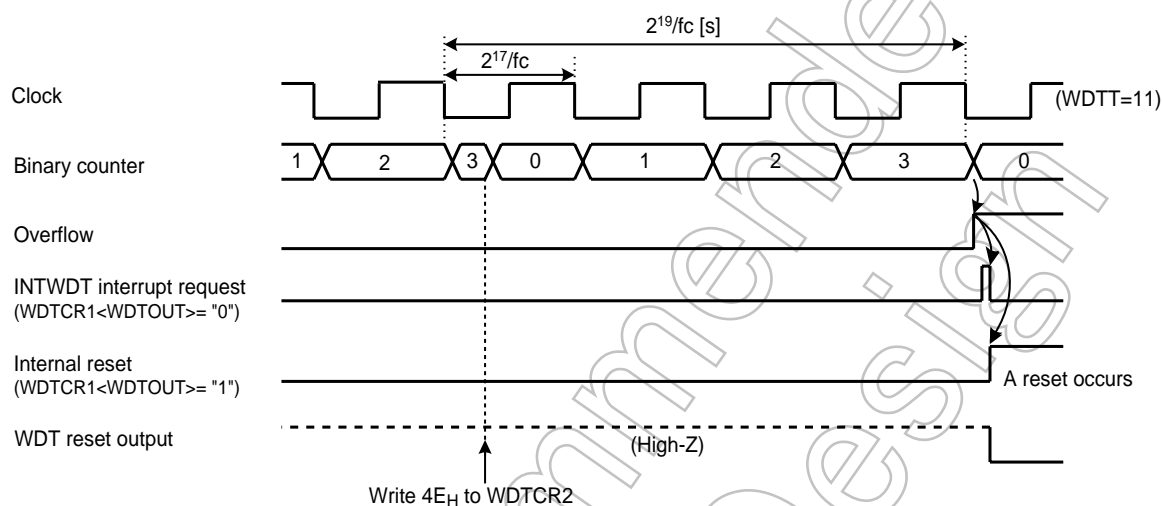


Figure 7-2 Watchdog Timer Interrupt/Reset

7.3 Address Trap

The Watchdog Timer Control Register 1 and 2 share the addresses with the control registers to generate address traps.

Watchdog Timer Control Register 1

WDTCR1 (0034H)	7	6	5	4	3	2	1	0	
			ATAS	ATOUT	(WDTEN)	(WDTT)	(WDTOUT)		(Initial value: **11 1001)

ATAS	Select address trap generation in the internal RAM area	0: Generate no address trap 1: Generate address traps (After setting ATAS to "1", writing the control code D2H to WDTCR2 is required)	Write only
ATOUT	Select operation at address trap	0: Interrupt request 1: Reset request	

Watchdog Timer Control Register 2

WDTCR2 (0035H)	7	6	5	4	3	2	1	0	
									(Initial value: **** *)

WDTCR2	Write Watchdog timer control code and address trap area control code	D2H: Enable address trap area selection (ATRAP control code) 4EH: Clear the watchdog timer binary counter (WDT clear code) B1H: Disable the watchdog timer (WDT disable code) Others: Invalid	Write only
--------	--	--	------------

7.3.1 Selection of Address Trap in Internal RAM (ATAS)

WDTCR1<ATAS> specifies whether or not to generate address traps in the internal RAM area. To execute an instruction in the internal RAM area, clear WDTCR1<ATAS> to "0". To enable the WDTCR1<ATAS> setting, set WDTCR1<ATAS> and then write D2H to WDTCR2.

Executing an instruction in the SFR or DBR area generates an address trap unconditionally regardless of the setting in WDTCR1<ATAS>.

7.3.2 Selection of Operation at Address Trap (ATOUT)

When an address trap is generated, either the interrupt request or the reset request can be selected by WDTCR1<ATOUT>.

7.3.3 Address Trap Interrupt (INTATRAP)

While WDTCR1<ATOUT> is "0", if the CPU should start looping for some cause such as noise and an attempt be made to fetch an instruction from the on-chip RAM (while WDTCR1<ATAS> is "1"), DBR or the SFR area, address trap interrupt (INTATRAP) will be generated.

An address trap interrupt is a non-maskable interrupt which can be accepted regardless of the interrupt master flag (IMF).

When an address trap interrupt is generated while the other interrupt including an address trap interrupt is already accepted, the new address trap is processed immediately and the previous interrupt is held pending. Therefore, if address trap interrupts are generated continuously without execution of the RETN instruction, too many levels of nesting may cause a malfunction of the microcontroller.

To generate address trap interrupts, set the stack pointer beforehand.

7.3.4 Address Trap Reset

While WDTCR1<ATOUT> is “1”, if the CPU should start looping for some cause such as noise and an attempt be made to fetch an instruction from the on-chip RAM (while WDTCR1<ATAS> is “1”), DBR or the SFR area, address trap reset will be generated.

When an address trap reset request is generated, the $\overline{\text{RESET}}$ pin outputs a low-level signal and the internal hardware is reset. The reset time is maximum $24/f_c$ [s] ($1.5 \mu\text{s}$ @ $f_c = 16.0 \text{ MHz}$).

Note: When an address trap reset is generated in the SLOW1 mode, the reset time is maximum $24/f_c$ (high-frequency clock) since the high-frequency clock oscillator is restarted. However, when crystals have inaccuracies upon start of the high-frequency clock oscillator, the reset time should be considered as an approximate value because it has slight errors.

Not Recommended
for New Design

Not Recommended
for New Design

Timer register

	7	6	5	4	3	2	1	0	
TREG1AH (0012H) R/W	—	—	—	—	—	—	TREG1AH		(Initial value: **** **00)

	7	6	5	4	3	2	1	0	
TREG1AM (0011H) R/W	TREG1AM								(Initial value: 0000 0000)

	7	6	5	4	3	2	1	0	
TREG1AL (0010H) R/W	TREG1AL								(Initial value: 0000 0000)

	7	6	5	4	3	2	1	0	
TREG1B (0013H)	Ta				Tb				(Initial value: 0000 0000)

		WGPSCK	NORMAL1/2, IDLE1/2 modes		SLOW1/2, SLEEP1/2 modes	R/W
			DV7CK=0	DV7CK=1		
Ta	Setting "H" level period of the window gate pulse	00	$(16 - Ta) \times 2^{12}/f_c$	$(16 - Ta) \times 2^4/f_s$	$(16 - Ta) \times 2^4/f_s$	
		01	$(16 - Ta) \times 2^{13}/f_c$	$(16 - Ta) \times 2^5/f_s$	$(16 - Ta) \times 2^5/f_s$	
		10	$(16 - Ta) \times 2^{14}/f_c$	$(16 - Ta) \times 2^6/f_s$	$(16 - Ta) \times 2^6/f_s$	
Tb	Setting "L" level period of the window gate pulse	00	$(16 - Tb) \times 2^{12}/f_c$	$(16 - Tb) \times 2^4/f_s$	$(16 - Tb) \times 2^4/f_s$	
		01	$(16 - Tb) \times 2^{13}/f_c$	$(16 - Tb) \times 2^5/f_s$	$(16 - Tb) \times 2^5/f_s$	
		10	$(16 - Tb) \times 2^{14}/f_c$	$(16 - Tb) \times 2^6/f_s$	$(16 - Tb) \times 2^6/f_s$	

Timer/counter 1 control register 1

	7	6	5	4	3	2	1	0	
TC1CR1 (0014H)	TC1C	TC1S	TC1CK			TC1M			(Initial value: 1000 1000)

TC1C	Counter/overflow flag control	0:	Clear Counter/overflow flag ("1" is automatically set after clearing.)			R/W
		1:	Not clear Counter/overflow flag			
TC1S	TC1 start control	00:	Stop and counter clear and overflow flag clear			R/W
		10:	Start			
		*1:	Reserved			
TC1CK	TC1 source clock select		NORMAL 1/2, IDLE 1/2 modes		SLOW 1/2 mode	SLEEP 1/2 mode
			DV7CK="0"			
			DV7CK="1"			
		000:	fc	fc	fc	fc
		001:	fs	fs	-	-
		010:	fc/2 ²³	fs/2 ¹⁵	fs/2 ¹⁵	fs/2 ¹⁵
		011:	fc/2 ¹³	fs/2 ⁵	fs/2 ⁵	fs/2 ⁵
		100:	fc/2 ¹¹	fs/2 ³	fs/2 ³	fs/2 ³
		101:	fc/2 ⁷	fc/2 ⁷	-	-
		110:	fc/2 ³	fc/2 ³	-	-
		111:	External clock (ECIN pin input)			
TC1M	TC1 mode select	00:	Timer/Event counter mode			R/W
		01:	Reserved			
		10:	Pulse width measurement mode			
		11:	Frequency measurement mode			

Note 1: fc; High-frequency clock [Hz] fs; Low-frequency clock [Hz] *; Don't care

Note 2: Writing to the low-byte of the timer register 1A (TREG1AL, TREG1AM), the compare function is inhibited until the high-byte (TREG1AH) is written.

Note 3: Set the mode and source clock, and edge (selection) when the TC1 stops (TC1CR1<TC1S>=00).

Note 4: "fc" can be selected as the source clock only in the timer mode during SLOW mode and in the pulse width measurement mode during NORMAL 1/2 or IDLE 1/2 mode.

Note 5: When a read instruction is executed to the timer register (TREG1A), the counter immediate value, not the register set value, is read out. Therefore it is impossible to read out the written value of TREG1A. To read the counter value, the read instruction should be executed when the counter stops to avoid reading unstable value.

Note 6: Set the timer register (TREG1A) to ≥1.

Note 7: When using the timer mode and pulse width measurement mode, set TC1CR1<TC1CK> (TC1 source clock select) to internal clock.

Note 8: When using the event counter mode, set TC1CR1<TC1CK> (TC1 source clock select) to external clock.

Note 9: Because the read value is different from the written value, do not use read-modify-write instructions to TREG1A.

Note 10: fc/2⁷, fc/2³ can not be used as source clock in SLOW/SLEEP mode.

Note 11: The read data of bits 7 to 2 in TREG1AH are always "0". (Data "1" can not be written.)

Timer/Counter 1 control register 2

	7	6	5	4	3	2	1	0	
TC1CR2 (0015H)	"0"	SGP	SGEDG	WGPSCK	"1"	"0"	(Initial value: *000 00**)		

SGP	Window gate pulse select	00: ECNT input 01: Internal window gate pulse (TREG1B) 10: Reserved 11: Reserved				R/W	
SGEDG	Window gate pulse interrupt edge select	0: Interrupts at the falling edge 1: Interrupts at the falling/rising edges					
WGPSCK	Window gate pulse source clock select		NORMAL1/2,IDLE1/2 modes		SLOW1/2 mode	SLEEP1/2 mode	R/W
			DV7CK="0" DV7CK="1"				
		00:	2 ¹² /fc	2 ⁴ /fs	2 ⁴ /fs	2 ⁴ /fs	
		01:	2 ¹³ /fc	2 ⁵ /fs	2 ⁵ /fs	2 ⁵ /fs	
		10:	2 ¹⁴ /fc	2 ⁶ /fs	2 ⁶ /fs	2 ⁶ /fs	
		11:	Reserved	Reserved	Reserved	Reserved	

Note 1: fc; High-frequency clock [Hz] fs; Low-frequency clock [Hz] *; Don't care
Note 2: Set the mode, source clock, and edge (selection) when the TC1 stops (TC1CR1<TC1S> = 00).
Note 3: Make sure to write TC1CR2 register bit 0,7 to "0" and also TC1CR2 register bit 1 to "1".

TC1 status register

	7	6	5	4	3	2	1	0	
TC1SR (0016H)	HECF	HEOVF	"0"	"0"	"0"	"0"	"0"	"0"	(Initial value: 0000 0000)

HECF	Operating Status monitor	0: Stop (during Tb) or disable 1: Under counting (during Ta)	Read only
HEOVF	Counter overflow monitor	0: No overflow 1: Overflow status	

8.3 Function

TC1 has four operating modes. The timer mode of the TC1 is used at warm-up when switching from SLOW mode to NORMAL2 mode.

8.3.1 Timer mode

In this mode, counting up is performed using the internal clock. The contents of TREG1A are compared with the contents of up-counter. If a match is found, an INTTC1 interrupt is generated, and the counter is cleared. Counting up resumes after the counter is cleared.

Table 8-1 Source clock (internal clock) of Timer/Counter 1

Source Clock				Resolution		Maximum Time Setting	
NORMAL1/2, IDLE1/2 Mode		SLOW Mode	SLEEP Mode	fc = 16 MHz	fs = 32.768 kHz	fc = 16 MHz	fs = 32.768 kHz
DV7CK = 0	DV7CK = 1						
fc/2 ²³ [Hz]	fs/2 ¹⁵ [Hz]	fs/2 ¹⁵ [Hz]	fs/2 ¹⁵ [Hz]	0.52 s	1 s	38.2 h	72.8 h
fc/2 ¹³	fs/2 ⁵	fs/2 ⁵	fs/2 ⁵	512 ms	0.98 ms	2.2 min	4.3 min
fc/2 ¹¹	fs/2 ³	fs/2 ³	fs/2 ³	128 ms	244 ms	0.6 min	1.07 min
fc/2 ⁷	fc/2 ⁷	-	-	8 ms	-	2.1 s	-
fc/2 ³	fc/2 ³	-	-	0.5 ms	-	131.1 ms	-
fc	fc	fc (Note)	-	62.5 ns	-	16.4 ms	-
fs	fs	-	-	-	30.5 ms	-	8 s

Note: When fc is selected for the source clock in SLOW mode, the lower bits 11 of TREG1A is invalid, and a match of the upper bits 7 makes interrupts.

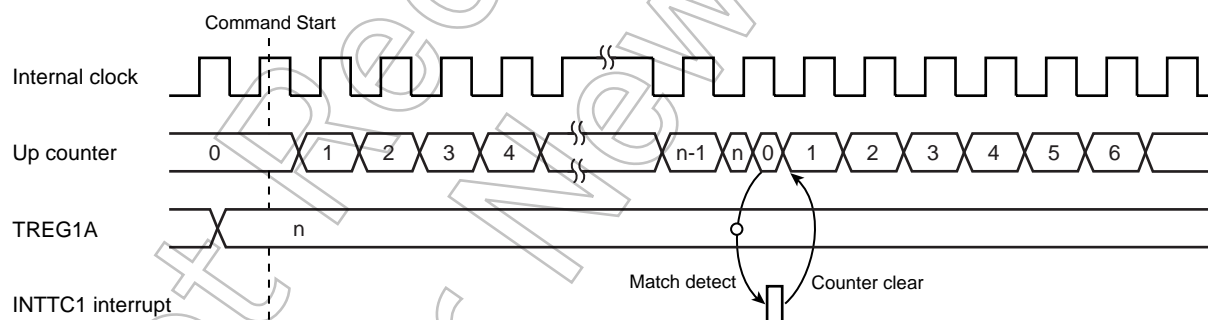


Figure 8-2 Timing chart for timer mode

8.3.2 Event Counter mode

It is a mode to count up at the falling edge of the ECIN pin input. When using this mode, set TC1CR1<TC1CK> to the external clock.

The contents of TREG1A are compared with the contents of up-counter. If a match is found, an INTTC1 interrupt is generated, and the counter is cleared. Counting up resumes for ECIN pin input edge each after the counter is cleared.

The maximum applied frequency is $fc/2^4$ [Hz] in NORMAL 1/2 or IDLE 1/2 mode and $fs/2^4$ [Hz] in SLOW or SLEEP mode. Two or more machine cycles are required for both the "H" and "L" levels of the pulse width.



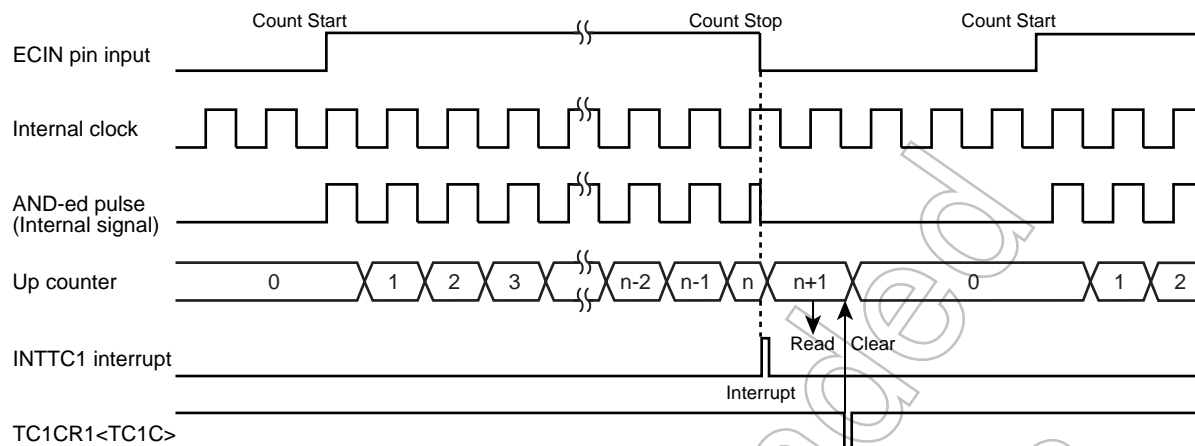


Figure 8-4 Pulse width measurement mode timing chart

8.3.4 Frequency Measurement mode

In this mode, the frequency of ECIN pin input pulse is measured. When using this mode, set TC1CR1<TC1CK> to the external clock.

The edge of the ECIN input pulse is counted during “H” level of the window gate pulse selected by TC1CR2<SGP>. To use ECNT input as a window gate pulse, TC1CR2<SGP> should be set to “00”.

An INTTC1 interrupt is generated on the falling edge or both the rising/falling edges of the window gate pulse, that can be selected by TC1CR2<SGEDG>. In the interrupt service program, read the contents of TREG1A while the count is stopped (window gate pulse is low), then clear the counter using TC1CR1<TC1C>. When the counter is not cleared, counting up resumes from previous stopping value.

The window pulse status can be monitored by TC1SR<HECF>.

When up counter is counted up from 3FFFFH to 00000H, an overflow occurs. At that time, TC1SR<HEOVF> is set to “1”. TC1SR<HEOVF> remains the previous data until the counter is required to be cleared by TC1CR1<TC1C>.

When the internal window gate pulse is selected, the window gate pulse is set as follows.

Table 8-2 Internal window gate pulse setting time

		WGPSCK	NORMAL1/2, IDLE1/2 modes		SLOW1/2, SLEEP1/2 modes	
			DV7CK=0	DV7CK=1		
Ta	Setting “H” level period of the window gate pulse	00	$(16 - Ta) \times 2^{12}/f_c$	$(16 - Ta) \times 2^4/f_s$	$(16 - Ta) \times 2^4/f_s$	R/W
		01	$(16 - Ta) \times 2^{13}/f_c$	$(16 - Ta) \times 2^5/f_s$	$(16 - Ta) \times 2^5/f_s$	
		10	$(16 - Ta) \times 2^{14}/f_c$	$(16 - Ta) \times 2^6/f_s$	$(16 - Ta) \times 2^6/f_s$	
Tb	Setting “L” level period of the window gate pulse	00	$(16 - Tb) \times 2^{12}/f_c$	$(16 - Tb) \times 2^4/f_s$	$(16 - Tb) \times 2^4/f_s$	
		01	$(16 - Tb) \times 2^{13}/f_c$	$(16 - Tb) \times 2^5/f_s$	$(16 - Tb) \times 2^5/f_s$	
		10	$(16 - Tb) \times 2^{14}/f_c$	$(16 - Tb) \times 2^6/f_s$	$(16 - Tb) \times 2^6/f_s$	

The internal window gate pulse consists of “H” level period (Ta) that is counting time and “L” level period (Tb) that is counting stop time. Ta or Tb can be individually set by TREG1B. One cycle contains Ta + Tb.

Note 1: Because the internal window gate pulse is generated in synchronization with the internal divider, it may be delayed for a maximum of one cycle of the source clock (TC1CR2<WGPSCK>) immediately after start of the timer.

Note 2: Set the internal window gate pulse when the timer counter is not operating or during the Tb period. When Tb is overwritten during the Tb period, the update is valid from the next Tb period.

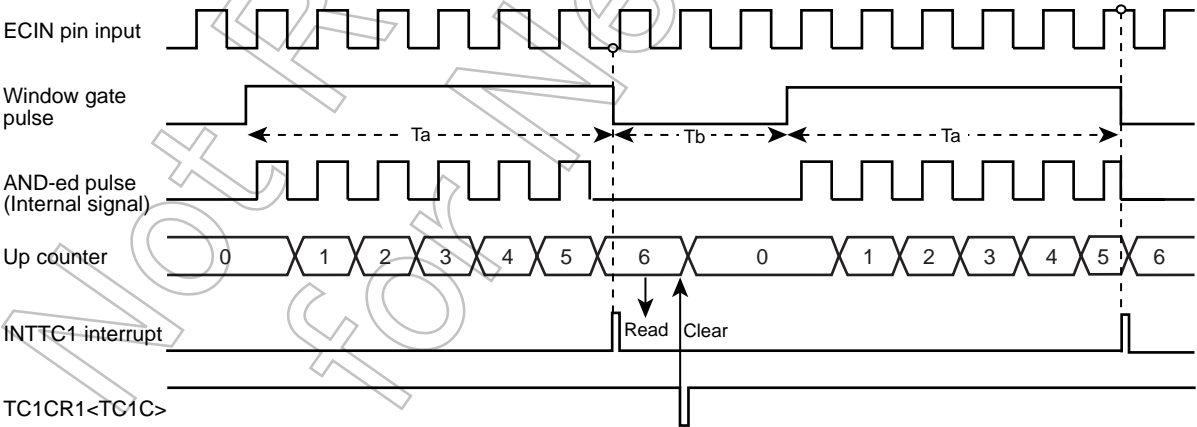
Note 3: Because the up counter is counted on the falling edge of logical AND-ed pulse (between ECIN pin input and window gate pulse), if window gate pulse becomes falling edge while ECIN input is "H" level, the up counter stops plus "1". Therefore, if ECIN input is always "H" level, count value becomes "1".

Table 8-3 Table Setting Ta and Tb (WGPSCK = 10, fc = 16 MHz)

Setting Value	Setting time	Setting Value	Setting time
0	16.38ms	8	8.19ms
1	15.36ms	9	7.17ms
2	14.34ms	A	6.14ms
3	13.31ms	B	5.12ms
4	12.29ms	C	4.10ms
5	11.26ms	D	3.07ms
6	10.24ms	E	2.05ms
7	9.22ms	F	1.02ms

Table 8-4 Table Setting Ta and Tb (WGPSCK = 10, fs = 32.768 kHz)

Setting Value	Setting time	Setting Value	Setting time
0	31.25ms	8	15.63ms
1	29.30ms	9	13.67ms
2	27.34ms	A	11.72ms
3	25.39ms	B	9.77ms
4	23.44ms	C	7.81ms
5	21.48ms	D	5.86ms
6	19.53ms	E	3.91ms
7	17.58ms	F	1.95ms



9.2 TimerCounter Control

The TimerCounter 3 is controlled by the TimerCounter 3 control register (TC3CR) and two 8-bit timer registers (TTREG3, PWREG3).

TimerCounter 3 Timer Register

TTREG3 (001CH) R/W	7	6	5	4	3	2	1	0	(Initial value: 1111 1111)

PWREG3 (002CH) R/W	7	6	5	4	3	2	1	0	(Initial value: 1111 1111)

Note 1: Do not change the timer register (TTREG3) setting while the timer is running.

Note 2: Do not change the timer register (PWREG3) setting in the operating mode except the 8-bit and 16-bit PWM modes while the timer is running.

TimerCounter 3 Control Register

TC3CR (0018H)	7	6	5	4	3	2	1	0	(Initial value: 0000 0000)
	TFF3		TC3CK		TC3S		TC3M		

TFF3	Time F/F3 control	0: Clear 1: Set	R/W
TC3CK	Operating clock selection [Hz]	NORMAL1/2, IDLE1/2 mode	
		DV7CK = 0 DV7CK = 1	
		000	fs/2 ¹¹
		001	fc/2 ⁷
		010	fc/2 ⁵
		011	fc/2 ³
		100	fs
		101	fc/2
		110	fc
		111	TC3 pin input
TC3S	TC3 start control	0: Operation stop and counter clear 1: Operation start	R/W
TC3M	TC3M operating mode select	000: 8-bit timer/event counter mode 001: 8-bit programmable divider output (PDO) mode 010: 8-bit pulse width modulation (PWM) output mode 011: 16-bit mode (Each mode is selectable with TC4M.) 1**: Reserved	R/W

Note 1: fc: High-frequency clock [Hz] fs: Low-frequency clock[Hz]

Note 2: Do not change the TC3M, TC3CK and TFF3 settings while the timer is running.

Note 3: To stop the timer operation (TC3S= 1 → 0), do not change the TC3M, TC3CK and TFF3 settings. To start the timer operation (TC3S= 0 → 1), TC3M, TC3CK and TFF3 can be programmed.

Note 4: To use the TimerCounter in the 16-bit mode, set the operating mode by programming TC4CR<TC4M>, where TC3M must be fixed to 011.

Note 5: To use the TimerCounter in the 16-bit mode, select the source clock by programming TC3CK. Set the timer start control and timer F/F control by programming TC4CR<TC4S> and TC4CR<TFF4>, respectively.

Note 6: The operating clock settings are limited depending on the timer operating mode. For the detailed descriptions, see Table 9-1 and Table 9-2.

Note 7: The timer register settings are limited depending on the timer operating mode. For the detailed descriptions, see Table 9-3.

Note 8: The operating clock f_c in the SLOW or SLEEP mode can be used only as the high-frequency warm-up mode.

Not Recommended
for New Design

The TimerCounter 4 is controlled by the TimerCounter 4 control register (TC4CR) and two 8-bit timer registers (TTREG4 and PWREG4).

TimerCounter 4 Timer Register

TTREG4 (001DH) R/W	7	6	5	4	3	2	1	0	(Initial value: 1111 1111)

PWREG4 (002DH) R/W	7	6	5	4	3	2	1	0	(Initial value: 1111 1111)

Note 1: Do not change the timer register (TTREG4) setting while the timer is running.

Note 2: Do not change the timer register (PWREG4) setting in the operating mode except the 8-bit and 16-bit PWM modes while the timer is running.

TimerCounter 4 Control Register

TC4CR (0019H)	7	6	5	4	3	2	1	0	(Initial value: 0000 0000)
	TFF4		TC4CK		TC4S		TC4M		

TFF4	Timer F/F4 control	0: Clear 1: Set	R/W
TC4CK	Operating clock selection [Hz]	NORMAL 1/2, IDLE 1/2 mode	
		DV7CK = 0	DV7CK = 1
		000	fs/2 ¹¹
		001	fc/2 ⁷
		010	fc/2 ⁵
		011	fc/2 ³
		100	fs
		101	fc/2
TC4S	TC4 start control	0: Operation stop and counter clear 1: Operation start	R/W
TC4M	TC4M operating mode select	000: 8-bit timer/event counter mode	R/W
		001: 8-bit programmable divider output (PDO) mode	
		010: 8-bit pulse width modulation (PWM) output mode	
		011: Reserved	
		100: 16-bit timer/event counter mode	
		101: Warm-up counter mode	
		110: 16-bit pulse width modulation (PWM) output mode	
		111: 16-bit PPG mode	

Note 1: fc: High-frequency clock [Hz] fs: Low-frequency clock [Hz]

Note 2: Do not change the TC4M, TC4CK and TFF4 settings while the timer is running.

Note 3: To stop the timer operation (TC4S= 1 → 0), do not change the TC4M, TC4CK and TFF4 settings.
To start the timer operation (TC4S= 0 → 1), TC4M, TC4CK and TFF4 can be programmed.

Note 4: When TC4M= 1** (upper byte in the 16-bit mode), the source clock becomes the TC3 overflow signal regardless of the TC4CK setting.

Note 5: To use the TimerCounter in the 16-bit mode, select the operating mode by programming TC4M, where TC3CR<TC3M> must be set to 011.

Note 6: To the TimerCounter in the 16-bit mode, select the source clock by programming TC3CR<TC3CK>. Set the timer start control and timer F/F control by programming TC4S and TFF4, respectively.

Note 7: The operating clock settings are limited depending on the timer operating mode. For the detailed descriptions, see Table 9-1 and Table 9-2.

Note 8: The timer register settings are limited depending on the timer operating mode. For the detailed descriptions, see Table 9-3.

Table 9-1 Operating Mode and Selectable Source Clock (NORMAL 1/2 and IDLE 1/2 Modes)

Operating mode	fc/2 ¹¹ or fs/2 ³	fc/2 ⁷	fc/2 ⁵	fc/2 ³	fs	fc/2	fc	TC3 pin input	TC4 pin input
8-bit timer	○	○	○	○	—	—	—	—	—
8-bit event counter	—	—	—	—	—	—	—	○	○
8-bit PDO	○	○	○	○	—	—	—	—	—
8-bit PWM	○	○	○	○	○	○	○	—	—
16-bit timer	○	○	○	○	—	—	—	—	—
16-bit event counter	—	—	—	—	—	—	—	○	—
Warm-up counter	—	—	—	—	○	—	—	—	—
16-bit PWM	○	○	○	○	○	○	○	○	—
16-bit PPG	○	○	○	○	—	—	—	○	—

Note 1: For 16-bit operations (16-bit timer/event counter, warm-up counter, 16-bit PWM and 16-bit PPG), set its source clock on lower bit (TC3CK).

Note 2: ○ : Available source clock

Table 9-2 Operating Mode and Selectable Source Clock (SLOW 1/2 and SLEEP 1/2 Modes)

Operating mode	fc/2 ¹¹ or fs/2 ³	fc/2 ⁷	fc/2 ⁵	fc/2 ³	fs	fc/2	fc	TC3 pin input	TC4 pin input
8-bit timer	○	—	—	—	—	—	—	—	—
8-bit event counter	—	—	—	—	—	—	—	○	○
8-bit PDO	○	—	—	—	—	—	—	—	—
8-bit PWM	○	—	—	—	○	—	—	—	—
16-bit timer	○	—	—	—	—	—	—	—	—
16-bit event counter	—	—	—	—	—	—	—	○	—
Warm-up counter	—	—	—	—	—	—	○	—	—
16-bit PWM	○	—	—	—	○	—	—	○	—
16-bit PPG	○	—	—	—	—	—	—	○	—

Note1: For 16-bit operations (16-bit timer/event counter, warm-up counter, 16-bit PWM and 16-bit PPG), set its source clock on lower bit (TC3CK).

Note2: ○ : Available source clock

Table 9-3 Constraints on Register Values Being Compared

Operating mode	Register Value
8-bit timer/event counter	$1 \leq (TTREGn) \leq 255$
8-bit PDO	$1 \leq (TTREGn) \leq 255$
8-bit PWM	$2 \leq (PWREGn) \leq 254$
16-bit timer/event counter	$1 \leq (TTREG4, 3) \leq 65535$
Warm-up counter	$256 \leq (TTREG4, 3) \leq 65535$
16-bit PWM	$2 \leq (PWREG4, 3) \leq 65534$
16-bit PPG	$1 \leq (PWREG4, 3) < (TTREG4, 3) \leq 65535$ and $(PWREG4, 3) + 1 < (TTREG4, 3)$

Note: n = 3 to 4

9.3 Function

The TimerCounter 3 and 4 have the 8-bit timer, 8-bit event counter, 8-bit programmable divider output (PDO), 8-bit pulse width modulation (PWM) output modes. The TimerCounter 3 and 4 (TC3, 4) are cascadable to form a 16-bit timer. The 16-bit timer has the operating modes such as the 16-bit timer, 16-bit event counter, warm-up counter, 16-bit pulse width modulation (PWM) output and 16-bit programmable pulse generation (PPG) modes.

9.3.1 8-Bit Timer Mode (TC3 and 4)

In the timer mode, the up-counter counts up using the internal clock. When a match between the up-counter and the timer register j (TTREG j) value is detected, an INTTC j interrupt is generated and the up-counter is cleared. After being cleared, the up-counter restarts counting.

Note 1: In the timer mode, fix TCjCR<TFFj> to 0. If not fixed, the PDO j , PWM j and PPG j pins may output pulses.

Note 2: In the timer mode, do not change the TTREG j setting while the timer is running. Since TTREG j is not in the shift register configuration in the timer mode, the new value programmed in TTREG j is in effect immediately after the programming. Therefore, if TTREG i is changed while the timer is running, an expected operation may not be obtained.

Note 3: $j = 3, 4$

Table 9-4 Source Clock for TimerCounter 3, 4 (Internal Clock)

Source Clock			Resolution		Maximum Time Setting	
NORMAL1/2, IDLE1/2 mode		SLOW1/2, SLEEP1/2 mode	$f_c = 16 \text{ MHz}$	$f_s = 32.768 \text{ kHz}$	$f_c = 16 \text{ MHz}$	$f_s = 32.768 \text{ kHz}$
DV7CK = 0	DV7CK = 1					
$f_c/2^{11} \text{ [Hz]}$	$f_s/2^3 \text{ [Hz]}$	$f_s/2^3 \text{ [Hz]}$	128 μs	244.14 μs	32.6 ms	62.3 ms
$f_c/2^7$	$f_c/2^7$	—	8 μs	—	2.0 ms	—
$f_c/2^5$	$f_c/2^5$	—	2 μs	—	510 μs	—
$f_c/2^3$	$f_c/2^3$	—	500 ns	—	127.5 μs	—

Example :Setting the timer mode with source clock $f_c/2^7 \text{ Hz}$ and generating an interrupt 80 μs later (TimerCounter4, $f_c = 16.0 \text{ MHz}$)

```
LD      (TTREG4), 0AH      : Sets the timer register ( $80 \mu\text{s} \div 2^7/f_c = 0AH$ ).
DI
SET     (EIRH), 3          : Enables INTTC4 interrupt.
EI
LD      (TC4CR), 00010000B : Sets the operating clock to  $f_c/2^7$ , and 8-bit timer mode.
LD      (TC4CR), 00011000B : Starts TC4.
```

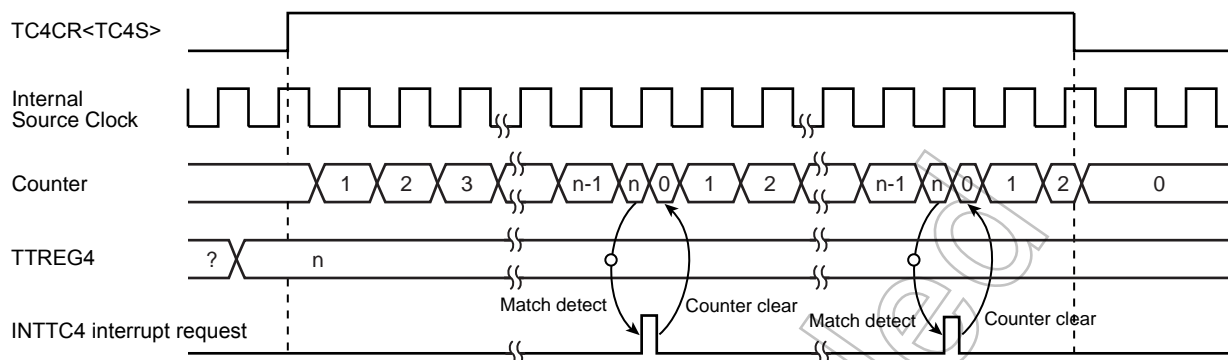


Figure 9-2 8-Bit Timer Mode Timing Chart (TC4)

9.3.2 8-Bit Event Counter Mode (TC3, 4)

In the 8-bit event counter mode, the up-counter counts up at the falling edge of the input pulse to the TCj pin. When a match between the up-counter and the TTREGj value is detected, an INTTCj interrupt is generated and the up-counter is cleared. After being cleared, the up-counter restarts counting at the falling edge of the input pulse to the TCj pin. Two machine cycles are required for the low- or high-level pulse input to the TCj pin. Therefore, a maximum frequency to be supplied is $f_c/2^4$ Hz in the NORMAL1/2 or IDLE1/2 mode, and $f_s/2^4$ Hz in the SLOW1/2 or SLEEP1/2 mode.

Note 1: In the event counter mode, fix TCjCR<TFFj> to 0. If not fixed, the \overline{PDOj} , \overline{PWMj} and \overline{PPGj} pins may output pulses.

Note 2: In the event counter mode, do not change the TTREGj setting while the timer is running. Since TTREGj is not in the shift register configuration in the event counter mode, the new value programmed in TTREGj is in effect immediately after the programming. Therefore, if TTREGj is changed while the timer is running, an expected operation may not be obtained.

Note 3: j = 3, 4

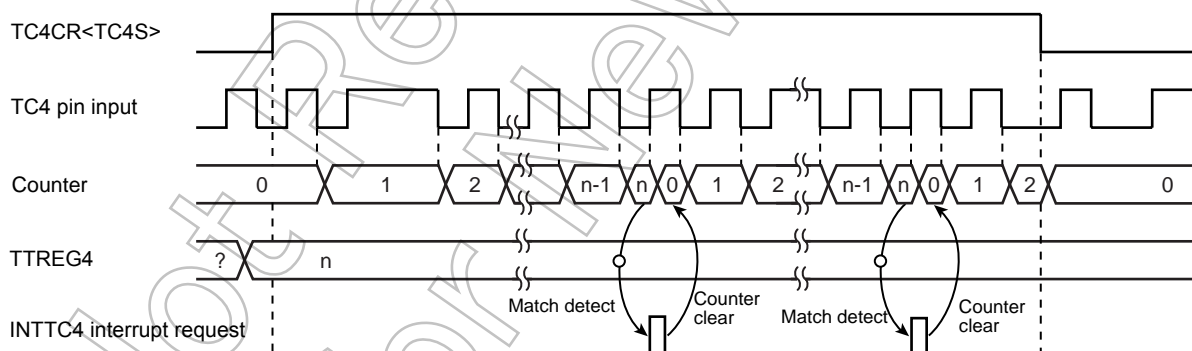


Figure 9-3 8-Bit Event Counter Mode Timing Chart (TC4)

9.3.3 8-Bit Programmable Divider Output (PDO) Mode (TC3, 4)

This mode is used to generate a pulse with a 50% duty cycle from the \overline{PDOj} pin.

In the PDO mode, the up-counter counts up using the internal clock. When a match between the up-counter and the TTREGj value is detected, the logic level output from the \overline{PDOj} pin is switched to the opposite state and the up-counter is cleared. The INTTCj interrupt request is generated at the time. The logic state opposite to the timer F/Fj logic level is output from the \overline{PDOj} pin. An arbitrary value can be set to the timer F/Fj by TCjCR<TFFj>. Upon reset, the timer F/Fj value is initialized to 0.

To use the programmable divider output, set the output latch of the I/O port to 1.

Example :Generating 1024 Hz pulse using TC4 ($f_c = 16.0 \text{ MHz}$)

Setting port

LD	(TTREG4), 3DH	: $1/1024 \div 2^7 / f_c \div 2 = 3DH$
LD	(TC4CR), 00010001B	: Sets the operating clock to $f_c/2^7$, and 8-bit PDO mode.
LD	(TC4CR), 00011001B	: Starts TC4.

Note 1: In the programmable divider output mode, do not change the TTREGj setting while the timer is running. Since TTREGj is not in the shift register configuration in the programmable divider output mode, the new value programmed in TTREGj is in effect immediately after programming. Therefore, if TTREGi is changed while the timer is running, an expected operation may not be obtained.

Note 2: When the timer is stopped during PDO output, the \overline{PDOj} pin holds the output status when the timer is stopped. To change the output status, program TCjCR<TFFj> after the timer is stopped. Do not change the TCjCR<TFFj> setting upon stopping of the timer.

Example: Fixing the \overline{PDOj} pin to the high level when the TimerCounter is stopped

CLR (TCjCR).3: Stops the timer.

CLR (TCjCR).7: Sets the \overline{PDOj} pin to the high level.

Note 3: j = 3, 4

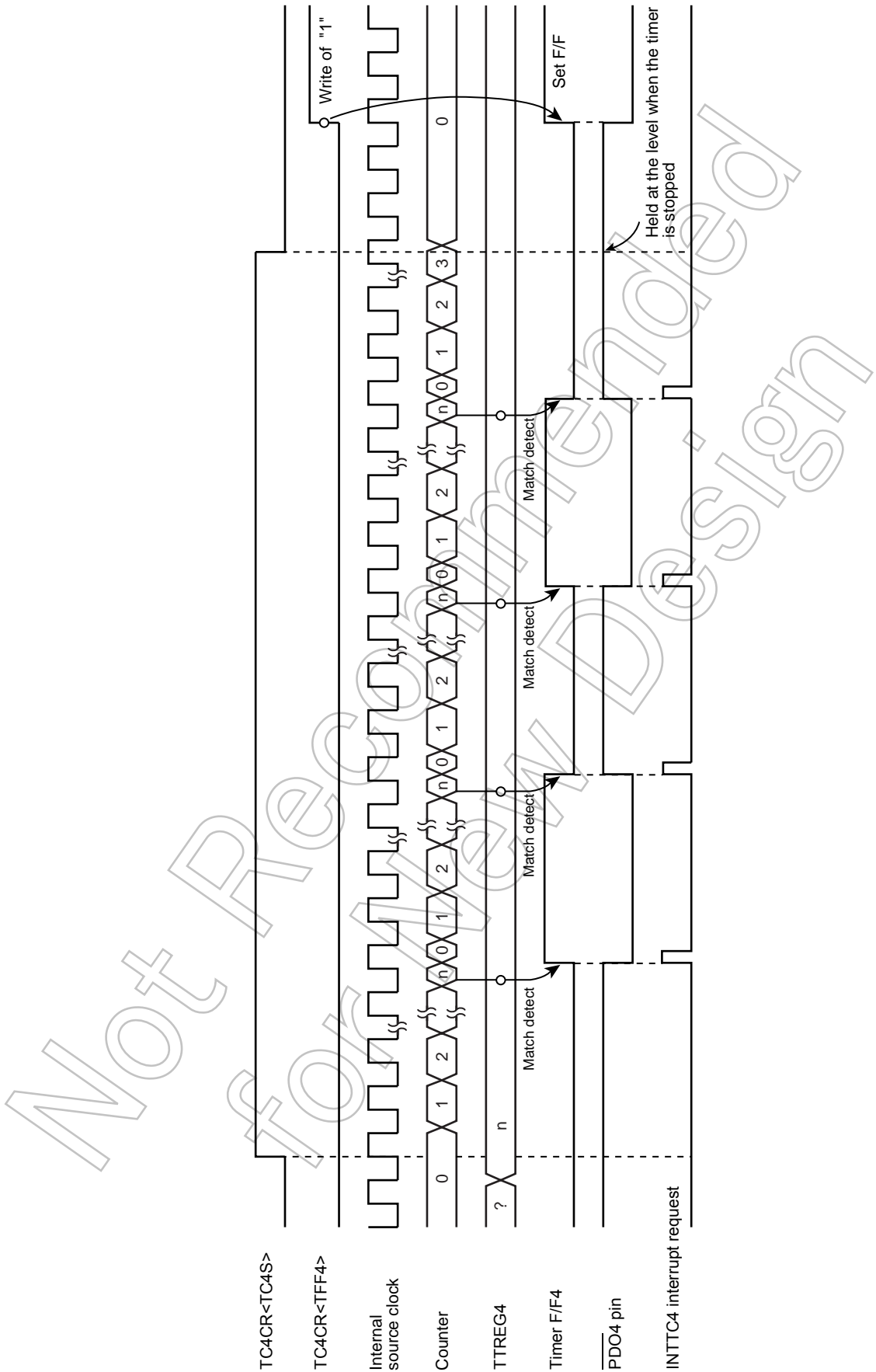


Figure 9-4 8-Bit PDO Mode Timing Chart (TC4)

9.3.4 8-Bit Pulse Width Modulation (PWM) Output Mode (TC3, 4)

This mode is used to generate a pulse-width modulated (PWM) signals with up to 8 bits of resolution. The up-counter counts up using the internal clock.

When a match between the up-counter and the PWREGj value is detected, the logic level output from the timer F/Fj is switched to the opposite state. The counter continues counting. The logic level output from the timer F/Fj is switched to the opposite state again by the up-counter overflow, and the counter is cleared. The INTTCj interrupt request is generated at this time.

Since the initial value can be set to the timer F/Fj by TCjCR<TFFj>, positive and negative pulses can be generated. Upon reset, the timer F/Fj is cleared to 0.

(The logic level output from the $\overline{\text{PWMj}}$ pin is the opposite to the timer F/Fj logic level.)

Since PWREGj in the PWM mode is serially connected to the shift register, the value set to PWREGj can be changed while the timer is running. The value set to PWREGj during a run of the timer is shifted by the INTTCj interrupt request and loaded into PWREGj. While the timer is stopped, the value is shifted immediately after the programming of PWREGj. If executing the read instruction to PWREGj during PWM output, the value in the shift register is read, but not the value set in PWREGj. Therefore, after writing to PWREGj, the reading data of PWREGj is previous value until INTTCj is generated.

For the pin used for PWM output, the output latch of the I/O port must be set to 1.

Note 1: In the PWM mode, program the timer register PWREGj immediately after the INTTCj interrupt request is generated (normally in the INTTCj interrupt service routine.) If the programming of PWREGj and the interrupt request occur at the same time, an unstable value is shifted, that may result in generation of the pulse different from the programmed value until the next INTTCj interrupt request is generated.

Note 2: When the timer is stopped during PWM output, the $\overline{\text{PWMj}}$ pin holds the output status when the timer is stopped. To change the output status, program TCjCR<TFFj> after the timer is stopped. Do not change the TCjCR<TFFj> upon stopping of the timer.

Example: Fixing the $\overline{\text{PWMj}}$ pin to the high level when the TimerCounter is stopped

CLR (TCjCR).3: Stops the timer.

CLR (TCjCR).7: Sets the $\overline{\text{PWMj}}$ pin to the high level.

Note 3: To enter the STOP mode during PWM output, stop the timer and then enter the STOP mode. If the STOP mode is entered without stopping the timer when fc, fc/2 or fs is selected as the source clock, a pulse is output from the $\overline{\text{PWMj}}$ pin during the warm-up period time after exiting the STOP mode.

Note 4: j = 3, 4

Table 9-5 PWM Output Mode

Source Clock			Resolution		Repeated Cycle	
NORMAL1/2, IDLE1/2 mode		SLOW1/2, SLEEP1/2 mode	fc = 16 MHz	fs = 32.768 kHz	fc = 16 MHz	fs = 32.768 kHz
DV7CK = 0	DV7CK = 1					
fc/2 ¹¹ [Hz]	fs/2 ³ [Hz]	fs/2 ³ [Hz]	128 μs	244.14 μs	32.8 ms	62.5 ms
fc/2 ⁷	fc/2 ⁷	—	8 μs	—	2.05 ms	—
fc/2 ⁵	fc/2 ⁵	—	2 μs	—	512 μs	—
fc/2 ³	fc/2 ³	—	500 ns	—	128 μs	—
fs	fs	fs	30.5 μs	30.5 μs	7.81 ms	7.81 ms
fc/2	fc/2	—	125 ns	—	32 μs	—
fc	fc	—	62.5 ns	—	16 μs	—

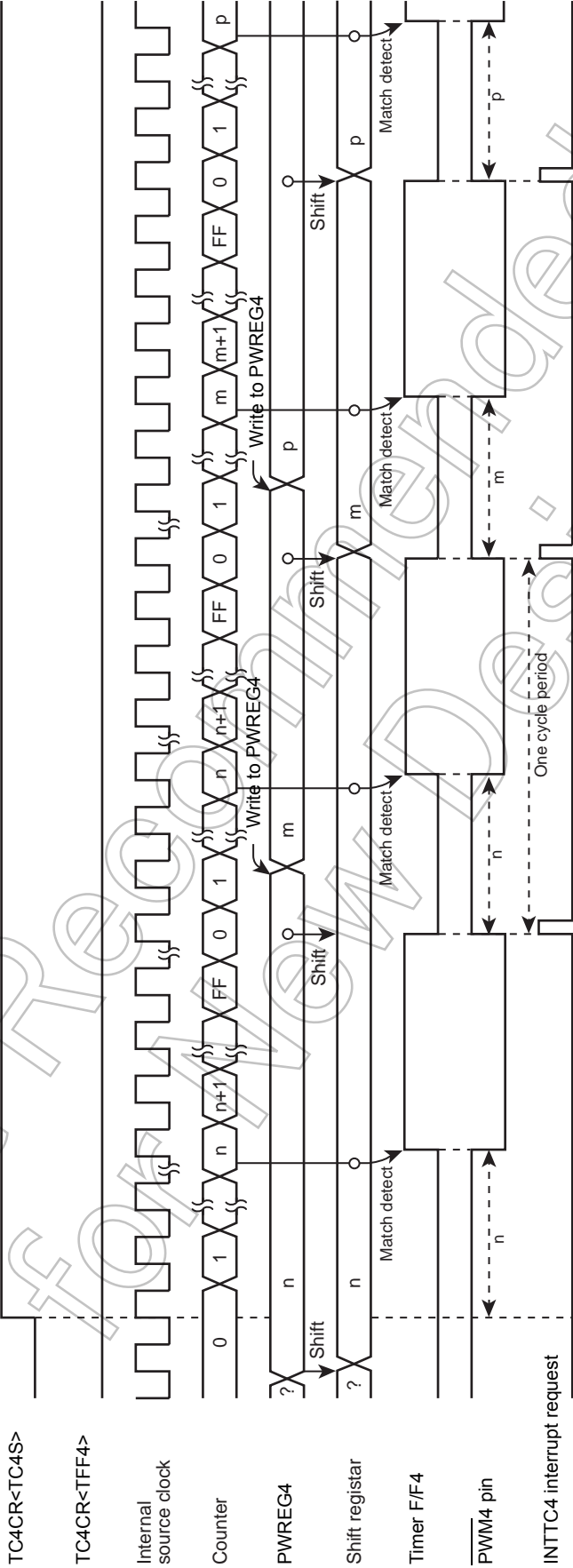


Figure 9-5 8-Bit PWM Mode Timing Chart (TC4)

9.3.5 16-Bit Timer Mode (TC3 and 4)

In the timer mode, the up-counter counts up using the internal clock. The TimerCounter 3 and 4 are cascaded to form a 16-bit timer.

When a match between the up-counter and the timer register (TTREG3, TTREG4) value is detected after the timer is started by setting TC4CR<TC4S> to 1, an INTTC4 interrupt is generated and the up-counter is cleared. After being cleared, the up-counter continues counting. Program the lower byte and upper byte in this order in the timer register. (Programming only the upper or lower byte should not be attempted.)

Note 1: In the timer mode, fix TCjCR<TFFj> to 0. If not fixed, the $\overline{P\bar{D}Oj}$, $\overline{P\bar{W}Mj}$, and $\overline{P\bar{P}Gj}$ pins may output a pulse.

Note 2: In the timer mode, do not change the TTREGj setting while the timer is running. Since TTREGj is not in the shift register configuration in the timer mode, the new value programmed in TTREGj is in effect immediately after programming of TTREGj. Therefore, if TTREGj is changed while the timer is running, an expected operation may not be obtained.

Note 3: j = 3, 4

Table 9-6 Source Clock for 16-Bit Timer Mode

Source Clock			Resolution		Maximum Time Setting	
NORMAL1/2, IDLE1/2 mode		SLOW1/2, SLEEP1/2 mode	fc = 16 MHz	fs = 32.768 kHz	fc = 16 MHz	fs = 32.768 kHz
DV7CK = 0	DV7CK = 1					
fc/2 ¹¹	fs/2 ³	fs/2 ³	128 μ s	244.14 μ s	8.39 s	16 s
fc/2 ⁷	fc/2 ⁷	—	8 μ s	—	524.3 ms	—
fc/2 ⁵	fc/2 ⁵	—	2 μ s	—	131.1 ms	—
fc/2 ³	fc/2 ³	—	500 ns	—	32.8 ms	—

Example :Setting the timer mode with source clock fc/2⁷ Hz, and generating an interrupt 300 ms later
(fc = 16.0 MHz)

LDW (TTREG3), 927CH : Sets the timer register (300 ms = 2⁷/fc = 927CH).

DI

SET (EIRH). 3 : Enables INTTC4 interrupt.

EI

LD (TC3CR), 13H : Sets the operating clock to fc/2⁷, and 16-bit timer mode (lower byte).

LD (TC4CR), 04H : Sets the 16-bit timer mode (upper byte).

LD (TC4CR), 0CH : Starts the timer.

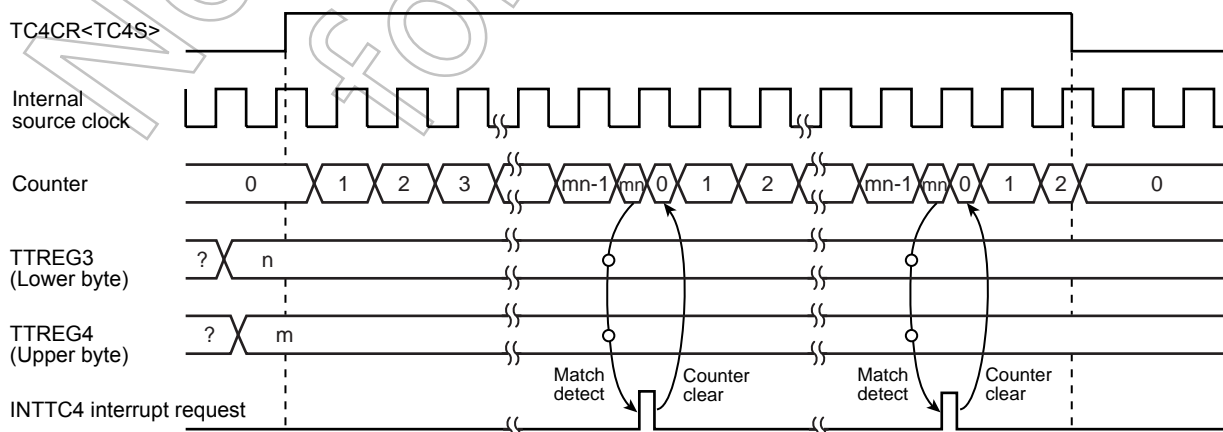


Figure 9-6 16-Bit Timer Mode Timing Chart (TC3 and TC4)

9.3.6 16-Bit Event Counter Mode (TC3 and 4)

In the event counter mode, the up-counter counts up at the falling edge to the TC3 pin. The TimerCounter 3 and 4 are cascable to form a 16-bit event counter.

When a match between the up-counter and the timer register (TTREG3, TTREG4) value is detected after the timer is started by setting TC4CR<TC4S> to 1, an INTTC4 interrupt is generated and the up-counter is cleared.

After being cleared, the up-counter restarts counting at the falling edge of the input pulse to the TC3 pin. Two machine cycles are required for the low- or high-level pulse input to the TC3 pin.

Therefore, a maximum frequency to be supplied is $f_c/2^4$ Hz in the NORMAL1/2 or IDLE1/2 mode, and $f_s/2^4$ in the SLOW1/2 or SLEEP1/2 mode. Program the lower byte (TTREG3), and upper byte (TTREG4) in this order in the timer register. (Programming only the upper or lower byte should not be attempted.)

Note 1: In the event counter mode, fix TCjCR<TFFj> to 0. If not fixed, the \overline{PDOj} , \overline{PWMj} and \overline{PPGj} pins may output pulses.

Note 2: In the event counter mode, do not change the TTREGj setting while the timer is running. Since TTREGj is not in the shift register configuration in the event counter mode, the new value programmed in TTREGj is in effect immediately after the programming. Therefore, if TTREGj is changed while the timer is running, an expected operation may not be obtained.

Note 3: j = 3, 4

9.3.7 16-Bit Pulse Width Modulation (PWM) Output Mode (TC3 and 4)

This mode is used to generate a pulse-width modulated (PWM) signals with up to 16 bits of resolution. The TimerCounter 3 and 4 are cascable to form the 16-bit PWM signal generator.

The counter counts up using the internal clock or external clock.

When a match between the up-counter and the timer register (PWREG3, PWREG4) value is detected, the logic level output from the timer F/F4 is switched to the opposite state. The counter continues counting. The logic level output from the timer F/F4 is switched to the opposite state again by the counter overflow, and the counter is cleared. The INTTC4 interrupt is generated at this time.

Two machine cycles are required for the high- or low-level pulse input to the TC3 pin. Therefore, a maximum frequency to be supplied is $f_c/2^4$ Hz in the NORMAL1/2 or IDLE1/2 mode, and $f_s/2^4$ to in the SLOW1/2 or SLEEP1/2 mode.

Since the initial value can be set to the timer F/F4 by TC4CR<TFF4>, positive and negative pulses can be generated. Upon reset, the timer F/F4 is cleared to 0.

(The logic level output from the $\overline{PWM4}$ pin is the opposite to the timer F/F4 logic level.)

Since PWREG4 and 3 in the PWM mode are serially connected to the shift register, the values set to PWREG4 and 3 can be changed while the timer is running. The values set to PWREG4 and 3 during a run of the timer are shifted by the INTTCj interrupt request and loaded into PWREG4 and 3. While the timer is stopped, the values are shifted immediately after the programming of PWREG4 and 3. Set the lower byte (PWREG3) and upper byte (PWREG4) in this order to program PWREG4 and 3. (Programming only the lower or upper byte of the register should not be attempted.)

If executing the read instruction to PWREG4 and 3 during PWM output, the values set in the shift register is read, but not the values set in PWREG4 and 3. Therefore, after writing to the PWREG4 and 3, reading data of PWREG4 and 3 is previous value until INTTC4 is generated.

For the pin used for PWM output, the output latch of the I/O port must be set to 1.

Note 1: In the PWM mode, program the timer register PWREG4 and 3 immediately after the INTTC4 interrupt request is generated (normally in the INTTC4 interrupt service routine.) If the programming of PWREGj and the interrupt request occur at the same time, an unstable value is shifted, that may result in generation of pulse different from the programmed value until the next INTTC4 interrupt request is generated.

Note 2: When the timer is stopped during PWM output, the $\overline{PWM4}$ pin holds the output status when the timer is stopped. To change the output status, program TC4CR<TFF4> after the timer is stopped. Do not program TC4CR<TFF4> upon stopping of the timer.

Example: Fixing the $\overline{PWM4}$ pin to the high level when the TimerCounter is stopped

CLR (TC4CR).3: Stops the timer.

CLR (TC4CR).7 : Sets the PWM4 pin to the high level.

Note 3: To enter the STOP mode, stop the timer and then enter the STOP mode. If the STOP mode is entered without stopping of the timer when f_c , $f_c/2$ or f_s is selected as the source clock, a pulse is output from the PWM4 pin during the warm-up period time after exiting the STOP mode.

Table 9-7 16-Bit PWM Output Mode

Source Clock			Resolution		Repeated Cycle	
NORMAL1/2, IDLE1/2 mode		SLOW1/2, SLEEP1/2 mode	$f_c = 16 \text{ MHz}$	$f_s = 32.768 \text{ kHz}$	$f_c = 16 \text{ MHz}$	$f_s = 32.768 \text{ kHz}$
DV7CK = 0	DV7CK = 1					
$f_c/2^{11}$	$f_s/2^3 \text{ [Hz]}$	$f_s/2^3 \text{ [Hz]}$	128 μs	244.14 μs	8.39 s	16 s
$f_c/2^7$	$f_c/2^7$	—	8 μs	—	524.3 ms	—
$f_c/2^5$	$f_c/2^5$	—	2 μs	—	131.1 ms	—
$f_c/2^3$	$f_c/2^3$	—	500 ns	—	32.8 ms	—
f_s	f_s	f_s	30.5 μs	30.5 μs	2 s	2 s
$f_c/2$	$f_c/2$	—	125 ns	—	8.2 ms	—
f_c	f_c	—	62.5 ns	—	4.1 ms	—

Example :Generating a pulse with 1-ms high-level width and a period of 32.768 ms ($f_c = 16.0 \text{ MHz}$)

Setting ports

LDW (PWREG3), 07D0H : Sets the pulse width.

LD (TC3CR), 33H : Sets the operating clock to $f_c/2^3$, and 16-bit PWM output mode (lower byte).

LD (TC4CR), 056H : Sets TFF4 to the initial value 0, and 16-bit PWM signal generation mode (upper byte).

LD (TC4CR), 05EH : Starts the timer.

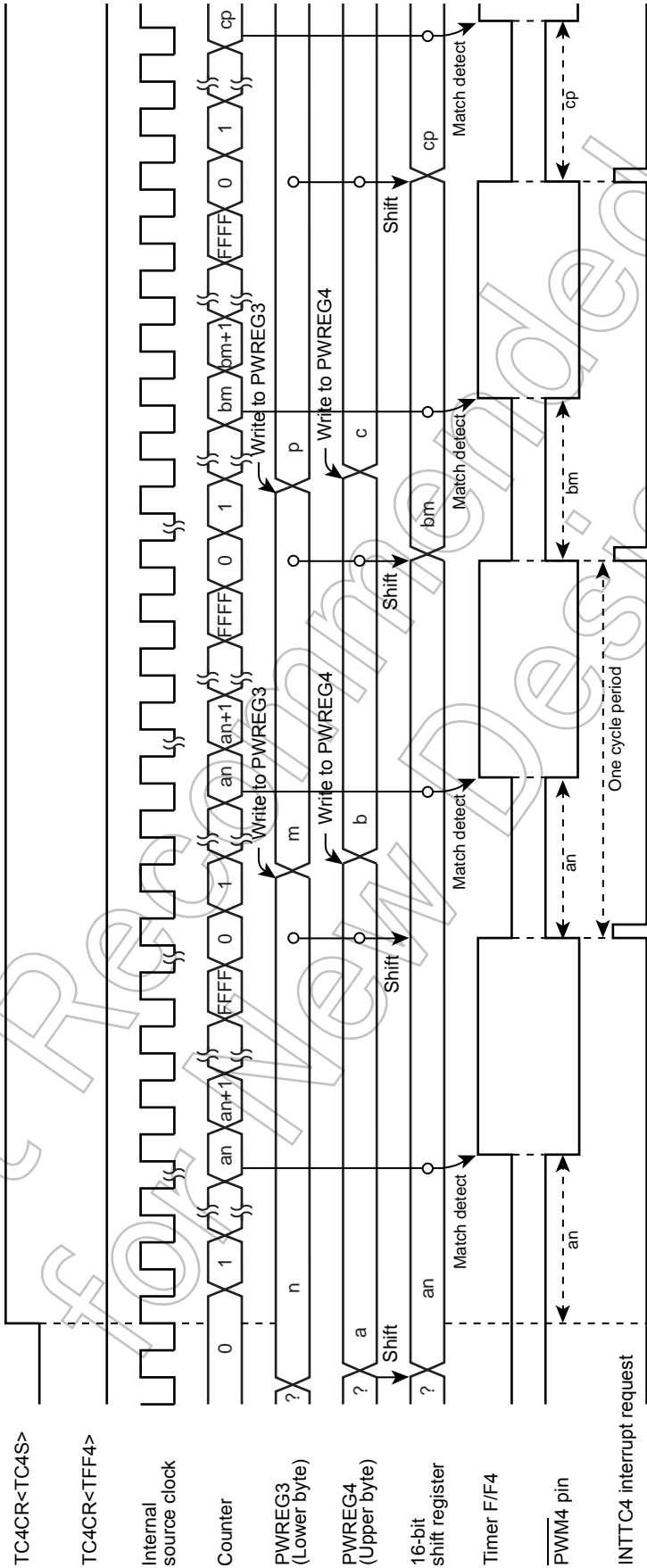


Figure 9-7 16-Bit PWM Mode Timing Chart (TC3 and TC4)

9.3.8 16-Bit Programmable Pulse Generate (PPG) Output Mode (TC3 and 4)

This mode is used to generate pulses with up to 16-bits of resolution. The timer counter 3 and 4 are cascaded to enter the 16-bit PPG mode.

The counter counts up using the internal clock or external clock. When a match between the up-counter and the timer register (PWREG3, PWREG4) value is detected, the logic level output from the timer F/F4 is switched to the opposite state. The counter continues counting. The logic level output from the timer F/F4 is switched to the opposite state again when a match between the up-counter and the timer register (TTREG3, TTREG4) value is detected, and the counter is cleared. The INTTC4 interrupt is generated at this time.

Two machine cycles are required for the high- or low-level pulse input to the TC3 pin. Therefore, a maximum frequency to be supplied is $f_c/2^4$ Hz in the NORMAL1/2 or IDLE1/2 mode, and $f_s/2^4$ to in the SLOW1/2 or SLEEP1/2 mode.

Since the initial value can be set to the timer F/F4 by TC4CR<TFF4>, positive and negative pulses can be generated. Upon reset, the timer F/F4 is cleared to 0.

(The logic level output from the $\overline{\text{PPG4}}$ pin is the opposite to the timer F/F4.)

Set the lower byte and upper byte in this order to program the timer register. (TTREG3 → TTREG4, PWREG3 → PWREG4) (Programming only the upper or lower byte should not be attempted.)

For PPG output, set the output latch of the I/O port to 1.

Example :Generating a pulse with 1-ms high-level width and a period of 16.385 ms ($f_c = 16.0$ MHz)

Setting ports		
LDW	(PWREG3), 07D0H	: Sets the pulse width.
LDW	(TTREG3), 8002H	: Sets the cycle period.
LD	(TC3CR), 33H	: Sets the operating clock to $f_c/2^3$, and 16-bit PPG mode (lower byte).
LD	(TC4CR), 057H	: Sets TFF4 to the initial value 0, and 16-bit PPG mode (upper byte).
LD	(TC4CR), 05FH	: Starts the timer.

Note 1: In the PPG mode, do not change the PWREGi and TTREGi settings while the timer is running. Since PWREGi and TTREGi are not in the shift register configuration in the PPG mode, the new values programmed in PWREGi and TTREGi are in effect immediately after programming PWREGi and TTREGi. Therefore, if PWREGi and TTREGi are changed while the timer is running, an expected operation may not be obtained.

Note 2: When the timer is stopped during PPG output, the $\overline{\text{PPG4}}$ pin holds the output status when the timer is stopped. To change the output status, program TC4CR<TFF4> after the timer is stopped. Do not change TC4CR<TFF4> upon stopping of the timer.

Example: Fixing the $\overline{\text{PPG4}}$ pin to the high level when the TimerCounter is stopped

CLR (TC4CR).3: Stops the timer

CLR (TC4CR).7: Sets the $\overline{\text{PPG4}}$ pin to the high level

Note 3: i = 3, 4

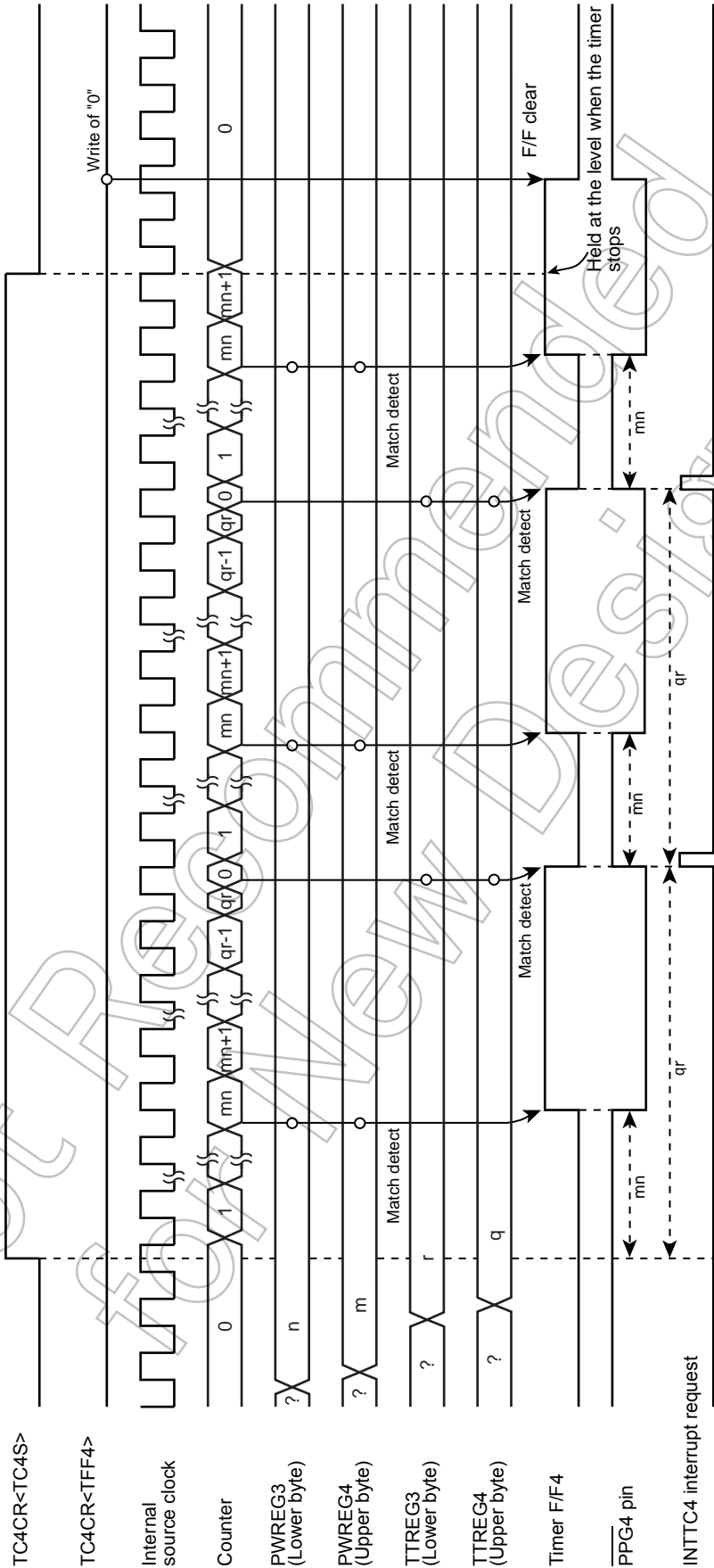


Figure 9-8 16-Bit PPG Mode Timing Chart (TC3 and TC4)

9.3.9 Warm-Up Counter Mode

In this mode, the warm-up period time is obtained to assure oscillation stability when the system clocking is switched between the high-frequency and low-frequency. The timer counter 3 and 4 are cascadable to form a 16-bit TimerCounter. The warm-up counter mode has two types of mode; switching from the high-frequency to low-frequency, and vice-versa.

Note 1: In the warm-up counter mode, fix TCiCR<TFFi> to 0. If not fixed, the $\overline{PD0i}$, \overline{PWMi} and \overline{PPGi} pins may output pulses.

Note 2: In the warm-up counter mode, only upper 8 bits of the timer register TTREG4 and 3 are used for match detection and lower 8 bits are not used.

Note 3: i = 3, 4

9.3.9.1 Low-Frequency Warm-up Counter Mode (NORMAL1 → NORMAL2 → SLOW2 → SLOW1)

In this mode, the warm-up period time from a stop of the low-frequency clock f_s to oscillation stability is obtained. Before starting the timer, set SYSCR2<XTEN> to 1 to oscillate the low-frequency clock. When a match between the up-counter and the timer register (TTREG4, 3) value is detected after the timer is started by setting TC4CR<TC4S> to 1, the counter is cleared by generating the INTTC4 interrupt request. After stopping the timer in the INTTC4 interrupt service routine, set SYSCR2<SYSCK> to 1 to switch the system clock from the high-frequency to low-frequency, and then clear of SYSCR2<XEN> to 0 to stop the high-frequency clock.

Table 9-8 Setting Time of Low-Frequency Warm-Up Counter Mode ($f_s = 32.768$ kHz)

Minimum Time Setting (TTREG4, 3 = 0100H)	Maximum Time Setting (TTREG4, 3 = FF00H)
7.81 ms	1.99 s

Example :After checking low-frequency clock oscillation stability with TC4 and 3, switching to the SLOW1 mode

```

SET      (SYSCR2).6      : SYSCR2<XTEN> ← 1
LD       (TC3CR), 43H    : Sets TFF3=0, source clock  $f_s$ , and 16-bit mode.
LD       (TC4CR), 05H    : Sets TFF4=0, and warm-up counter mode.
LD       (TTREG3), 8000H : Sets the warm-up time.
                        : (The warm-up time depends on the oscillator characteristic.)
DI       : IMF ← 0
SET      (EIRH). 3      : Enables the INTTC4.
EI       : IMF ← 1
SET      (TC4CR).3      : Starts TC4 and 3.
:         :
PINTTC4: CLR      (TC4CR).3 : Stops TC4 and 3.
SET      (SYSCR2).5      : SYSCR2<SYSCK> ← 1
                        : (Switches the system clock to the low-frequency clock.)
CLR      (SYSCR2).7      : SYSCR2<XEN> ← 0 (Stops the high-frequency clock.)
RETI
:         :
VINTTC4: DW       PINTTC4 : INTTC4 vector table

```


9.3.9.2 High-Frequency Warm-Up Counter Mode
(SLOW1 → SLOW2 → NORMAL2 → NORMAL1)

In this mode, the warm-up period time from a stop of the high-frequency clock *fc* to the oscillation stability is obtained. Before starting the timer, set SYSCR2<XEN> to 1 to oscillate the high-frequency clock. When a match between the up-counter and the timer register (TTREG4, 3) value is detected after the timer is started by setting TC4CR<TC4S> to 1, the counter is cleared by generating the INTTC4 interrupt request. After stopping the timer in the INTTC4 interrupt service routine, clear SYSCR2<SYSCK> to 0 to switch the system clock from the low-frequency to high-frequency, and then SYSCR2<XTEN> to 0 to stop the low-frequency clock.

Table 9-9 Setting Time in High-Frequency Warm-Up Counter Mode

Minimum time Setting (TTREG4, 3 = 0100H)	Maximum time Setting (TTREG4, 3 = FF00H)
16 μs	4.08 ms

Example :After checking high-frequency clock oscillation stability with TC4 and 3, switching to the NORMAL1 mode

```

SET      (SYSCR2).7      : SYSCR2<XEN> ← 1
LD       (TC3CR), 63H    : Sets TFF3=0, source clock fc, and 16-bit mode.
LD       (TC4CR), 05H    : Sets TFF4=0, and warm-up counter mode.
LD       (TTREG3), 0F800H : Sets the warm-up time.
                                (The warm-up time depends on the oscillator characteristic.)
DI       : IMF ← 0
SET      (EIRH). 3      : Enables the INTTC4.
EI       : IMF ← 1
SET      (TC4CR).3      : Starts the TC4 and 3.
:       :
PINTTC4: CLR      (TC4CR).3 : Stops the TC4 and 3.
        CLR      (SYSCR2).5 : SYSCR2<SYSCK> ← 0
                                (Switches the system clock to the high-frequency clock.)
        CLR      (SYSCR2).6 : SYSCR2<XTEN> ← 0
                                (Stops the low-frequency clock.)
        RETI
:       :
VINTTC4: DW       PINTTC4 : INTTC4 vector table
    
```

10. Synchronous Serial Interface (SIO)

The TMP86P820FG has a clocked-synchronous 8-bit serial interface. Serial interface has an 8-byte transmit and receive data buffer that can automatically and continuously transfer up to 64 bits of data.

Serial interface is connected to outside peripheral devices via SO, SI, SCK port.

10.1 Configuration

SIO control / status register

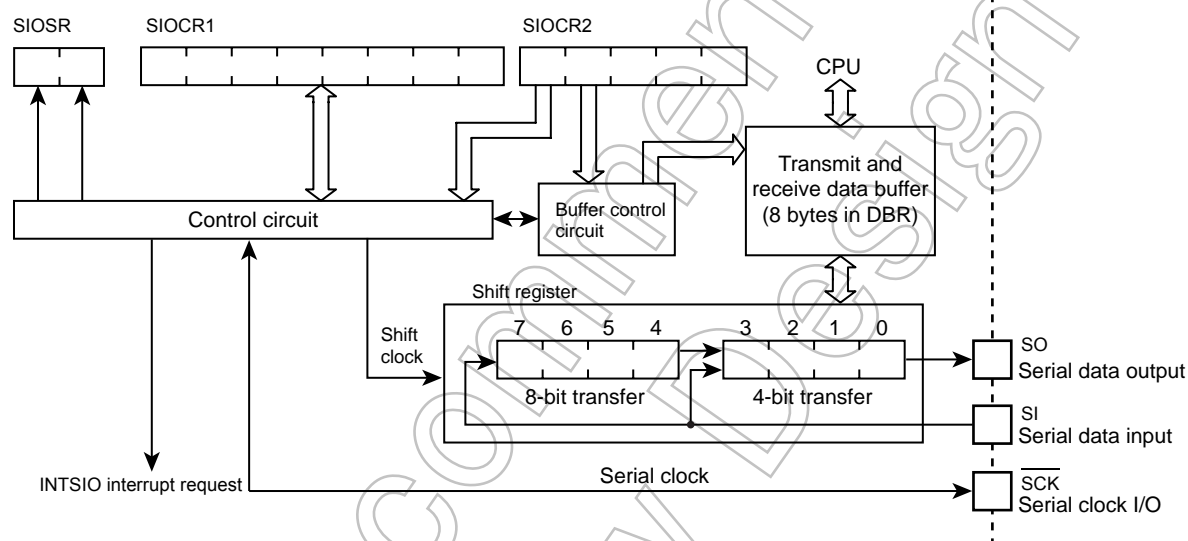


Figure 10-1 Serial Interface

10.2 Control

The serial interface is controlled by SIO control registers (SIOCR1/SIOCR2). The serial interface status can be determined by reading SIO status register (SIOSR).

The transmit and receive data buffer is controlled by the SIOCR2<BUF>. The data buffer is assigned to address 0F90H to 0F97H for SIO in the DBR area, and can continuously transfer up to 8 words (bytes or nibbles) at one time. When the specified number of words has been transferred, a buffer empty (in the transmit mode) or a buffer full (in the receive mode or transmit/receive mode) interrupt (INTSIO) is generated.

When the internal clock is used as the serial clock in the 8-bit receive mode and the 8-bit transmit/receive mode, a fixed interval wait can be applied to the serial clock for each word transferred. Four different wait times can be selected with SIOCR2<WAIT>.

SIO Control Register 1

SIOCR1	7	6	5	4	3	2	1	0
(0F98H)	SIOS	SIOINH	SIOM			SCK		

(Initial value: 0000 0000)

SIOS	Indicate transfer start / stop	0: Stop 1: Start	Write only			
SIOINH	Continue / abort transfer	0: Continuously transfer 1: Abort transfer (Automatically cleared after abort)				
SIOM	Transfer mode select	000: 8-bit transmit mode 010: 4-bit transmit mode 100: 8-bit transmit / receive mode 101: 8-bit receive mode 110: 4-bit receive mode Except the above: Reserved				
SCK	Serial clock select		NORMAL1/2, IDLE1/2 mode	SLOW1/2 SLEEP1/2 mode	Write only	
			DV7CK = 0			DV7CK = 1
		000	$fc/2^{13}$	$fs/2^5$		$fs/2^5$
		001	$fc/2^8$	$fc/2^8$		-
		010	$fc/2^7$	$fc/2^7$		-
		011	$fc/2^6$	$fc/2^6$		-
		100	$fc/2^5$	$fc/2^5$		-
		101	$fc/2^4$	$fc/2^4$		-
		110	Reserved			
111	External clock (Input from SCK pin)					

- Note 1: fc; High-frequency clock [Hz], fs; Low-frequency clock [Hz]
Note 2: Set SIOS to "0" and SIOINH to "1" when setting the transfer mode or serial clock.
Note 3: SIOCR1 is write-only register, which cannot access any of in read-modify-write instruction such as bit operate, etc.

SIO Control Register 2

SIOCR2	7	6	5	4	3	2	1	0
(0F99H)				WAIT		BUF		

(Initial value: ***0 0000)

WAIT	Wait control	Always sets "00" except 8-bit transmit / receive mode. 00: $T_f = T_D$ (Non wait) 01: $T_f = 2T_D$ (Wait) 10: $T_f = 4T_D$ (Wait) 11: $T_f = 8T_D$ (Wait)	
BUF	Number of transfer words (Buffer address in use)	000: 1 word transfer 0F90H 001: 2 words transfer 0F90H ~ 0F91H 010: 3 words transfer 0F90H ~ 0F92H 011: 4 words transfer 0F90H ~ 0F93H 100: 5 words transfer 0F90H ~ 0F94H 101: 6 words transfer 0F90H ~ 0F95H 110: 7 words transfer 0F90H ~ 0F96H 111: 8 words transfer 0F90H ~ 0F97H	Write only

Note 1: The lower 4 bits of each buffer are used during 4-bit transfers. Zeros (0) are stored to the upper 4bits when receiving.

Note 2: Transmitting starts at the lowest address. Received data are also stored starting from the lowest address to the highest address. (The first buffer address transmitted is 0F90H).

Note 3: The value to be loaded to BUF is held after transfer is completed.

Note 4: SIOCR2 must be set when the serial interface is stopped (SIOF = 0).

Note 5: *: Don't care

Note 6: SIOCR2 is write-only register, which cannot access any of in read-modify-write instruction such as bit operate, etc.

SIO Status Register

SIOSR	7	6	5	4	3	2	1	0
(0F99H)	SIOF	SEF						

SIOF	Serial transfer operating status monitor	0: Transfer terminated 1: Transfer in process	Read only
SEF	Shift operating status monitor	0: Shift operation terminated 1: Shift operation in process	

Note 1: T_f : Frame time, T_D : Data transfer time

Note 2: After SIOS is cleared to "0", SIOF is cleared to "0" at the termination of transfer or the setting of SIOINH to "1".

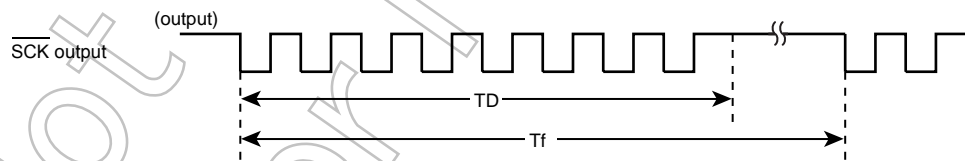


Figure 10-2 Frame time (T_f) and Data transfer time (T_D)

10.3 Serial clock

10.3.1 Clock source

Internal clock or external clock for the source clock is selected by SIOCR1<SCK>.

10.3.1.1 Internal clock

Any of six frequencies can be selected. The serial clock is output to the outside on the SCK pin. The SCK pin goes high when transfer starts.

When data writing (in the transmit mode) or reading (in the receive mode or the transmit/receive mode) cannot keep up with the serial clock rate, there is a wait function that automatically stops the serial clock and holds the next shift operation until the read/write processing is completed.

Table 10-1 Serial Clock Rate

	NORMAL 1/2, IDLE 1/2 mode				SLOW 1/2, SLEEP 1/2 mode	
	DV7CK = 0		DV7CK = 1			
SCK	Clock	Baud Rate	Clock	Baud Rate	Clock	Baud Rate
000	$f_c/2^{13}$	1.91 Kbps	$f_s/2^5$	1024 bps	$f_s/2^5$	1024 bps
001	$f_c/2^8$	61.04 Kbps	$f_c/2^8$	61.04 Kbps	-	-
010	$f_c/2^7$	122.07 Kbps	$f_c/2^7$	122.07 Kbps	-	-
011	$f_c/2^6$	244.14 Kbps	$f_c/2^6$	244.14 Kbps	-	-
100	$f_c/2^5$	488.28 Kbps	$f_c/2^5$	488.28 Kbps	-	-
101	$f_c/2^4$	976.56 Kbps	$f_c/2^4$	976.56 Kbps	-	-
110	-	-	-	-	-	-
111	External	External	External	External	External	External

Note: 1 Kbit = 1024 bit ($f_c = 16$ MHz, $f_s = 32.768$ kHz)

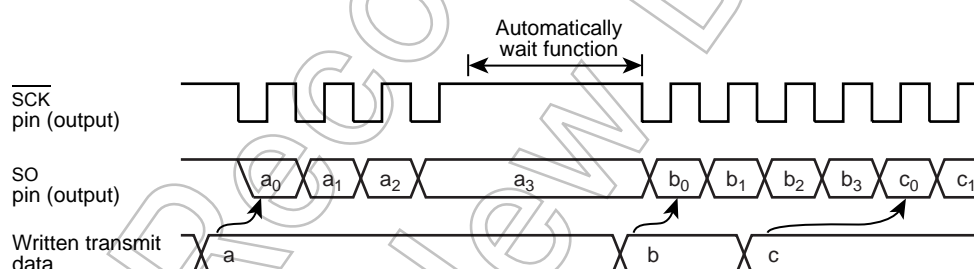


Figure 10-3 Automatic Wait Function (at 4-bit transmit mode)

10.3.1.2 External clock

An external clock connected to the SCK pin is used as the serial clock. In this case, output latch of this port should be set to "1". To ensure shifting, a pulse width of at least 4 machine cycles is required. This pulse is needed for the shift operation to execute certainly. Actually, there is necessary processing time for interrupting, writing, and reading. The minimum pulse is determined by setting the mode and the program. Therefore, maximum transfer frequency will be 488.3K bit/sec (at $f_c=16$ MHz).

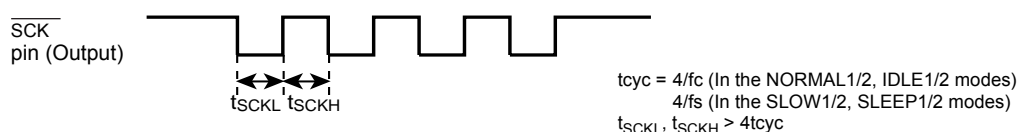


Figure 10-4 External clock pulse width

10.3.2 Shift edge

The leading edge is used to transmit, and the trailing edge is used to receive.

10.3.2.1 Leading edge

Transmitted data are shifted on the leading edge of the serial clock (falling edge of the $\overline{\text{SCK}}$ pin input/output).

10.3.2.2 Trailing edge

Received data are shifted on the trailing edge of the serial clock (rising edge of the $\overline{\text{SCK}}$ pin input/output).

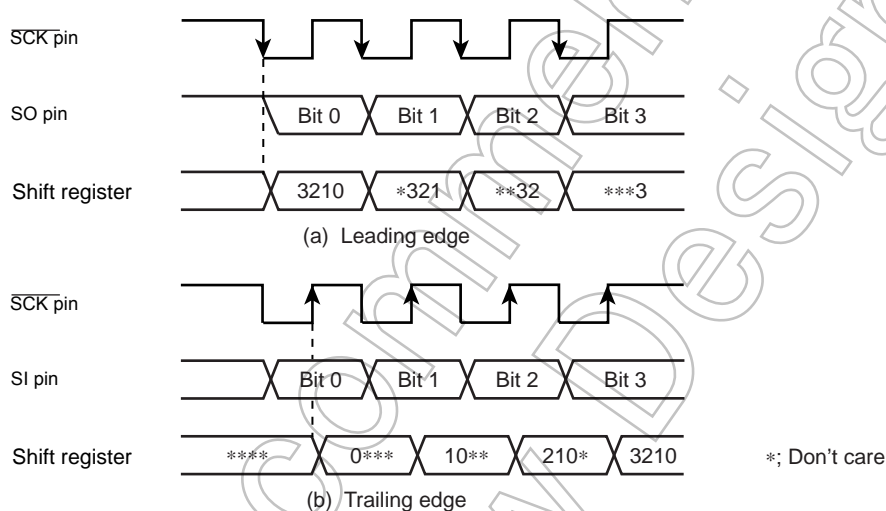


Figure 10-5 Shift edge

10.4 Number of bits to transfer

Either 4-bit or 8-bit serial transfer can be selected. When 4-bit serial transfer is selected, only the lower 4 bits of the transmit/receive data buffer register are used. The upper 4 bits are cleared to "0" when receiving. The data is transferred in sequence starting at the least significant bit (LSB).

10.5 Number of words to transfer

Up to 8 words consisting of 4 bits of data (4-bit serial transfer) or 8 bits (8-bit serial transfer) of data can be transferred continuously. The number of words to be transferred can be selected by $\text{SIOCR2}<\text{BUF}>$.

An INTSIO interrupt is generated when the specified number of words has been transferred. If the number of words is to be changed during transfer, the serial interface must be stopped before making the change. The number of words can be changed during automatic-wait operation of an internal clock. In this case, the serial interface is not required to be stopped.

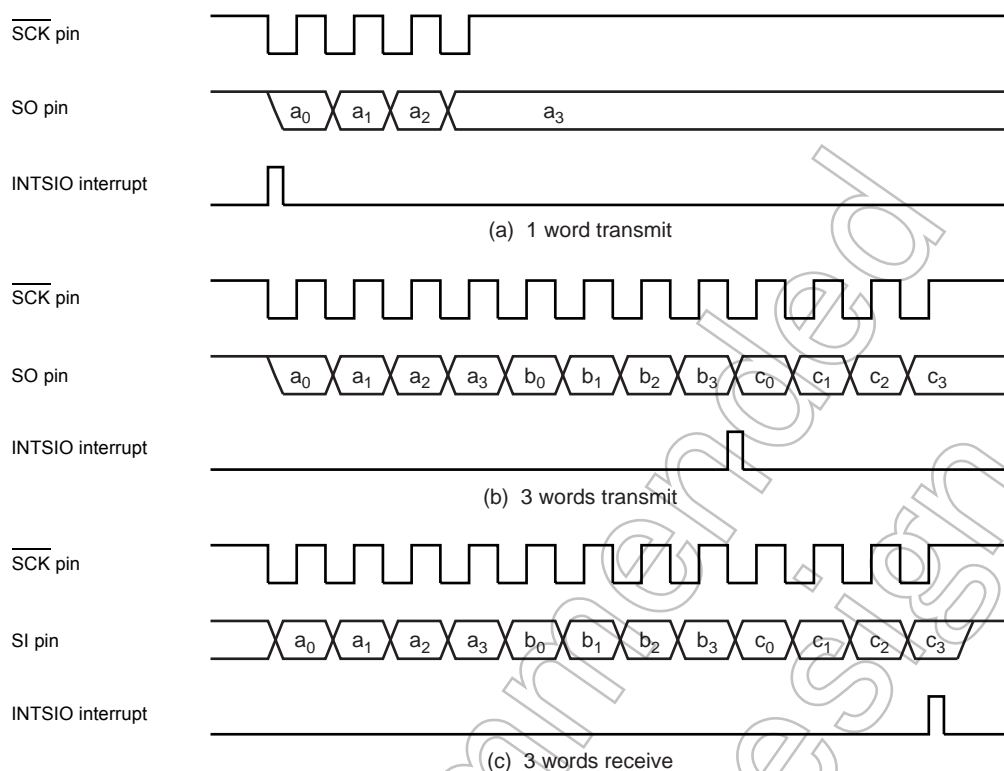


Figure 10-6 Number of words to transfer (Example: 1word = 4bit)

10.6 Transfer Mode

SIOCR1<SIOM> is used to select the transmit, receive, or transmit/receive mode.

10.6.1 4-bit and 8-bit transfer modes

In these modes, firstly set the SIO control register to the transmit mode, and then write first transmit data (number of transfer words to be transferred) to the data buffer registers (DBR).

After the data are written, the transmission is started by setting SIOCR1<SIOS> to “1”. The data are then output sequentially to the SO pin in synchronous with the serial clock, starting with the least significant bit (LSB). As soon as the LSB has been output, the data are transferred from the data buffer register to the shift register. When the final data bit has been transferred and the data buffer register is empty, an INTSIO (Buffer empty) interrupt is generated to request the next transmitted data.

When the internal clock is used, the serial clock will stop and an automatic-wait will be initiated if the next transmitted data are not loaded to the data buffer register by the time the number of data words specified with the SIOCR2<BUF> has been transmitted. Writing even one word of data cancels the automatic-wait; therefore, when transmitting two or more words, always write the next word before transmission of the previous word is completed.

Note: Automatic waits are also canceled by writing to a DBR not being used as a transmit data buffer register; therefore, during SIO do not use such DBR for other applications. For example, when 3 words are transmitted, do not use the DBR of the remained 5 words.

When an external clock is used, the data must be written to the data buffer register before shifting next data. Thus, the transfer speed is determined by the maximum delay time from the generation of the interrupt request to writing of the data to the data buffer register by the interrupt service program.

The transmission is ended by clearing SIOCR1<SIOS> to “0” or setting SIOCR1<SIOINH> to “1” in buffer empty interrupt service program.

SIOCR1<SIOS> is cleared, the operation will end after all bits of words are transmitted.

That the transmission has ended can be determined from the status of SIOSR<SIOF> because SIOSR<SIOF> is cleared to “0” when a transfer is completed.

When SIOCR1<SIOINH> is set, the transmission is immediately ended and SIOSR<SIOF> is cleared to “0”.

When an external clock is used, it is also necessary to clear SIOCR1<SIOS> to “0” before shifting the next data; If SIOCR1<SIOS> is not cleared before shift out, dummy data will be transmitted and the operation will end.

If it is necessary to change the number of words, SIOCR1<SIOS> should be cleared to “0”, then SIOCR2<BUF> must be rewritten after confirming that SIOSR<SIOF> has been cleared to “0”.

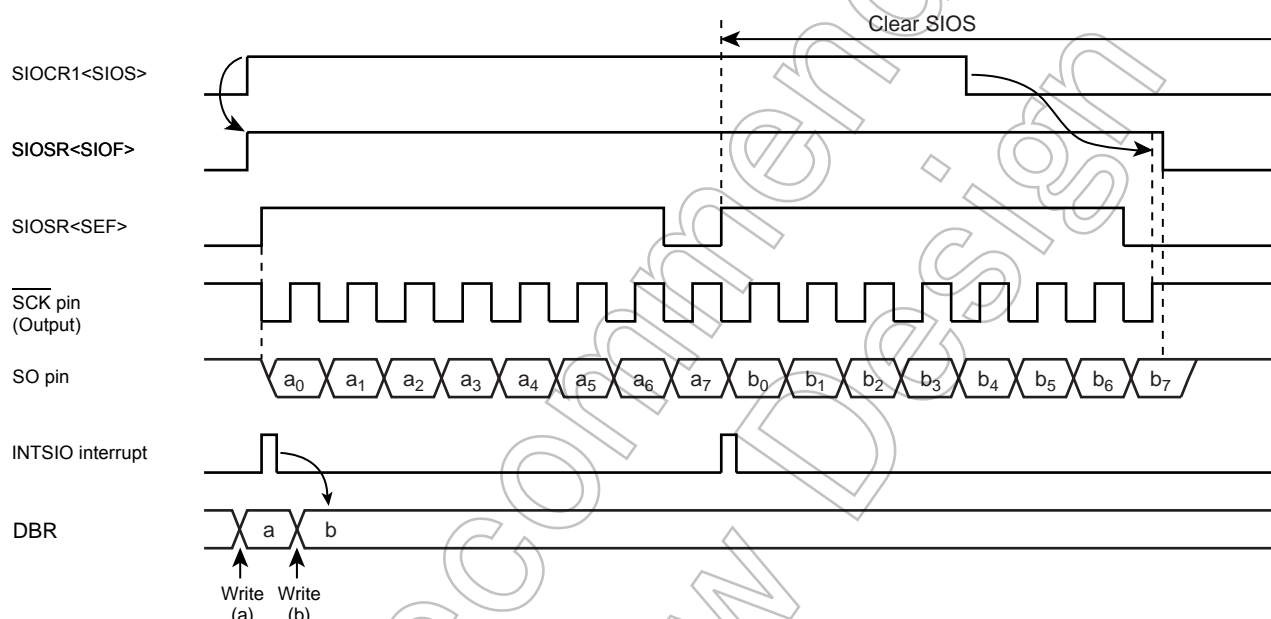


Figure 10-7 Transfer Mode (Example: 8bit, 1word transfer, Internal clock)

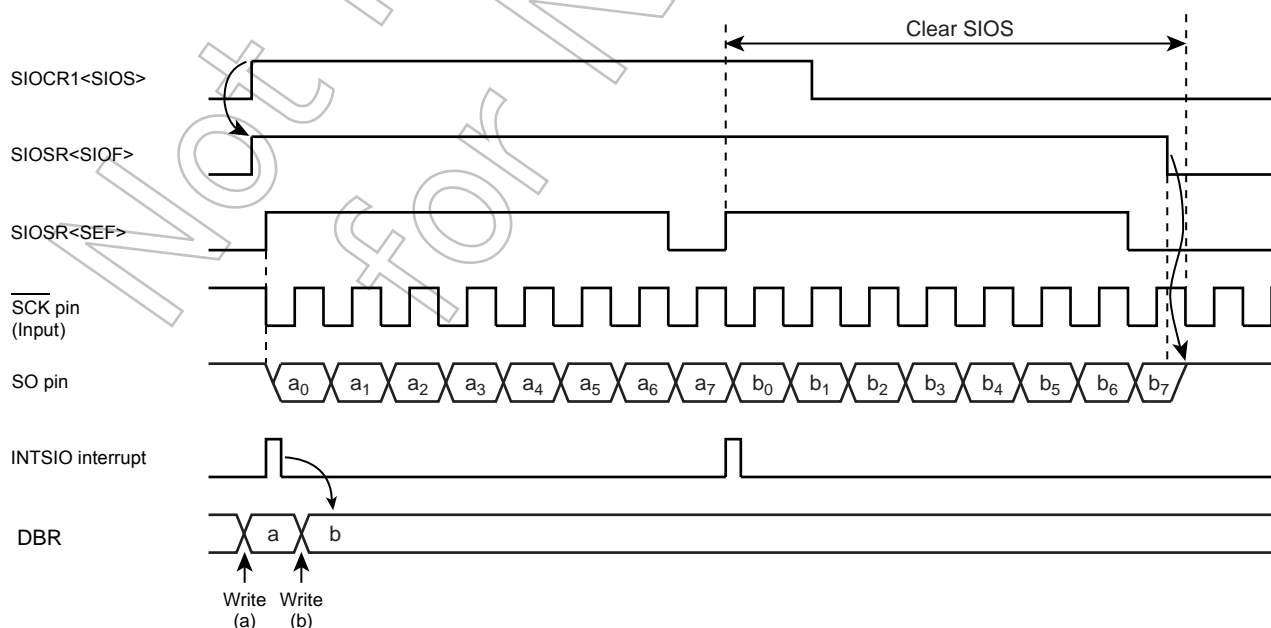


Figure 10-8 Transfer Mode (Example: 8bit, 1word transfer, External clock)

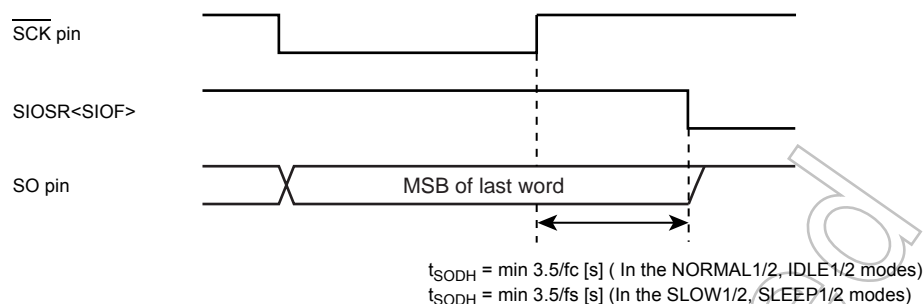


Figure 10-9 Transmitted Data Hold Time at End of Transfer

10.6.2 4-bit and 8-bit receive modes

After setting the control registers to the receive mode, set SIOCR1<SIOS> to “1” to enable receiving. The data are then transferred to the shift register via the SI pin in synchronous with the serial clock. When one word of data has been received, it is transferred from the shift register to the data buffer register (DBR). When the number of words specified with the SIOCR2<BUF> has been received, an INTSIO (Buffer full) interrupt is generated to request that these data be read out. The data are then read from the data buffer registers by the interrupt service program.

When the internal clock is used, and the previous data are not read from the data buffer register before the next data are received, the serial clock will stop and an automatic-wait will be initiated until the data are read. A wait will not be initiated if even one data word has been read.

Note: Waits are also canceled by reading a DBR not being used as a received data buffer register is read; therefore, during SIO do not use such DBR for other applications.

When an external clock is used, the shift operation is synchronized with the external clock; therefore, the previous data are read before the next data are transferred to the data buffer register. If the previous data have not been read, the next data will not be transferred to the data buffer register and the receiving of any more data will be canceled. When an external clock is used, the maximum transfer speed is determined by the delay between the time when the interrupt request is generated and when the data received have been read.

The receiving is ended by clearing SIOCR1<SIOS> to “0” or setting SIOCR1<SIOINH> to “1” in buffer full interrupt service program.

When SIOCR1<SIOS> is cleared, the current data are transferred to the buffer. After SIOCR1<SIOS> cleared, the receiving is ended at the time that the final bit of the data has been received. That the receiving has ended can be determined from the status of SIOSR<SIOF>. SIOSR<SIOF> is cleared to “0” when the receiving is ended. After confirmed the receiving termination, the final receiving data is read. When SIOCR1<SIOINH> is set, the receiving is immediately ended and SIOSR<SIOF> is cleared to “0”. (The received data is ignored, and it is not required to be read out.)

If it is necessary to change the number of words in external clock operation, SIOCR1<SIOS> should be cleared to “0” then SIOCR2<BUF> must be rewritten after confirming that SIOSR<SIOF> has been cleared to “0”. If it is necessary to change the number of words in internal clock, during automatic-wait operation which occurs after completion of data receiving, SIOCR2<BUF> must be rewritten before the received data is read out.

Note: The buffer contents are lost when the transfer mode is switched. If it should become necessary to switch the transfer mode, end receiving by clearing SIOCR1<SIOS> to “0”, read the last data and then switch the transfer mode.

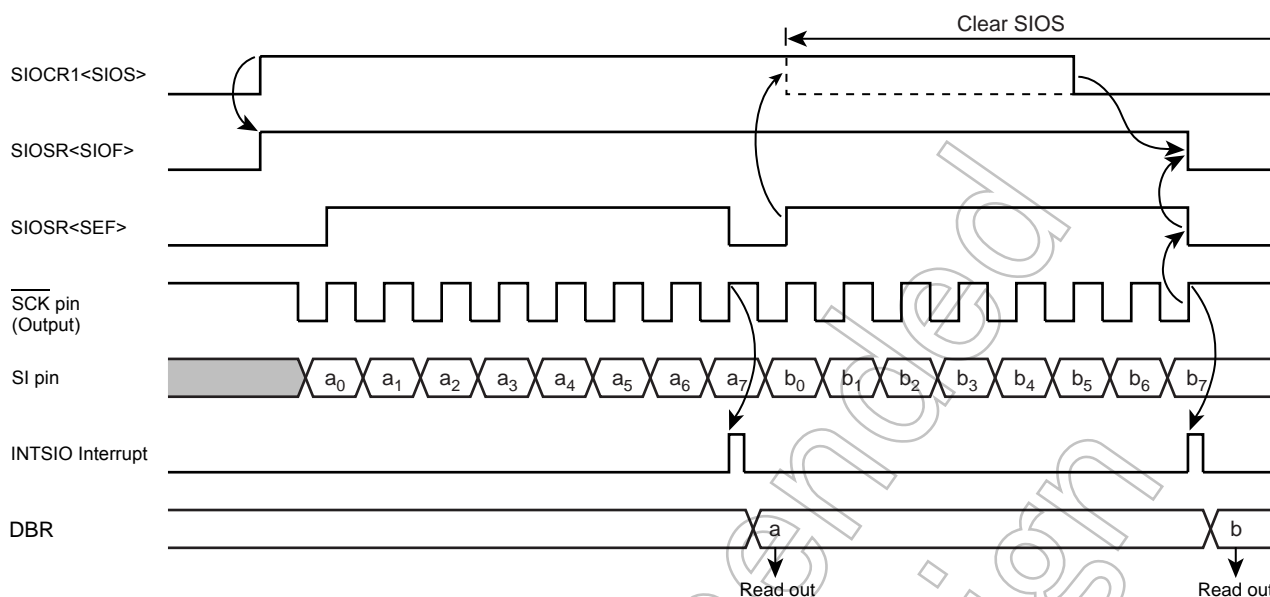


Figure 10-10 Receive Mode (Example: 8bit, 1word transfer, Internal clock)

10.6.3 8-bit transfer / receive mode

After setting the SIO control register to the 8-bit transmit/receive mode, write the data to be transmitted first to the data buffer registers (DBR). After that, enable the transmit/receive by setting SIOCR1<SIOS> to "1". When transmitting, the data are output from the SO pin at leading edges of the serial clock. When receiving, the data are input to the SI pin at the trailing edges of the serial clock. When the all receive is enabled, 8-bit data are transferred from the shift register to the data buffer register. An INTSIO interrupt is generated when the number of data words specified with the SIOCR2<BUF> has been transferred. Usually, read the receive data from the buffer register in the interrupt service. The data buffer register is used for both transmitting and receiving; therefore, always write the data to be transmitted after reading the all received data.

When the internal clock is used, a wait is initiated until the received data are read and the next transfer data are written. A wait will not be initiated if even one transfer data word has been written.

When an external clock is used, the shift operation is synchronized with the external clock; therefore, it is necessary to read the received data and write the data to be transmitted next before starting the next shift operation. When an external clock is used, the transfer speed is determined by the maximum delay between generation of an interrupt request and the received data are read and the data to be transmitted next are written.

The transmit/receive operation is ended by clearing SIOCR1<SIOS> to "0" or setting SIOCR1<SIOINH> to "1" in INTSIO interrupt service program.

When SIOCR1<SIOS> is cleared, the current data are transferred to the buffer. After SIOCR1<SIOS> cleared, the transmitting/receiving is ended at the time that the final bit of the data has been transmitted.

That the transmitting/receiving has ended can be determined from the status of SIOSR<SIOF>. SIOSR<SIOF> is cleared to "0" when the transmitting/receiving is ended.

When SIOCR1<SIOINH> is set, the transmit/receive operation is immediately ended and SIOSR<SIOF> is cleared to "0".

If it is necessary to change the number of words in external clock operation, SIOCR1<SIOS> should be cleared to "0", then SIOCR2<BUF> must be rewritten after confirming that SIOSR<SIOF> has been cleared to "0".

If it is necessary to change the number of words in internal clock, during automatic-wait operation which occurs after completion of transmit/receive operation, SIOCR2<BUF> must be rewritten before reading and writing of the receive/transmit data.

Note: The buffer contents are lost when the transfer mode is switched. If it should become necessary to switch the transfer mode, end receiving by clearing SIOCR1<SIOS> to "0", read the last data and then switch the transfer mode.

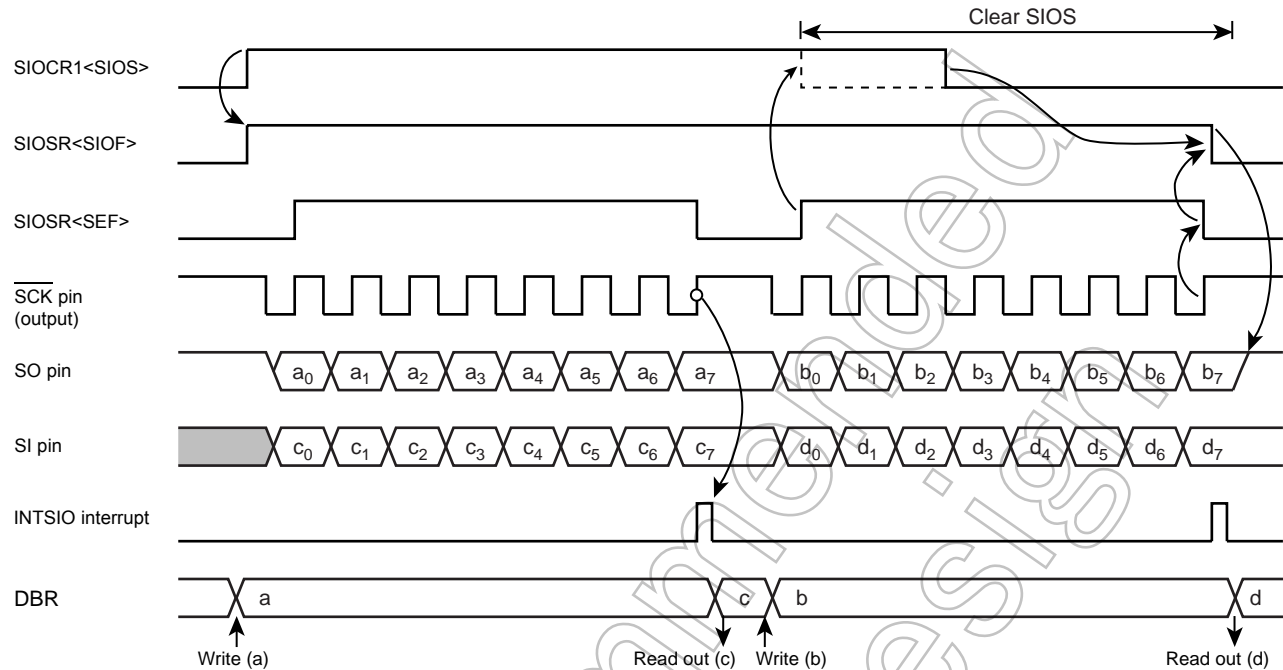


Figure 10-11 Transfer / Receive Mode (Example: 8bit, 1word transfer, Internal clock)

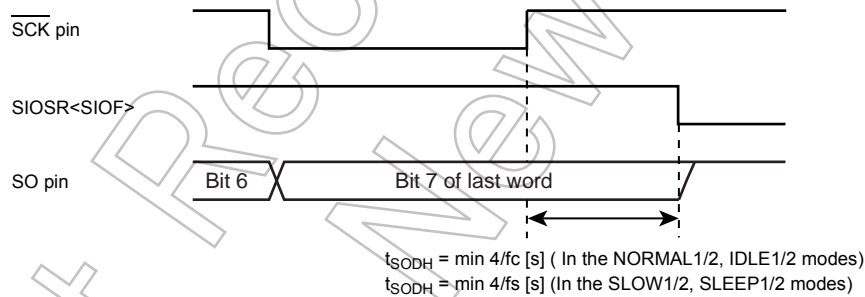


Figure 10-12 Transmitted Data Hold Time at End of Transfer / Receive

11. 8-Bit AD Converter (ADC)

The TMP86P820FG have a 8-bit successive approximation type AD converter.

11.1 Configuration

The circuit configuration of the 8-bit AD converter is shown in Figure 11-1.

It consists of control registers ADCCR1 and ADCCR2, converted value registers ADCDR1 and ADCDR2, a DA converter, a sample-and-hold circuit, a comparator, and a successive comparison circuit.

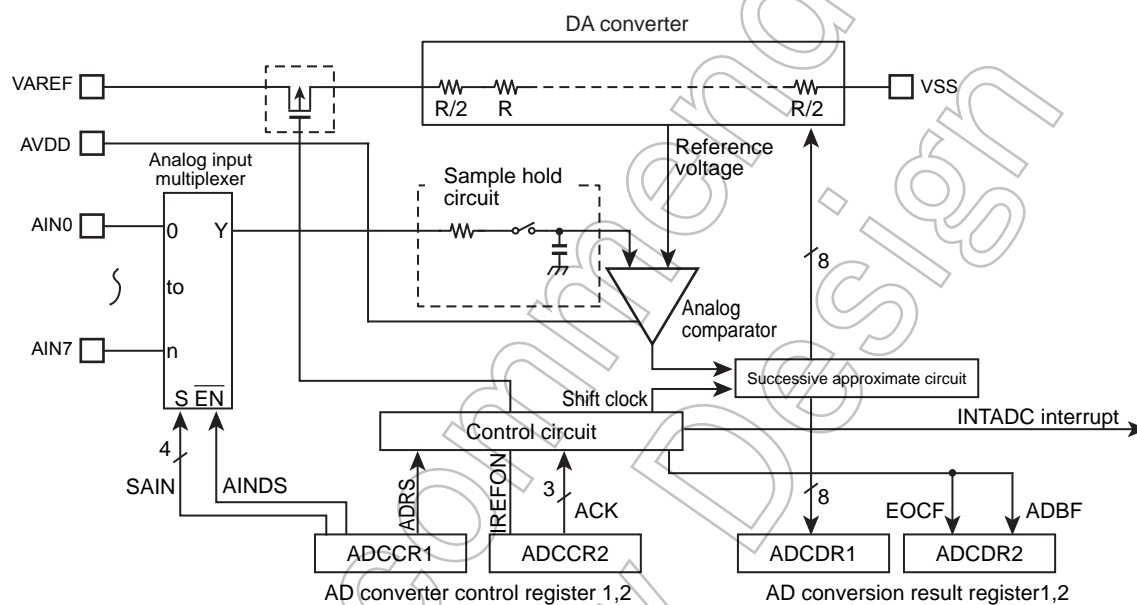


Figure 11-1 8-bit AD Converter (ADC)

11.2 Control

The AD converter consists of the following four registers:

1. AD converter control register 1 (ADCCR1)
This register selects the analog channels in which to perform AD conversion and controls the AD converter as it starts operating.
2. AD converter control register 2 (ADCCR2)
This register selects the AD conversion time and controls the connection of the DA converter (ladder resistor network).
3. AD converted value register (ADCDR1)
This register is used to store the digital value after being converted by the AD converter.
4. AD converted value register (ADCDR2)
This register monitors the operating status of the AD converter.

AD Converter Control Register 1

ADCCR1 (000EH)	7	6	5	4	3	2	1	0	
	ADRS	"0"	"1"	AINDS	SAIN				(Initial value: 0001 0000)

ADRS	AD conversion start	0: Start 1: Start	R/W
AINDS	Analog input control	0: Analog input enable 1: Analog input disable	
SAIN	Analog input channel select	0000: AIN0 0001: AIN1 0010: AIN2 0011: AIN3 0100: AIN4 0101: AIN5 0110: AIN6 0111: AIN7 1000: Reserved 1001: Reserved 1010: Reserved 1011: Reserved 1100: Reserved 1101: Reserved 1110: Reserved 1111: Reserved	

- Note 1: Select analog input when AD converter stops (ADCDR2<ADBF> = "0").
- Note 2: When the analog input is all use disabling, the ADCCR1<AINDS> should be set to "1".
- Note 3: During conversion, do not perform output instruction to maintain a precision for all of the pins. And port near to analog input, do not input intense signaling of change.
- Note 4: The ADRS is automatically cleared to "0" after starting conversion.
- Note 5: Do not set ADCCR1<ADRS> newly again during AD conversion. Before setting ADCCR1<ADRS> newly again, check ADCDR2<EOCF> to see that the conversion is completed or wait until the interrupt signal (INTADC) is generated (e.g., interrupt handling routine).
- Note 6: After STOP or SLOW/SLEEP mode are started, AD converter control register 1 (ADCCR1) is all initialized and no data can be written in this register. Therefore, to use AD converter again, set the ADCCR1 newly after returning to NORMAL1 or NORMAL2 mode.
- Note 7: Although ADCCR1<SAIN> is initialized to "Reserved value" after reset, set the suitable analog input channel when using AD converter.
- Note 8: Always set bit 5 in ADCCR1 to "1" and set bit 6 in ADCCR1 to "0".

AD Converter Control Register 2

ADCCR2 (000FH)	7	6	5	4	3	2	1	0	
			IREFON	"1"		ACK		"0"	(Initial value: **0* 000*)

IREFON	DA converter (ladder resistor) connection control	0: Connected only during AD conversion 1: Always connected	R/W
ACK	AD conversion time select	000: 39/fc 001: Reserved 010: 78/fc 011: 156/fc 100: 312/fc 101: 624/fc 110: 1248/fc 111: Reserved	R/W

Note 1: Always set bit 0 in ADCCR2 to "0" and set bit 4 in ADCCR2 to "1".

Note 2: When a read instruction for ADCCR2, bit 6 to 7 in ADCCR2 read in as undefined data.

Note 3: After STOP or SLOW/SLEEP mode are started, AD converter control register 2 (ADCCR2) is all initialized and no data can be written in this register. Therefore, to use AD converter again, set the ADCCR2 newly after returning to NORMAL1 or NORMAL2 mode.

Table 11-1 Conversion Time according to ACK Setting and Frequency

Condition ACK	Conversion time'	16MHz	8MHz	4 MHz	2 MHz	10MHz	5 MHz	2.5 MHz
000	39/fc	-	-	-	19.5 μ s	-	-	15.6 μ s
001	Reserved							
010	78/fc	-	-	19.5 μ s	39.0 μ s	-	15.6 μ s	31.2 μ s
011	156/fc	-	19.5 μ s	39.0 μ s	78.0 μ s	15.6 μ s	31.2 μ s	62.4 μ s
100	312/fc	19.5 μ s	39.0 μ s	78.0 μ s	156.0 μ s	31.2 μ s	62.4 μ s	124.8 μ s
101	624/fc	39.0 μ s	78.0 μ s	156.0 μ s	-	62.4 μ s	124.8 μ s	-
110	1248/fc	78.0 μ s	156.0 μ s	-	-	124.8 μ s	-	-
111	Reserved							

Note 1: Settings for "-" in the above table are inhibited.

Note 2: Set conversion time by Analog Reference Voltage (V_{AREF}) as follows.

- $V_{AREF} = 4.5$ to 5.5 V (15.6 μ s or more)
- $V_{AREF} = 2.7$ to 5.5 V (31.2 μ s or more)
- $V_{AREF} = 1.8$ to 5.5 V (124.8 μ s or more)

AD Conversion Result Register

ADCDR1 (0020H)	7	6	5	4	3	2	1	0	
	AD07	AD06	AD05	AD04	AD03	AD02	AD01	AD00	(Initial value: 0000 0000)

AD Conversion Result Register

ADCDR2 (0021H)	7	6	5	4	3	2	1	0	
			EOCF	ADBF					(Initial value: **00 ****)

EOCF	AD conversion end flag	0: Before or during conversion 1: Conversion completed	Read only
ADBF	AD conversion busy flag	0: During stop of AD conversion 1: During AD conversion	

Note 1: The ADCDR2<EOCF> is cleared to "0" when reading the ADCDR1.

Therefore, the AD conversion result should be read to ADCDR2 more first than ADCDR1.

Note 2: ADCDR2<ADBF> is set to "1" when AD conversion starts and cleared to "0" when the AD conversion is finished. It also is cleared upon entering STOP or SLOW mode.

Note 3: If a read instruction is executed for ADCDR2, read data of bits 7, 6 and 3 to 0 are unstable.

11.3 Function

11.3.1 AD Converter Operation

When ADCCR1<ADRS> is set to "1", AD conversion of the voltage at the analog input pin specified by ADCCR1<SAIN> is thereby started.

After completion of the AD conversion, the conversion result is stored in AD converted value registers (ADCDR1) and at the same time ADCDR2<EOCF> is set to "1", the AD conversion finished interrupt (INTADC) is generated.

ADCCR1<ADRS> is automatically cleared after AD conversion has started. Do not set ADCCR1<ADRS> newly again (restart) during AD conversion. Before setting ADRS newly again, check ADCDR<EOCF> to see that the conversion is completed or wait until the interrupt signal (INTADC) is generated (e.g., interrupt handling routine).

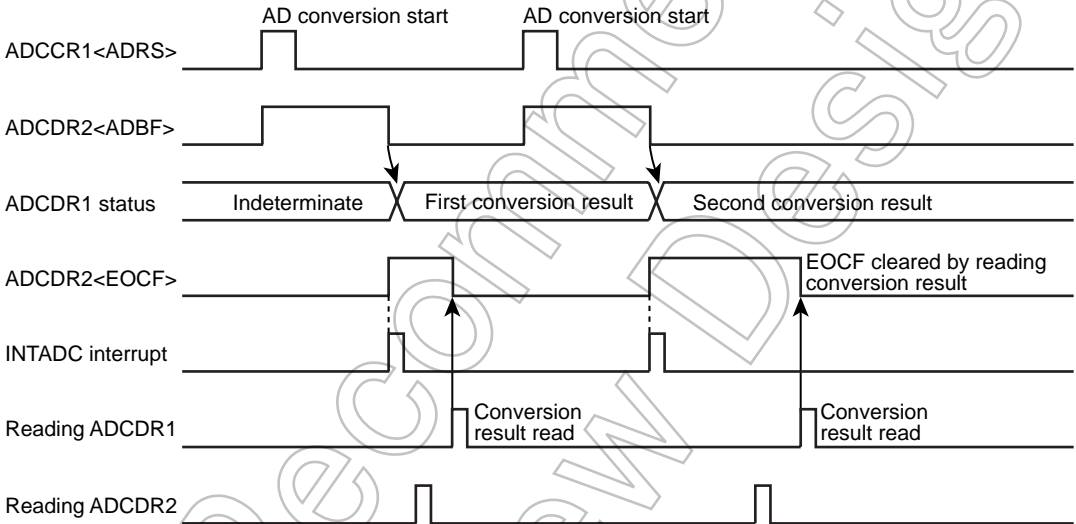


Figure 11-2 AD Converter Operation

11.3.2 AD Converter Operation

1. Set up the AD converter control register 1 (ADCCR1) as follows:
 - Choose the channel to AD convert using AD input channel select (SAIN).
 - Specify analog input enable for analog input control (AINDS).
2. Set up the AD converter control register 2 (ADCCR2) as follows:
 - Set the AD conversion time using AD conversion time (ACK). For details on how to set the conversion time, refer to Table 11-1.
 - Choose IREFON for DA converter control.
3. After setting up 1. and 2. above, set AD conversion start (ADRS) of AD converter control register 1 (ADCCR1) to "1".
4. After an elapse of the specified AD conversion time, the AD converted value is stored in AD converted value register 1 (ADCDR1) and the AD conversion finished flag (EOCF) of AD converted value register 2 (ADCDR2) is set to "1", upon which time AD conversion interrupt INTADC is generated.
5. EOCF is cleared to "0" by a read of the conversion result. However, if reconverted before a register read, although EOCF is cleared the previous conversion result is retained until the next conversion is completed.

Example :After selecting the conversion time of 19.5 μ s at 16 MHz and the analog input channel AIN3 pin, perform AD conversion once. After checking EOCF, read the converted value and store the 8-bit data in address 009FH on RAM.

```

; AIN SELECT
:
:
LD      (ADCCR1), 00100011B ; Select AIN3
LD      (ADCCR2), 11011000B ; Select conversion time (312/fc) and operation mode
; AD CONVERT START
SET      (ADCCR1), 7 ; ADRS = 1
SLOOP:  TEST      (ADCDR2), 5 ; EOCF = 1 ?
JRS      T, SLOOP
; RESULT DATA READ
LD      A, (ADCDR1)
LD      (9FH), A
```

11.3.3 STOP and SLOW Mode during AD Conversion

When the STOP or SLOW mode is entered forcibly during AD conversion, the AD convert operation is suspended and the AD converter is initialized (ADCCR1 and ADCCR2 are initialized to initial value.). Also, the conversion result is indeterminate. (Conversion results up to the previous operation are cleared, so be sure to read the conversion results before entering STOP or SLOW mode.) When restored from STOP or SLOW mode, AD conversion is not automatically restarted, so it is necessary to restart AD conversion. Note that since the analog reference voltage is automatically disconnected, there is no possibility of current flowing into the analog reference voltage.

11.3.4 Analog Input Voltage and AD Conversion Result

The analog input voltage is corresponded to the 8-bit digital value converted by the AD as shown in Figure 11-3.

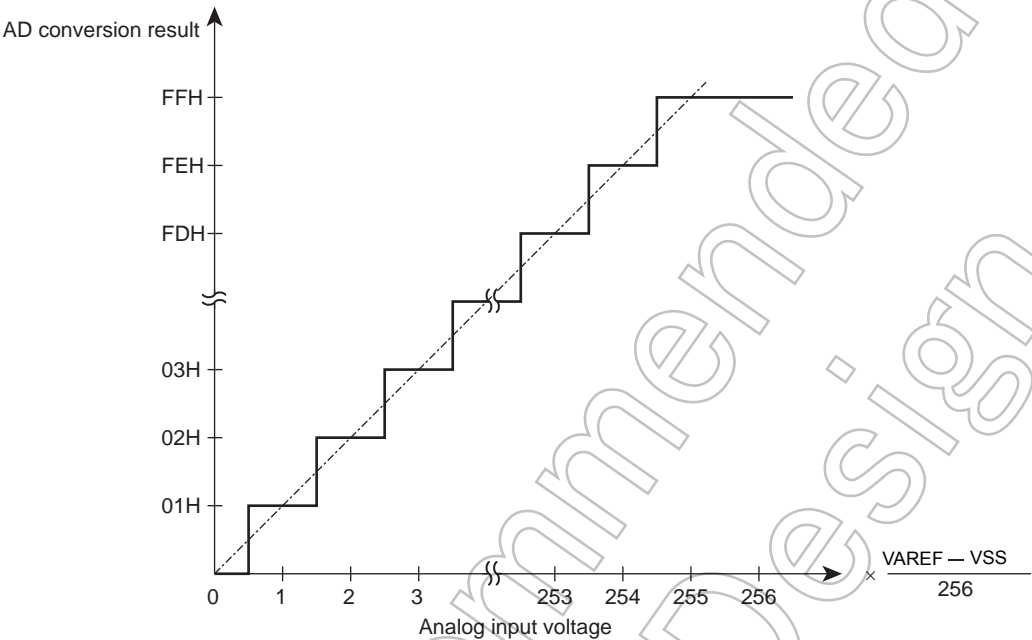


Figure 11-3 Analog Input Voltage and AD Conversion Result (typ.)

11.4 Precautions about AD Converter

11.4.1 Analog input pin voltage range

Make sure the analog input pins (AIN0 to AIN7) are used at voltages within VSS below VAREF. If any voltage outside this range is applied to one of the analog input pins, the converted value on that pin becomes uncertain. The other analog input pins also are affected by that.

11.4.2 Analog input shared pins

The analog input pins (AIN0 to AIN7) are shared with input/output ports. When using any of the analog inputs to execute AD conversion, do not execute input/output instructions for all other ports. This is necessary to prevent the accuracy of AD conversion from degrading. Not only these analog input shared pins, some other pins may also be affected by noise arising from input/output to and from adjacent pins.

11.4.3 Noise countermeasure

The internal equivalent circuit of the analog input pins is shown in Figure 11-4. The higher the output impedance of the analog input source, more easily they are susceptible to noise. Therefore, make sure the output impedance of the signal source in your design is 5 k Ω or less. Toshiba also recommends attaching a capacitor external to the chip.

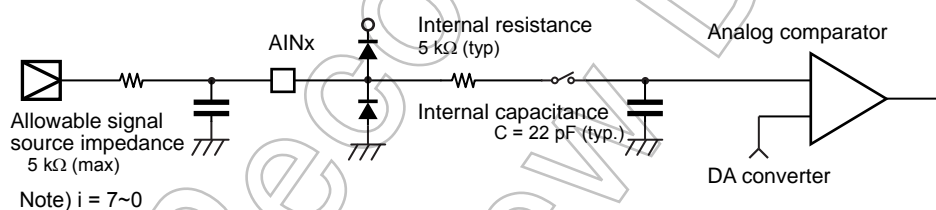


Figure 11-4 Analog Input Equivalent Circuit and Example of Input Pin Processing

Not Recommended
for New Design

Also, each level of the STOP2 to STOP5 pins can be confirmed by reading corresponding I/O port data register, check all STOP2 to STOP5 pins "H" that is enabled by STOPPCR before the STOP mode is started (Note2,3).

Note 1: When the STOP mode released by the edge release mode (SYSCR1<RELM> = "0"), inhibit input from STOP2 to STOP5 pins by Key-on Wakeup Control Register (STOPPCR) or must be set "H" level into STOP2 to STOP5 pins that are available input during STOP mode.

Note 2: When the $\overline{\text{STOP}}$ pin input is high or STOP2 to STOP5 pins input which is enabled by STOPPCR is low, executing an instruction which starts STOP mode will not place in STOP mode but instead will immediately start the release sequence (Warm up).

Note 3: The input circuit of Key-on Wakeup input and Port input is separated, so each input voltage threshold value is different. Therefore, a value comes from port input before STOP mode start may be different from a value which is detected by Key-on Wakeup input (Figure 12-2).

Note 4: $\overline{\text{STOP}}$ pin doesn't have the control register such as STOPPCR, so when STOP mode is released by STOP2 to STOP5 pins, $\overline{\text{STOP}}$ pin also should be used as STOP mode release function.

Note 5: In STOP mode, Key-on Wakeup pin which is enabled as input mode (for releasing STOP mode) by Key-on Wakeup Control Register (STOPPCR) may generate the penetration current, so the said pin must be disabled AD conversion input (analog voltage input).

Note 6: When the STOP mode is released by STOP2 to STOP5 pins, the level of $\overline{\text{STOP}}$ pin should hold "L" level (Figure 12-3).

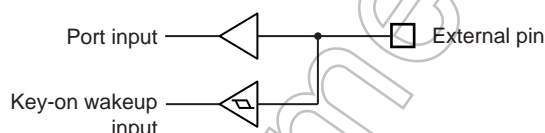


Figure 12-2 Key-on Wakeup Input and Port Input

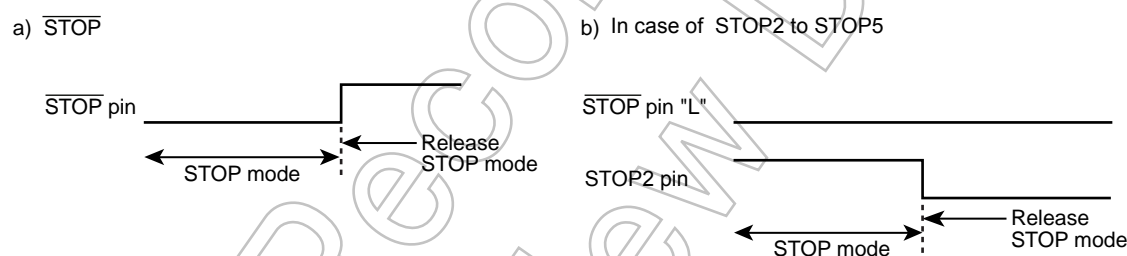


Figure 12-3 Priority of $\overline{\text{STOP}}$ pin and STOP2 to STOP5 pins

Table 12-1 Release level (edge) of STOP mode

Pin name	Release level (edge)	
	SYSCR1<RELM>="1" (Note2)	SYSCR1<RELM>="0"
$\overline{\text{STOP}}$	"H" level	Rising edge
STOP2	"L" level	Don't use (Note1)
STOP3	"L" level	Don't use (Note1)
STOP4	"L" level	Don't use (Note1)
STOP5	"L" level	Don't use (Note1)

13.2 Control

The LCD driver is controlled using the LCD control register (LCDCR). The LCD driver's display is enabled using the EDSP.

LCD Driver Control Register

LCDCR (0028H)	7	6	5	4	3	2	1	0	
	EDSP	BRES	VFSEL		DUTY		SLF		(Initial value: 0000 0000)

EDSP	LCD Display Control	0: Blanking 1: Enables LCD display (Blanking is released)			R/W	
BRES	Booster circuit control	0: Disable (use divider resistance) 1: Enable				
VFSEL	Selection of boost frequency		NORMAL 1/2, IDLE 1/2 mode			SLOW1/2, SLEEP0/1/2 mode
			DV7CK = 0	DV7CK = 1		
		00	$f_c/2^{13}$	$f_s/2^5$		$f_s/2^5$
		01	$f_c/2^{11}$	$f_s/2^3$		$f_s/2^3$
		10	$f_c/2^{10}$	$f_s/2^2$		$f_s/2^2$
		11	$f_c/2^9$	$f_c/2^9$		—
DUTY	Selection of driving methods	00: 1/4 Duty (1/3 Bias) 01: 1/3 Duty (1/3 Bias) 10: 1/2 Duty (1/2 Bias) 11: Static				
SLF	Selection of LCD frame frequency		NORMAL 1/2, IDLE 1/2 mode			SLOW1/2, SLEEP0/1/2 mode
			DV7CK = 0	DV7CK = 1		
		00	$f_c/2^{17}$	$f_s/2^9$		$f_s/2^9$
		01	$f_c/2^{16}$	$f_s/2^8$		$f_s/2^8$
		10	$f_c/2^{15}$	$f_c/2^{15}$		—
		11	$f_c/2^{13}$	$f_c/2^{13}$		—

Note 1: When <BRES>(Booster circuit control) is set to "0", $V_{DD} \geq V_3 \geq V_2 \geq V_1 \geq V_{SS}$ should be satisfied.

When <BRES> is set to "1", $5.5[V] \geq V_3 \geq V_{DD}$ should be satisfied.

If these conditions are not satisfied, it not only affects the quality of LCD display but also may damage the device due to over voltage of the port.

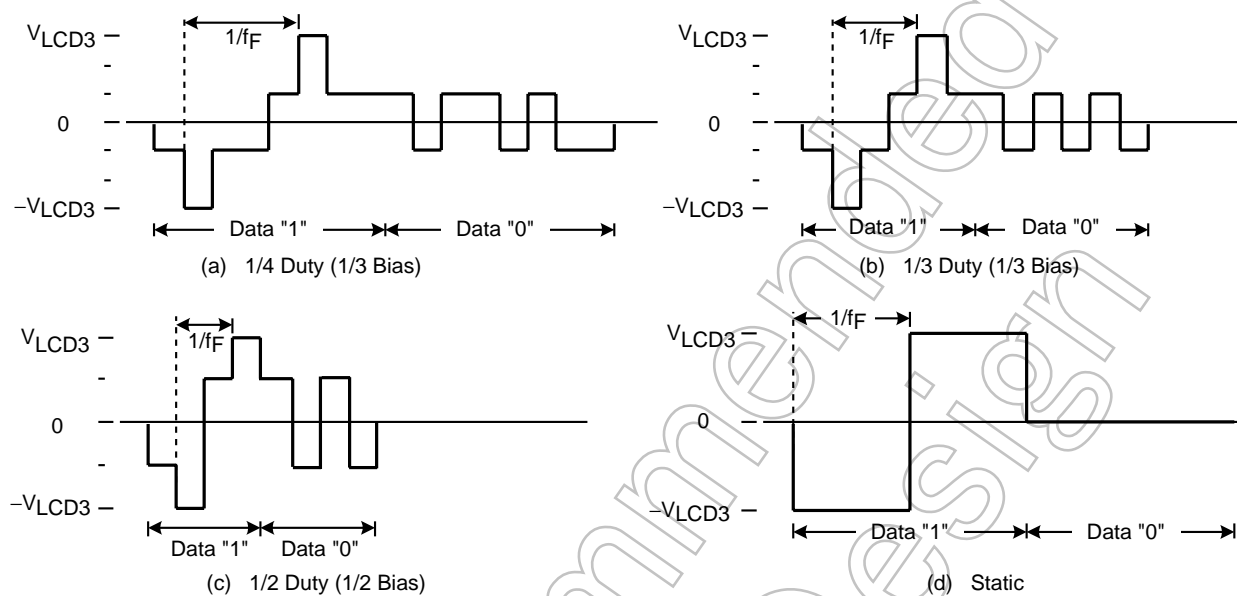
Note 2: When used as the booster circuit, bias should be composed to 1/3. Therefore, do not set LCDCR<DUTY> to "10" or "11" when the booster circuit is enable.

Note 3: Do not set SLF to "10" or "11" in SLOW1/2 modes.

Note 4: Do not set VFSEL to "11" SLOW1/2 modes.

13.2.1 LCD driving methods

As for LCD driving method, 4 types can be selected by LCDCR<DUTY>. The driving method is initialized in the initial program according to the LCD used.



Note 1: f_F : Frame frequency

Note 2: V_{LCD3} : LCD drive voltage

Figure 13-2 LCD Drive Waveform (COM-SEG pins)

13.2.2 Frame frequency

Frame frequency (f_F) is set according to driving method and base frequency as shown in the following Table 13-1. The base frequency is selected by LCDCR<SLF> according to the frequency f_c and f_s of the basic clock to be used.

Table 13-1 Setting of LCD Frame Frequency

(a) At the single clock mode. At the dual clock mode (DV7CK = 0).

SLF	Base frequency [Hz]	Frame frequency [Hz]			
		1/4 Duty	1/3 Duty	1/2 Duty	Static
00	$\frac{f_c}{2^{17}}$	$\frac{f_c}{2^{17}}$	$\frac{4}{3} \cdot \frac{f_c}{2^{17}}$	$\frac{4}{2} \cdot \frac{f_c}{2^{17}}$	$\frac{f_c}{2^{17}}$
	($f_c = 16$ MHz)	122	163	244	122
	($f_c = 8$ MHz)	61	81	122	61
01	$\frac{f_c}{2^{16}}$	$\frac{f_c}{2^{16}}$	$\frac{4}{3} \cdot \frac{f_c}{2^{16}}$	$\frac{4}{2} \cdot \frac{f_c}{2^{16}}$	$\frac{f_c}{2^{16}}$
	($f_c = 8$ MHz)	122	163	244	122
	($f_c = 4$ MHz)	61	81	122	61
10	$\frac{f_c}{2^{15}}$	$\frac{f_c}{2^{15}}$	$\frac{4}{3} \cdot \frac{f_c}{2^{15}}$	$\frac{4}{2} \cdot \frac{f_c}{2^{15}}$	$\frac{f_c}{2^{15}}$
	($f_c = 4$ MHz)	122	163	244	122
	($f_c = 2$ MHz)	61	81	122	61
11	$\frac{f_c}{2^{13}}$	$\frac{f_c}{2^{13}}$	$\frac{4}{3} \cdot \frac{f_c}{2^{13}}$	$\frac{4}{2} \cdot \frac{f_c}{2^{13}}$	$\frac{f_c}{2^{13}}$
	($f_c = 1$ MHz)	122	163	244	122

Note: f_c : High-frequency clock [Hz]

Table 13-2

(b) At the dual clock mode (DV7CK = 1 or SYSCK = 1)

SLF	Base frequency [Hz]	Frame frequency [Hz]			
		1/4 Duty	1/3 Duty	1/2 Duty	Static
00	$\frac{f_s}{2^9}$	$\frac{f_s}{2^9}$	$\frac{4}{3} \cdot \frac{f_s}{2^9}$	$\frac{4}{2} \cdot \frac{f_s}{2^9}$	$\frac{f_s}{2^9}$
	($f_s = 32.768$ kHz)	64	85	128	64
01	$\frac{f_s}{2^8}$	$\frac{f_s}{2^8}$	$\frac{4}{3} \cdot \frac{f_s}{2^8}$	$\frac{4}{2} \cdot \frac{f_s}{2^8}$	$\frac{f_s}{2^8}$
	($f_s = 32.768$ kHz)	128	171	256	128

Note: f_s : Low-frequency clock [Hz]

13.2.3 Driving method for LCD driver

In the TMP86P820FG, LCD driving voltages can be generated using either an internal booster circuit or an external resistor divider. This selection is made in LCDCR<BRES>.

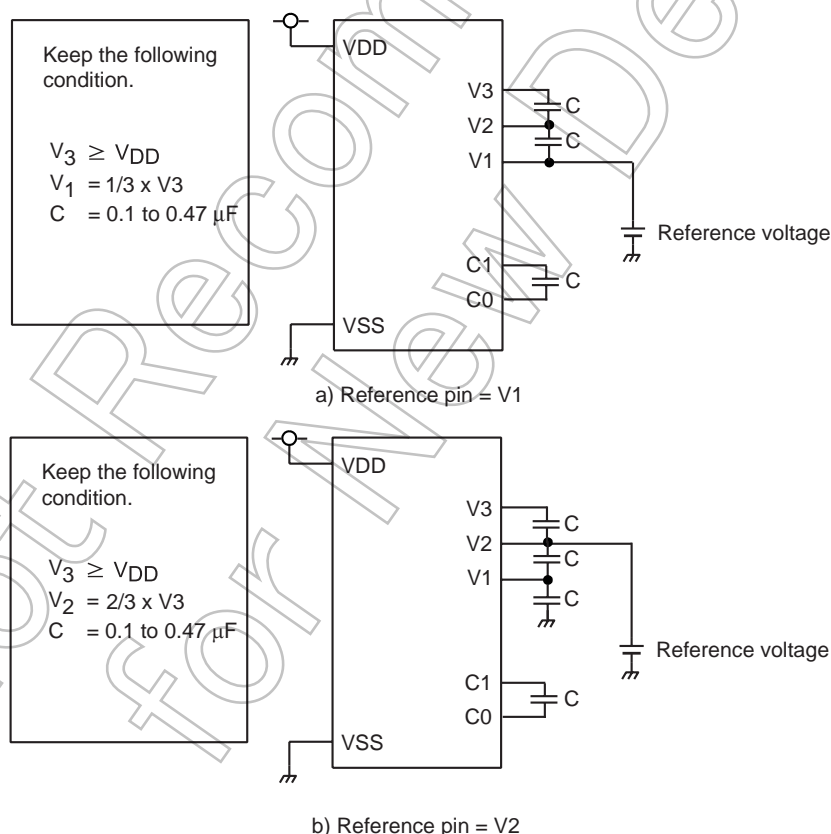
13.2.3.1 When using the booster circuit (LCDCR<BRES>="1")

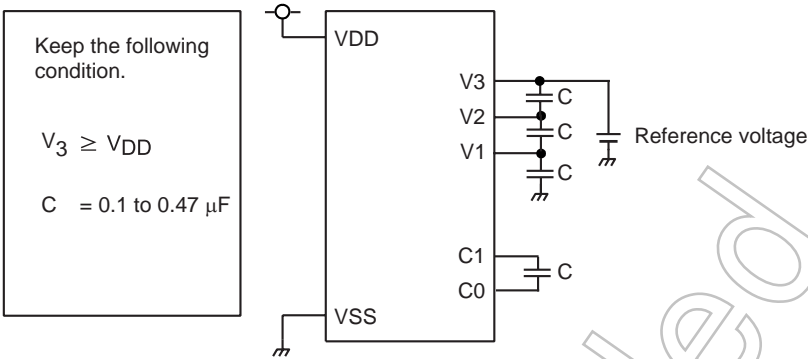
When the reference voltage is connected to the V1 pin, the booster circuit boosts the reference voltage twofold (V2) or threefold (V3) to generate the output voltages for segment/common signals. When the reference voltage is connected to the V2 pin, it is reduced to 1/2 (V1) or boosted to 3/2 (V3). When the reference voltage is connected to the V3 pin, it is reduced to 1/3 (V1) or 2/3 (V2).

LCDCR<VFSEL> is used to select the reference frequency in the booster circuit. The faster the boosting frequency, the higher the segment/common drive capability, but power consumption is increased. Conversely, the slower the boosting frequency, the lower the segment/common drive capability, but power consumption is reduced. If the drive capability is insufficient, the LCD may not be displayed clearly. Therefore, select an optimum boosting frequency for the LCD panel to be used.

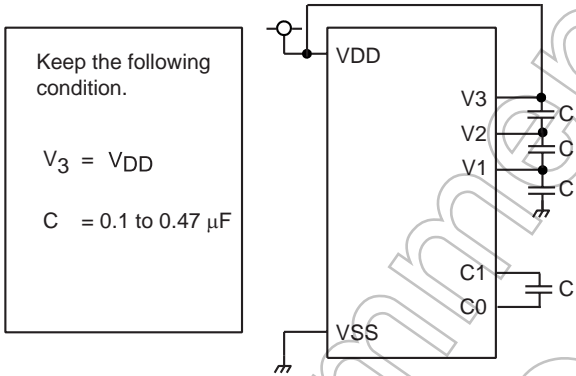
Table 13-3 shows the V3 pin current capacity and boosting frequency.

Note: When used as the booster circuit, bias should be composed to 1/3. Therefore, do not set LCDCR<DUTY> to "10" or "11" when the booster circuit is enable (LCDCR<BRES>="1").





c) Reference pin = V3



d) Reference pin = V3

- Note 1: When the TMP86P820FG uses the booster circuit to drive the LCD, the power supply and capacitor for the booster circuit should be connected as shown above.
- Note 2: When the reference voltage is connected to a pin other than V1, add a capacitor between V1 and GND.
- Note 3: The connection examples shown above are different from those shown in the datasheets of the previous version. Since the above connection method enhances the boosting characteristics, it is recommended that new boards be designed using the above connection method. (Using the existing connection method does not affect LCD display.)

Figure 13-3 Connection Examples When Using the Booster Circuit (LCDCR<BRES> = "1")

Table 13-3 V3 Pin Current Capacity and Boosting Frequency (typ.)

VFSEL	Boosting frequency	fc = 16 MHz	fc = 8 MHz	fc = 4 MHz	fc = 32.768 MHz
00	fc/2 ¹³ or fs/2 ⁵	-37 mV/ μ A	-80 mV/ μ A	-138 mV/ μ A	-76 mV/ μ A
01	fc/2 ¹¹ or fs/2 ³	-19 mV/ μ A	-24 mV/ μ A	-37 mV/ μ A	-23 mV/ μ A
10	fc/2 ¹⁰ or fs/2 ²	-17 mV/ μ A	-19 mV/ μ A	-24 mV/ μ A	-18 mV/ μ A
11	fc/2 ⁹	-16 mV/ μ A	-17 mV/ μ A	-19 mV/ μ A	-

- Note 1: The current capacity is the amount of voltage that falls per 1 μ A.
- Note 2: The boosting frequency should be selected depending on your LCD panel.
- Note 3: For the reference pin V1 or V2, a current capacity ten times larger than the above is recommended to ensure stable operation.
- For example, when the boosting frequency is fc/2⁹ (at fc = 8 MHz), -1.7 mV/ μ A or more is recommended for the current capacity of the reference pin V1.

13.2.3.2 When using an external resistor divider (LCDCR<BRES>="0")

When an external resistor divider is used, the voltage of an external power supply is divided and input on V1, V2, and V3 to generate the output voltages for segment/common signals.

The smaller the external resistor value, the higher the segment/common drive capability, but power consumption is increased. Conversely, the larger the external resistor value, the lower the segment/common drive capability, but power consumption is reduced. If the drive capability is insufficient, the LCD may not be displayed clearly. Therefore, select an optimum resistor value for the LCD panel to be used.

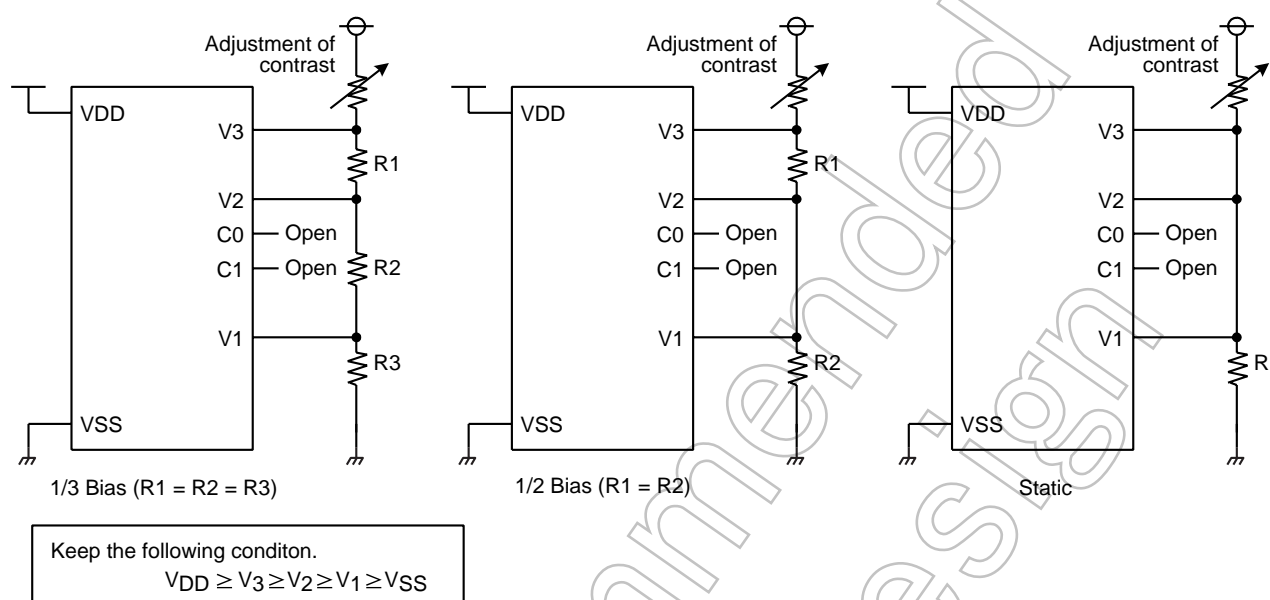


Figure 13-4 Connection Examples When Using an External Resistor Divider
(LCDCR<BRES> = "0")

13.3 LCD Display Operation

13.3.1 Display data setting

Display data is stored to the display data area (assigned to address 0F80H to 0F8FH, 16bytes) in the DBR. The display data which are stored in the display data area is automatically read out and sent to the LCD driver by the hardware. The LCD driver generates the segment signal and common signal according to the display data and driving method. Therefore, display patterns can be changed by only over writing the contents of display data area by the program. Table 13-5 shows the correspondence between the display data area and SEG/COM pins.

LCD light when display data is "1" and turn off when "0". According to the driving method of LCD, the number of pixels which can be driven becomes different, and the number of bits in the display data area which is used to store display data also becomes different.

Therefore, the bits which are not used to store display data as well as the data buffer which corresponds to the addresses not connected to LCD can be used to store general user process data (see Table 13-4).

Note: The display data memory contents become unstable when the power supply is turned on; therefore, the display data memory should be initialized by an initiation routine.

Table 13-4 Driving Method and Bit for Display Data

Driving methods	Bit 7/3	Bit 6/2	Bit 5/1	Bit 4/0
1/4 Duty	COM3	COM2	COM1	COM0
1/3 Duty	—	COM2	COM1	COM0
1/2 Duty	—	—	COM1	COM0
Static	—	—	—	COM0

Note: –: This bit is not used for display data

Table 13-5 LCD Display Data Area (DBR)

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0F80H	SEG1				SEG0			
0F81H	SEG3				SEG2			
0F82H	SEG5				SEG4			
0F83H	SEG7				SEG6			
0F84H	SEG9				SEG8			
0F85H	SEG11				SEG10			
0F86H	SEG13				SEG12			
0F87H	SEG15				SEG14			
0F88H	SEG17				SEG16			
0F89H	SEG19				SEG18			
0F8AH	SEG21				SEG20			
0F8BH	SEG23				SEG22			
0F8CH	SEG25				SEG24			
0F8DH	SEG27				SEG26			
0F8EH	SEG29				SEG28			
0F8FH	SEG31				SEG30			
	COM3	COM2	COM1	COM0	COM3	COM2	COM1	COM0

13.3.2 Blanking

Blanking is enabled when EDSP is cleared to “0”.

Blanking turns off LCD through outputting a GND level to SEG/COM pin.

When in STOP mode, EDSP is cleared to “0” and automatically blanked. To redisplay ICD after exiting STOP mode, it is necessary to set EDSP back to “1”.

Note: During reset, the LCD segment outputs and LCD common outputs are fixed “0” level. But the multiplex terminal of input/output port and LCD segment output becomes high impedance. Therefore, when the reset input is long remarkably, ghost problem may appear in LCD display.

13.4 Control Method of LCD Driver

13.4.1 Initial setting

Figure 13-5 shows the flowchart of initialization.

Example : To operate a 1/4 duty LCD of 32 segments × 4 com-mons at frame frequency $fc/2^{16}$ [Hz], and booster frequency $fc/2^{13}$ [Hz]

LD	(LCDCR), 01000001B	; Sets LCD driving method and frame frequency. Boost frequency
LD	(P*LCR), 0FFH	; Sets segment output control register. (*; Port No.)
:	:	
:	:	; Sets the initial value of display data.
LD	(LCDCR), 11000001B	; Display enable

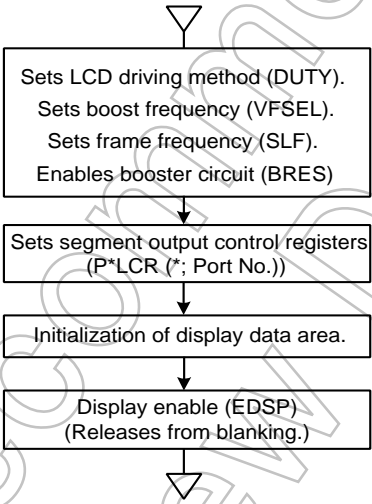


Figure 13-5 Initial Setting of LCD Driver

13.4.2 Store of display data

Generally, display data are prepared as fixed data in program memory (ROM) and stored in display data area by load command.

Example :To display using 1/4 duty LCD a numerical value which corresponds to the LCD data stored in data memory at address 80H (when pins COM and SEG are connected to LCD as in Figure 13-6), display data become as shown in Table 13-6.

```

LD      A, (80H)
ADD     A, TABLE-$-7
LD      HL, 0F80H
LD      W, (PC + A)
LD      (HL), W
RET

TABLE:  DB      11011111B, 00000110B,
              11100011B, 10100111B,
              00110110B, 10110101B,
              11110101B, 00010111B,
              11110111B, 10110111B

```

Note:DB is a byte data difinition instruction.

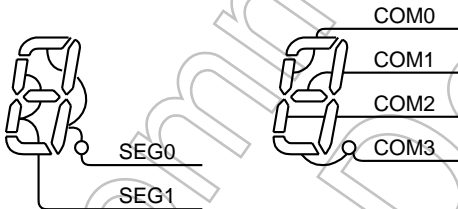


Figure 13-6 Example of COM, SEG Pin Connection (1/4 Duty)

Table 13-6 Example of Display Data (1/4 Duty)

No.	display	Display data	No.	display	Display data
0		11011111	5		10110101
1		00000110	6		11110101
2		11100011	7		00000111
3		10100111	8		11110111
4		00110110	9		10110111

Example 2: Table 13-6 shows an example of display data which are displayed using 1/2 duty LCD in the same way as Table 13-7. The connection between pins COM and SEG are the same as shown in Figure 13-7.

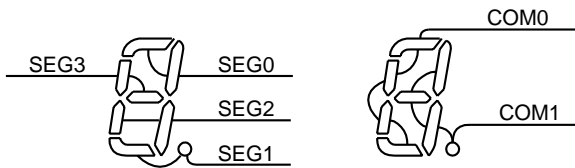


Figure 13-7 Example of COM, SEG Pin Connection

Table 13-7 Example of Display Data (1/2 Duty)

Number	Display data		Number	Display data	
	High order address	Low order address		High order address	Low order address
0	**01**11	**01**11	5	**11**10	**01**01
1	**00**10	**00**10	6	**11**11	**01**01
2	**10**01	**01**11	7	**01**10	**00**11
3	**10**10	**01**11	8	**11**11	**01**11
4	**11**10	**00**10	9	**11**10	**01**11

Note: *: Don't care

13.4.3 Example of LCD drive output

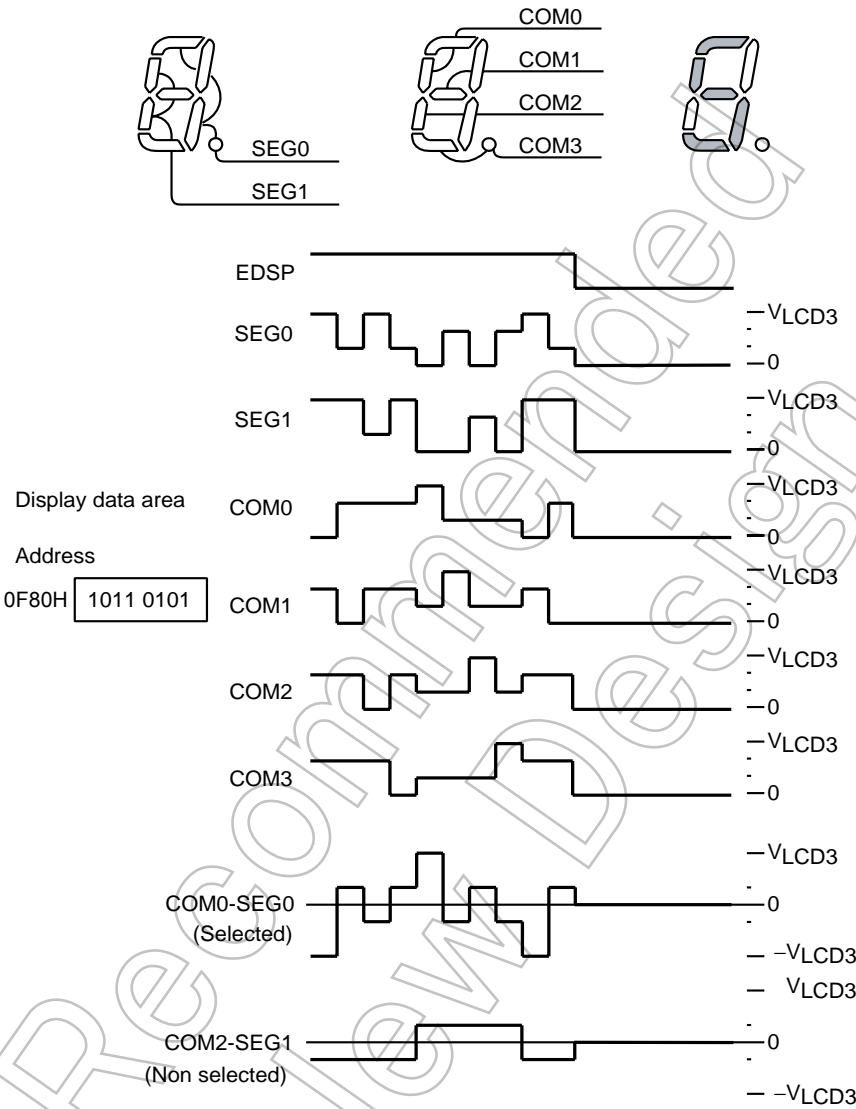


Figure 13-8 1/4 Duty (1/3 bias) Drive

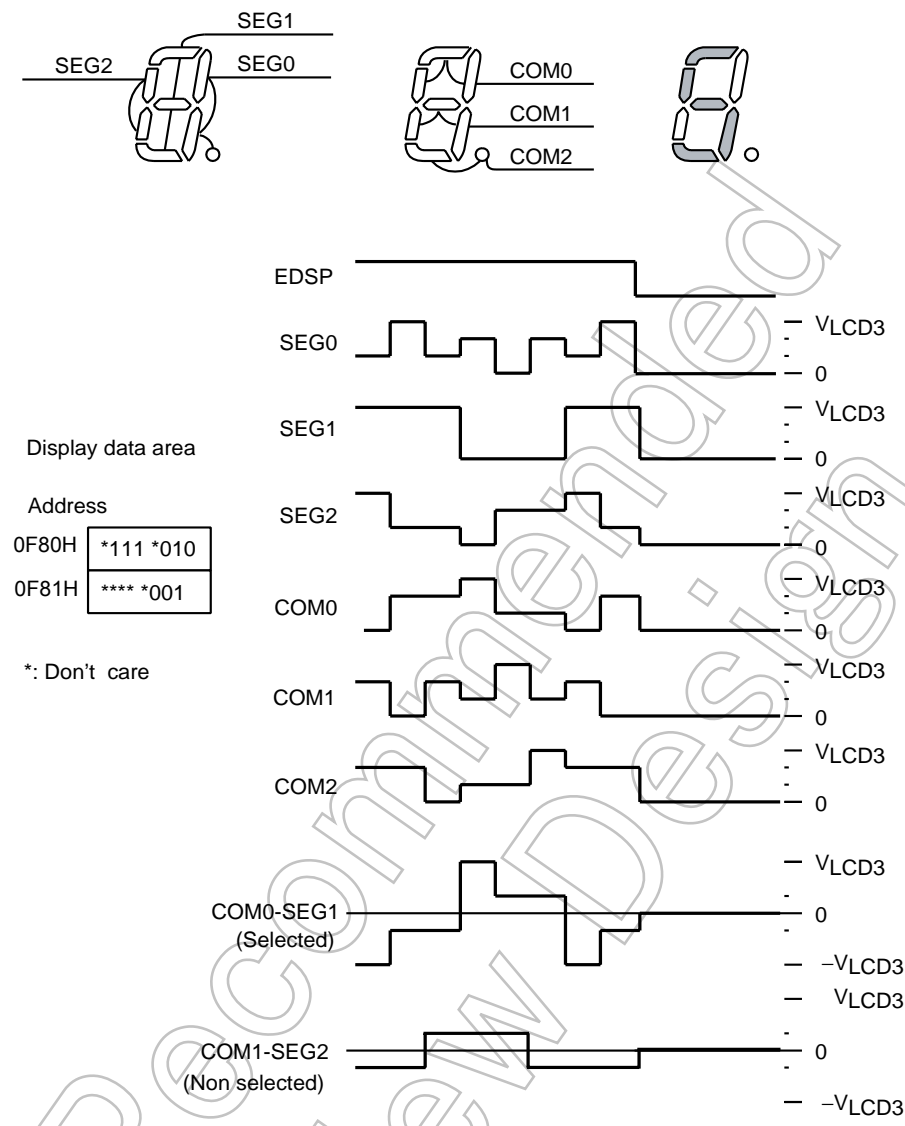


Figure 13-9 1/3 Duty (1/3 bias) Drive

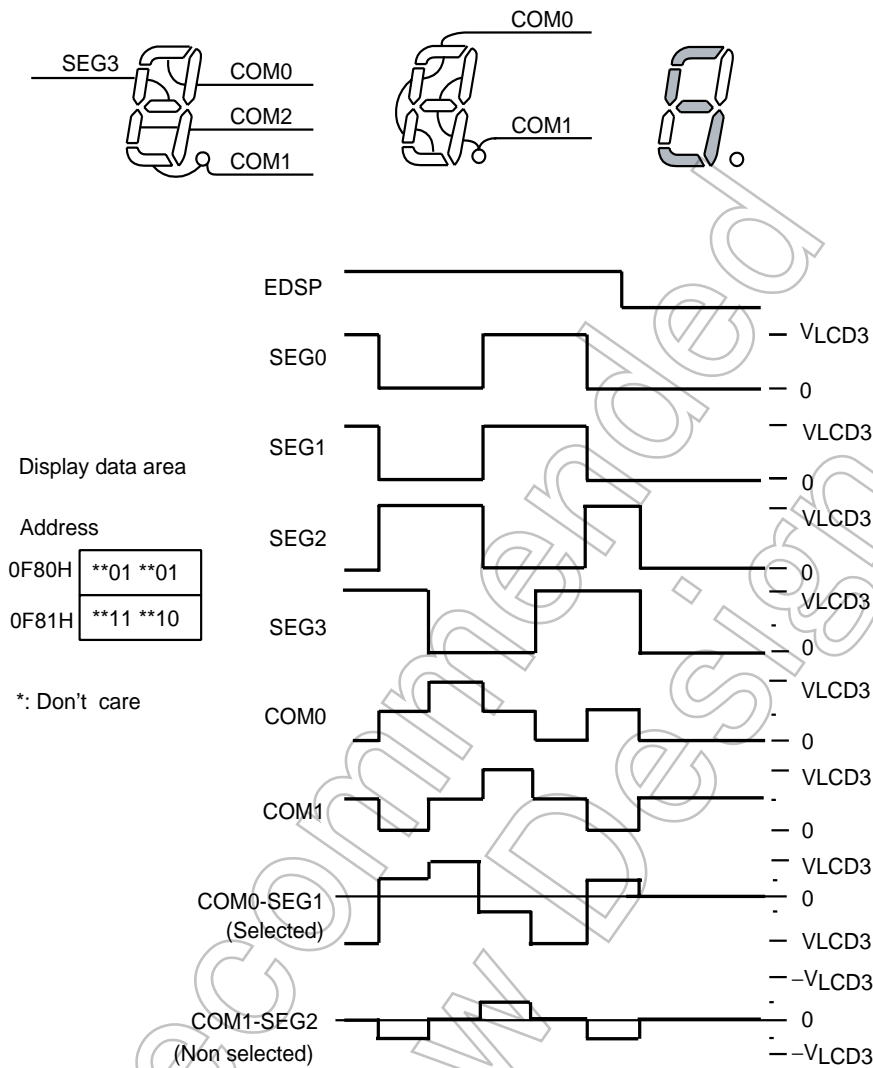


Figure 13-10 1/2 Duty (1/2 bias) Drive

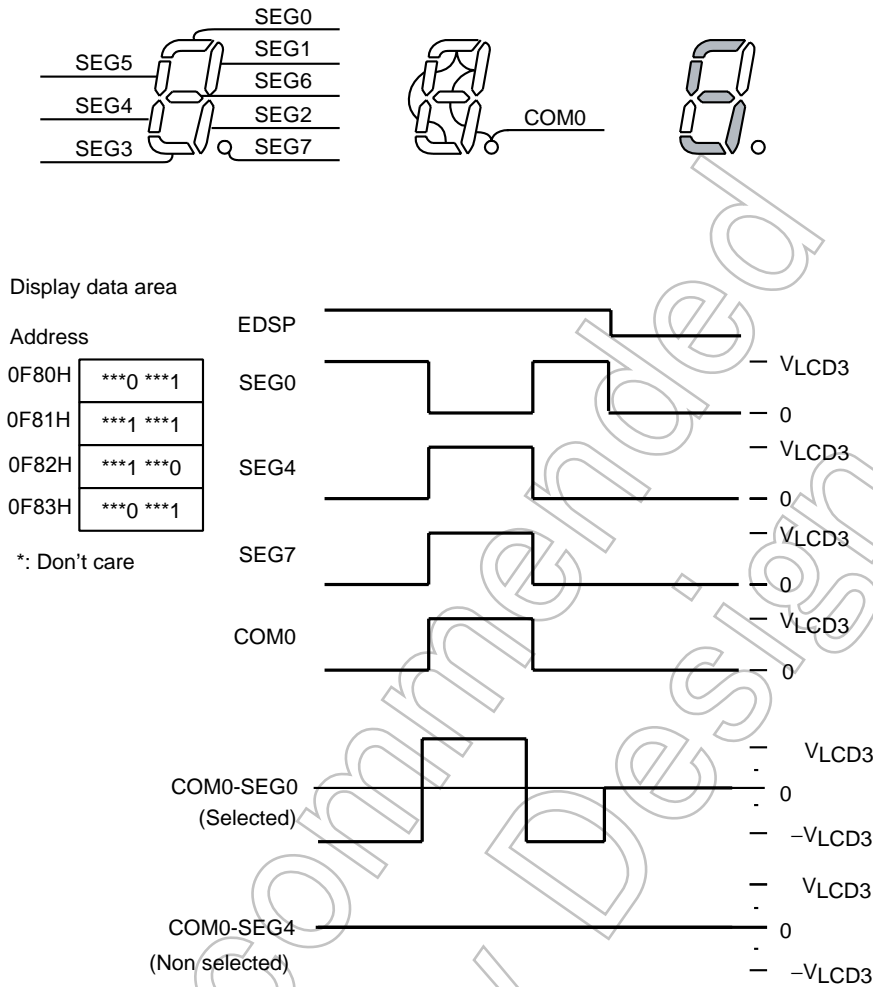


Figure 13-11 Static Drive

Not Recommended
for New Design

14. OTP operation

This section describes the function and basic operational blocks of TMP86P820FG. The TMP86P820FG has PROM in place of the mask ROM which is included in the TMP86C820/420. The configuration and function are the same as the TMP86C820/420. In addition, TMP86P820FG operates as the single clock mode when releasing reset. When using the dual clock mode, oscillate a low-frequency clock by [SET. (SYSCR2). XTEN] command at the beginning of program.

14.1 Operating mode

The TMP86P820FG has MCU mode and PROM mode.

14.1.1 MCU mode

The MCU mode is set by fixing the TEST/VPP pin to the low level. (TEST/VPP pin cannot be used open because it has no built-in pull-down resistor).

14.1.1.1 Program Memory

The TMP86P820FG has 8K bytes built-in one-time-PROM (addresses E000 to FFFFH in the MCU mode, addresses 0000 to 1FFFFH in the PROM mode).

When using TMP86P820FG for evaluation of mask ROM products, the program is written in the program storing area shown in Figure 14-1.

Since the TMP86P820FG supports several mask ROM sizes, check the difference in memory size and program storing area between the one-time PROM and the mask ROM to be used.

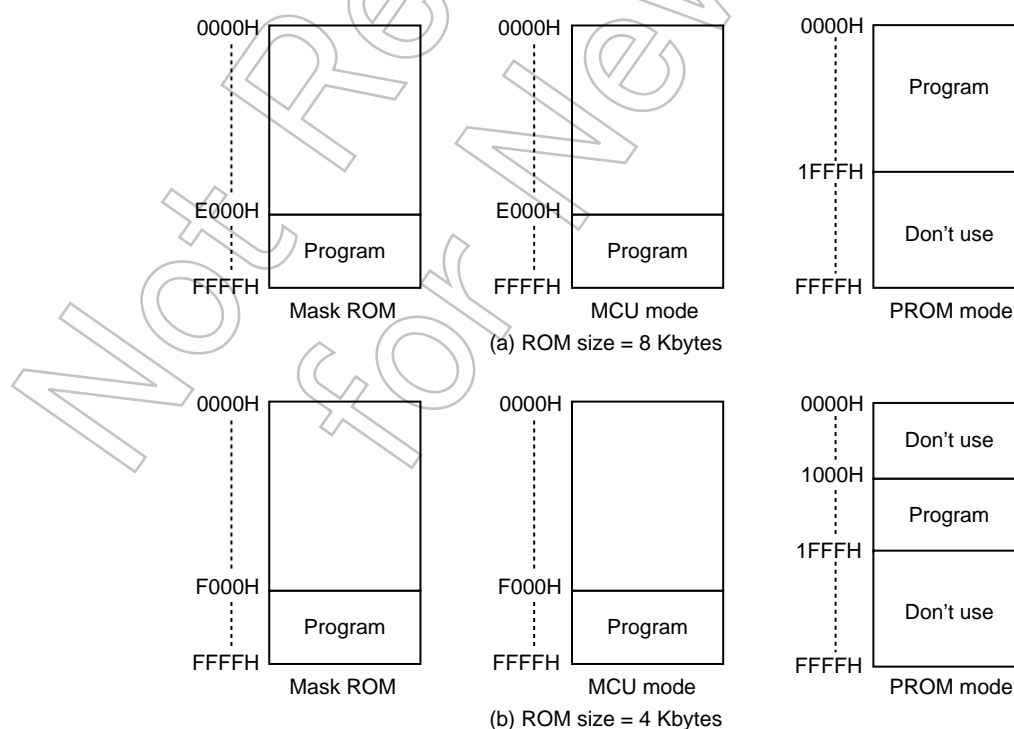


Figure 14-1 Program Memory Area

Note: The area that is not in use should be set data to FFH, or a general-purpose PROM programmer should be set only in the program memory area to access.

14.1.1.2 Data Memory

TMP86P820FG has a built-in 256 bytes Data memory (static RAM).

14.1.1.3 Input/Output Circuitry

1. Control pins

The control pins of the TMP86P820FG are the same as those of the TMP86C820/420 except that the TEST pin does not have a built-in pull-down resistor.

2. I/O ports

The I/O circuitries of the TMP86P820FG I/O ports are the same as those of the TMP86C820/420.

14.1.2 PROM mode

The PROM mode is set by setting the $\overline{\text{RESET}}$ pin, TEST pin and other pins as shown in Table 14-1 and Figure 14-2. The programming and verification for the internal PROM is achieved by using a general-purpose PROM programmer with the adaptor socket.

Table 14-1 Pin name in PROM mode

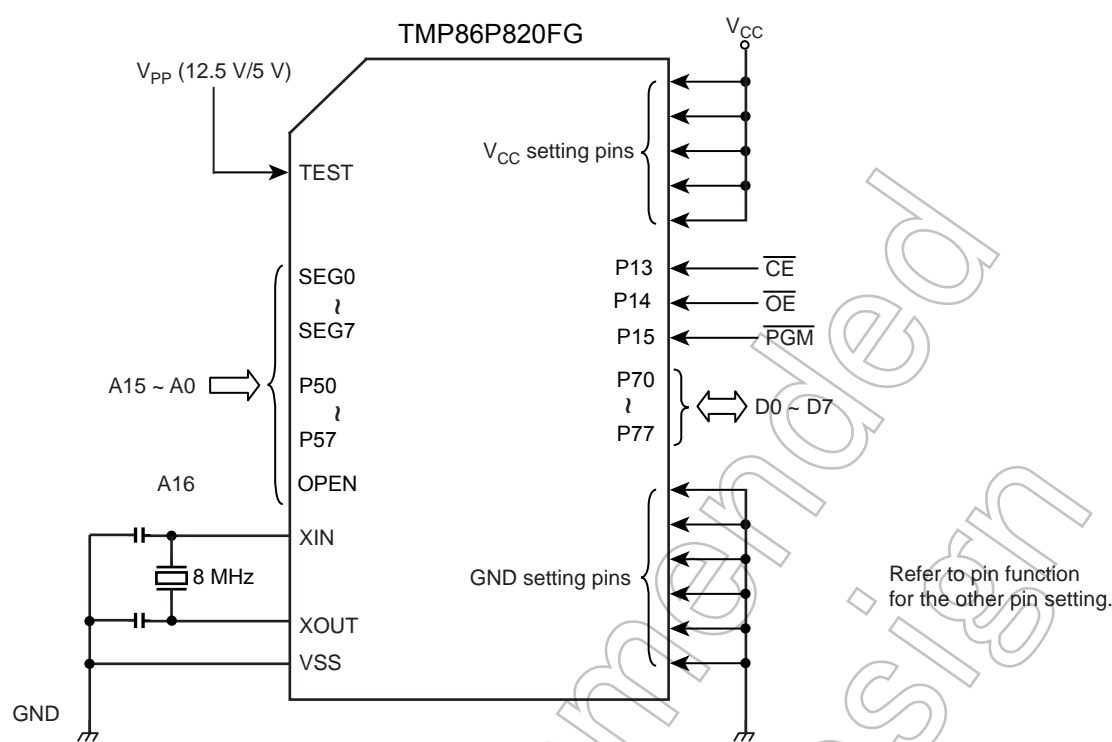
Pin name (PROM mode)	I/O	Function	Pin name (MCU mode)
A15 to A8	Input	Program memory address input	P57 to P50
A7 to A0	Input	Program memory address input	SEG7 to SEG0
D7 to D0	Input/Output	Program memory data input/output	P77 to P70
$\overline{\text{CE}}$	Input	Chip enable signal input	P13
$\overline{\text{OE}}$	Input	Output enable signal input	P14
$\overline{\text{PGM}}$	Input	Program mode signal input	P15
VPP	Power supply	+12.75V/5V (Power supply of program)	TEST
VCC	Power supply	+6.25V/5V	VDD
GND	Power supply	0V	VSS
VCC	Setting pin	Fix to "H" level in PROM mode	AVDD,P11,P21
GND	Setting pin	Fix to "L" level in PROM mode	VAREF,P10,P20,P22,P61
$\overline{\text{RESET}}$	Setting pin	Fix to "L" level in PROM mode	$\overline{\text{RESET}}$
XIN (CLK)	Input	Set oscillation with resonator In case of external CLK input, set CLK to XIN	XIN
XOUT	Output	and set XOUT to open.	XOUT

Note 1: The high-speed program mode can be used. The setting is different according to the type of PROM programmer to use, refer to each description of PROM programmer.

TMP86P820FG does not support the electric signature mode, apply the ROM type of PROM programmer to TC571000D/AD.

Note 2: No pin is applied to A16 pin of TC571000D/AD(Open) in PROM mode.

Always set the adapter socket switch to the "N" side when using TOSHIBA's adaptor socket.



Note 1: EPROM adaptor socket (TC571000 • 1M bit EPROM)

Note 2: PROM programmer connection adaptor sockets
BM11663 for TMP86P820FG

Note 3: Inside pin name for TMP86P820FG
Outside pin name for EPROM

Figure 14-2 PROM mode setting

14.1.2.1 Programming Flowchart (High-speed program writing)

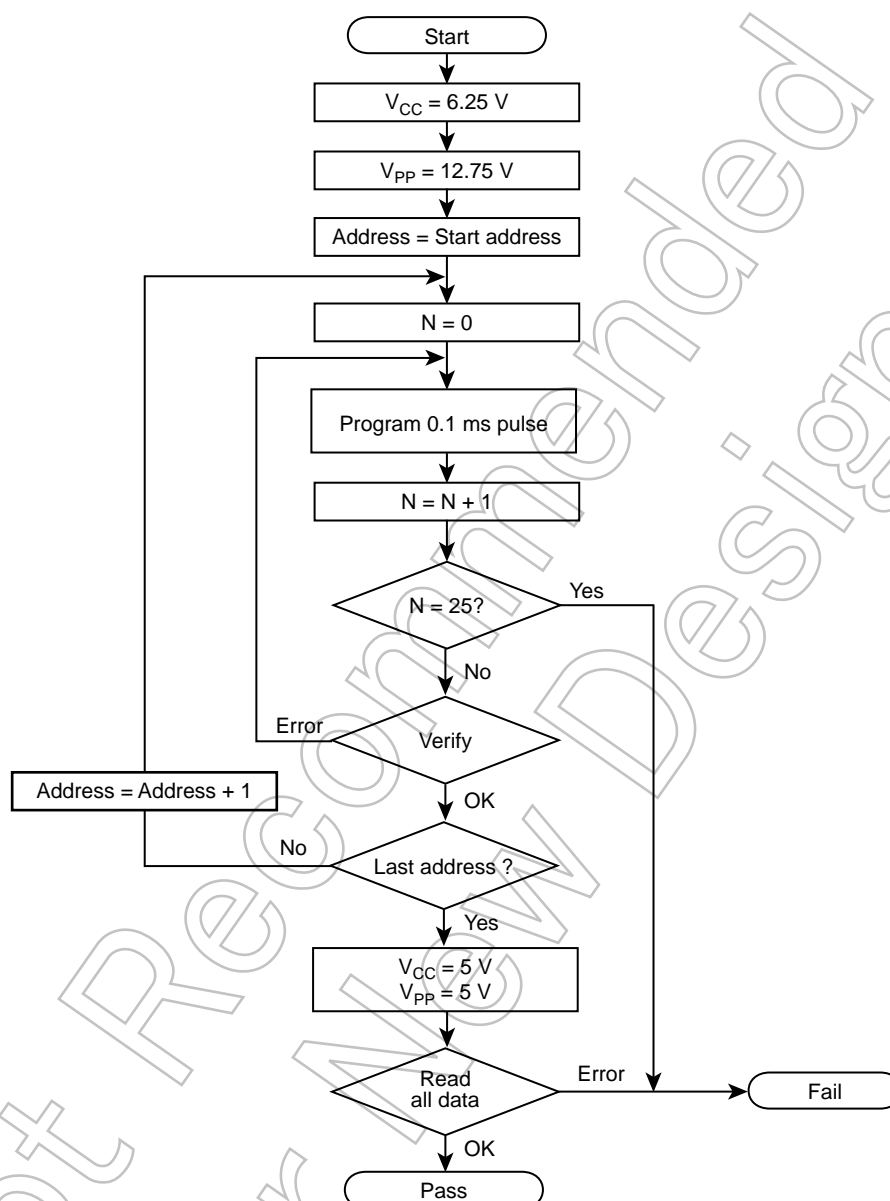


Figure 14-3 Programming Flowchart

The high-speed programming mode is set by applying $V_{pp}=12.75V$ (programming voltage) to the V_{pp} pin when the $V_{cc} = 6.25 V$. After the address and data are fixed, the data in the address is written by applying $0.1[msec]$ of low level program pulse to \overline{PGM} pin. Then verify if the data is written. If the programmed data is incorrect, another $0.1[msec]$ pulse is applied to \overline{PGM} pin. This programming procedure is repeated until correct data is read from the address (maximum of 25 times). Subsequently, all data are programmed in all address. When all data were written, verify all address under the condition $V_{cc}=V_{pp}=5V$.

14.1.2.2 Program Writing using a General-purpose PROM Programmer

1. Recommended OTP adaptor

BM11663 for TMP86P820FG

2. Setting of OTP adaptor

Set the switch (SW1) to "N" side.

3. Setting of PROM programmer

a. Set PROM type to TC571000D/AD.

Vpp: 12.75 V (high-speed program writing mode)

b. Data transmission (or Copy) (Note 1)

The PROM of TMP86P820FG is located on different address; it depends on operating mode: MCU mode and PROM mode. When you write the data of ROM for mask ROM products, the data should be transferred (or copied) from the address for MCU mode to that for PROM mode before writing operation is executed. For the applicable program areas of MCU mode and PROM mode are different, refer to TMP86P820FG" Figure 14-1 Program Memory Area".

Example: In the block transfer (copy) mode, executed as below.

8KB ROM capacity : 0E000~0FFFFH → 00000~01FFFH

4KB ROM capacity : 0F000~0FFFFH → 01000~01FFFH

c. Setting of the program address (Note 1)

Start address: 0000H (When 4KB ROM capacity, start address is 1000H)

End address: 1FFFH

4. Writing

Write and verify according to the above procedure "Setting of PROM programmer".

Note 1: For the setting method, refer to each description of PROM programmer.

Make sure to set the data of address area that is not in use to FFH.

Note 2: When setting MCU to the adaptor or when setting the adaptor to the PROM programmer, set the first pin of the adaptor and that of PROM programmer socket matched. If the first pin is conversely set, MCU or adaptor or programmer would be damaged.

Note 3: The TMP86P820FG does not support the electric signature mode.

If PROM programmer uses the signature, the device would be damaged because of applying voltage of 12±0.5V to pin 9(A9) of the address. Don't use the signature.

Not Recommended
for New Design

15. Input/Output Circuitry

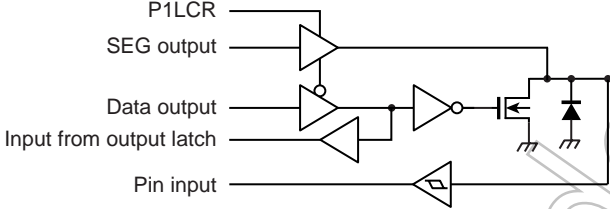
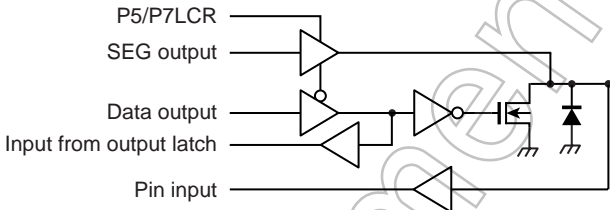
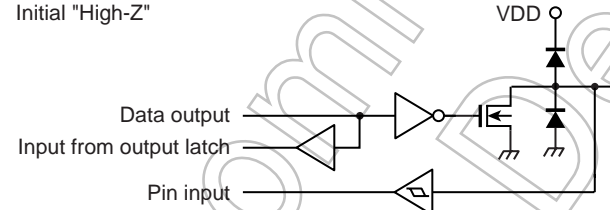
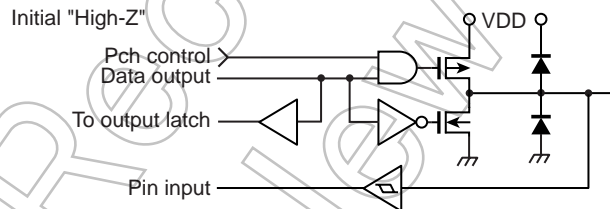
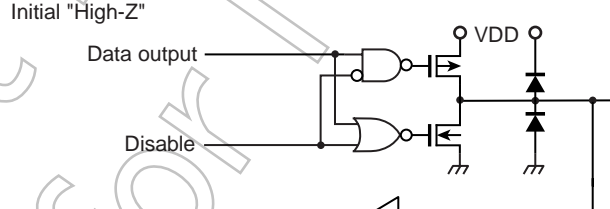
15.1 Control Pins

The input/output circuitries of the TMP86P820FG control pins are shown below.

Control Pin	I/O	Input/Output Circuitry	Remarks
XIN XOUT	Input Output		Resonator connecting pins (high-frequency) $R_f = 1.2\text{ M}\Omega$ (typ.) $R_O = 1.0\text{ k}\Omega$ (typ.)
XTIN XTOUT	Input Output		Resonator connecting pins (Low-frequency) $R_f = 6\text{ M}\Omega$ (typ.) $R_O = 220\text{ k}\Omega$ (typ.)
$\overline{\text{RESET}}$	Input Output		Sink open drain output Hysteresis input Pull-up resistor $R_{IN} = 220\text{ k}\Omega$ (typ.) $R = 1\text{ k}\Omega$ (typ.)
TEST	Input		Without pull-down resistor $R = 1\text{ k}\Omega$ (typ.) Fix the TEST pin at low-level in MCU mode.

Note: The TEST pin of the TMP86P820 does not have a pull-down resistor and protect diode (D_1). Fix the TEST pin at low-level in MCU mode.

15.2 Input/Output Ports

Port	I/O	Input/Output Circuitry	Remarks
P1	I/O	<p>Initial "High-Z"</p> 	Sink open drain output Hysteresis input
P5 P7	I/O	<p>Initial "High-Z"</p> 	Sink open drain output
P2	I/O	<p>Initial "High-Z"</p> 	Sink open drain output Hysteresis input
P3	I/O	<p>Initial "High-Z"</p> 	Sink open drain or C-MOS output Hysteresis input High current output (Nch) (Programmable port option)
P6	I/O	<p>Initial "High-Z"</p> 	Tri-state I/O Hysteresis input

Note: Port P1, P5 and P7 are sink open drain output. But they are also used as a segment output of LCD. Therefore, absolute maximum ratings of port input voltage should be used in -0.3 to $V_{DD} + 0.3$ volts.

16. Electrical Characteristics

16.1 Absolute Maximum Ratings

The absolute maximum ratings are rated values which must not be exceeded during operation, even for an instant. Any one of the ratings must not be exceeded. If any absolute maximum rating is exceeded, a device may break down or its performance may be degraded, causing it to catch fire or explode resulting in injury to the user. Thus, when designing products which include this device, ensure that no absolute maximum rating value will ever be exceeded.

($V_{SS} = 0\text{ V}$)

Parameter	Symbol	Pins	Rating	Unit
Supply Voltage	V_{DD}		-0.3 to 6.5	V
Program voltage	V_{PP}	TEST/ V_{PP}	-0.3 to 13.0	V
Input Voltage	V_{IN}		-0.3 to $V_{DD} + 0.3$	V
Output Voltage	V_{OUT1}		-0.3 to $V_{DD} + 0.3$	V
Output Current (Per 1 pin)	I_{OUT1}	P3, P6 Port	-1.8	mA
	I_{OUT2}	P1, P2, P5, P6, P7 Port	3.2	
	I_{OUT3}	P3 Port	30	
Output Current (Total)	ΣI_{OUT2}	P1, P2, P5, P6, P7 Port	60	
	ΣI_{OUT3}	P3 Port	80	
Power Dissipation [$T_{opr} = 85^{\circ}\text{C}$]	PD		350	mW
Soldering Temperature (Time)	T_{sld}		260 (10 s)	$^{\circ}\text{C}$
Storage Temperature	T_{stg}		-55 to 125	
Operating Temperature	T_{opr}		-40 to 85	

16.2 Recommended Operating Condition

The recommended operating conditions for a device are operating conditions under which it can be guaranteed that the device will operate as specified. If the device is used under operating conditions other than the recommended operating conditions (supply voltage, operating temperature range, specified AC/DC values etc.), malfunction may occur. Thus, when designing products which include this device, ensure that the recommended operating conditions for the device are always adhered to.

($V_{SS} = 0$ V, $T_{opr} = -40$ to 85°C)

Parameter	Symbol	Pins	Condition		Min	Max	Unit
Supply Voltage	V _{DD}		fc = 16 MHz	NORMAL1, 2 mode	4.5	5.5	V
				IDLE0, 1, 2 mode			
			fc = 8 MHz	NORMAL1, 2 mode	2.7		
				IDLE0, 1, 2 mode			
			fc = 4.2 MHz	NORMAL1, 2 mode	1.8		
				IDLE0, 1, 2 mode			
			fs = 32.768 kHz	SLOW1, 2 mode			
				SLEEP0, 1, 2 mode			
	STOP mode						
Input High Level	V _{IH1}	Except Hysteresis input	V _{DD} ≥ 4.5 V	V _{DD} × 0.70	V _{DD}		
	V _{IH2}	Hysteresis input		V _{DD} × 0.75			
	V _{IH3}		V _{DD} < 4.5 V	V _{DD} × 0.90			
Input Low Level	V _{IL1}	Except Hysteresis input	V _{DD} ≥ 4.5 V	0	V _{DD} × 0.30		
	V _{IL2}	Hysteresis input			V _{DD} × 0.25		
	V _{IL3}		V _{DD} < 4.5 V		V _{DD} × 0.10		
Clock Frequency	fc	XIN, XOUT	V _{DD} = 1.8 V to 5.5 V	1.0	4.2	MHz	
			V _{DD} = 2.7 V to 5.5 V		8.0		
			V _{DD} = 4.5 V to 5.5 V		16.0		
	fs	XTIN, XTOUT		30.0	34.0	kHz	

16.3 DC Characteristics

 $(V_{SS} = 0\text{ V}, T_{opr} = -40\text{ to }85^{\circ}\text{C})$

Parameter	Symbol	Pins	Condition	Min	Typ.	Max	Unit
Hysteresis Voltage	V_{HS}	Hysteresis input		–	0.9	–	V
Input Current	I_{IN1}	TEST	$V_{DD} = 5.5\text{ V}, V_{IN} = 5.5\text{ V}/0\text{ V}$	–	–	± 2	μA
	I_{IN2}	Sink Open Drain, Tri-state Port					
	I_{IN3}	$\overline{\text{RESET}}, \overline{\text{STOP}}$					
Input Resistance	R_{IN2}	$\overline{\text{RESET}}$ Pull-Up	$V_{DD} = 5.5\text{ V}, V_{IN} = 0\text{ V}$	100	220	450	$\text{k}\Omega$
Output Leakage Current	I_{LO}	Sink Open Drain, Tri-state Port	$V_{DD} = 5.5\text{ V}, V_{OUT} = 5.5\text{ V}/0\text{ V}$	–	–	± 2	μA
Output High Voltage	V_{OH2}	C-MOS, Tri-state Port	$V_{DD} = 4.5\text{ V}, I_{OH} = -0.7\text{ mA}$	4.1	–	–	V
Output Low Voltage	V_{OL}	Except XOUT and P3 Port	$V_{DD} = 4.5\text{ V}, I_{OL} = 1.6\text{ mA}$	–	–	0.4	
Output Low Current	I_{OL}	High Current Port (P3 Port)	$V_{DD} = 4.5\text{ V}, V_{OL} = 1.0\text{ V}$	–	20	–	mA
Supply Current in NORMAL1, 2 mode	I_{DD}		$V_{DD} = 5.5\text{ V}$ $V_{IN} = 5.3\text{ V}/0.2\text{ V}$ $f_c = 16\text{ MHz}$ $f_s = 32.768\text{ kHz}$	–	7.5	9	mA
Supply Current in IDLE0, 1, 2 mode				–	5.5	6.5	
Supply Current in SLOW1 mode			$V_{DD} = 3.0\text{ V}$ $V_{IN} = 2.8\text{ V}/0.2\text{ V}$ $f_s = 32.768\text{ kHz}$	–	18	42	μA
Supply Current in SLEEP1 mode				–	16	25	
Supply Current in SLEEP0 mode				–	12	20	
Supply Current in STOP mode			$V_{DD} = 5.5\text{ V}$ $V_{IN} = 5.3\text{ V}/0.2\text{ V}$	–	0.5	10	

Note 1: Typical values show those at $T_{opr} = 25^{\circ}\text{C}$, $V_{DD} = 5\text{ V}$

Note 2: Input current (I_{IN3}): The current through pull-up resistor is not included.

Note 3: I_{DD} does not include I_{REF} current.

Note 4: The supply currents of SLOW2 and SLEEP2 modes are equivalent to IDLE0, 1, 2.

16.4 AD Conversion Characteristics

($V_{SS} = 0.0\text{ V}$, $4.5\text{ V} \leq V_{DD} \leq 5.5\text{ V}$, $T_{opr} = -40\text{ to }85^{\circ}\text{C}$)

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Analog Reference Voltage	V _{AREF}		A _{VDD} – 1.5	–	A _{VDD}	V
Power Supply Voltage of Analog Control Circuit	A _{VDD}		V _{DD}			
Analog Reference Voltage Range (Note 4)	ΔV _{AREF}		3.0	–	–	
Analog Input Voltage	V _{AIN}		V _{SS}	–	V _{AREF}	
Power Supply Current of Analog Reference Voltage	I _{REF}	V _{DD} = A _{VDD} = V _{AREF} = 5.5 V V _{SS} = 0.0 V	–	0.6	1.0	mA
Non linearity Error		V _{DD} = A _{VDD} = 5.0 V V _{SS} = 0.0 V V _{AREF} = 5.0 V	–	–	±1	LSB
Zero Point Error			–	–	±1	
Full Scale Error			–	–	±1	
Total Error			–	–	±2	

($V_{SS} = 0.0\text{ V}$, $2.7\text{ V} \leq V_{DD} < 4.5\text{ V}$, $T_{opr} = -40\text{ to }85^{\circ}\text{C}$)

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Analog Reference Voltage	V _{AREF}		A _{VDD} – 1.5	–	A _{VDD}	V
Power Supply Voltage of Analog Control Circuit	A _{VDD}		V _{DD}			
Analog Reference Voltage Range (Note 4)	ΔV _{AREF}		2.5	–	–	
Analog Input Voltage	V _{AIN}		V _{SS}	–	V _{AREF}	
Power Supply Current of Analog Reference Voltage	I _{REF}	V _{DD} = A _{VDD} = V _{AREF} = 4.5 V V _{SS} = 0.0 V	–	0.5	0.8	mA
Non linearity Error		V _{DD} = A _{VDD} = 2.7 V V _{SS} = 0.0 V V _{AREF} = 2.7 V	–	–	±1	LSB
Zero Point Error			–	–	±1	
Full Scale Error			–	–	±1	
Total Error			–	–	±2	

($V_{SS} = 0.0\text{ V}$, $2.0\text{ V} \leq V_{DD} < 2.7\text{ V}$, $T_{opr} = -40\text{ to }85^{\circ}\text{C}$) (Note 5)

($V_{SS} = 0.0\text{ V}$, $1.8\text{ V} \leq V_{DD} < 2.0\text{ V}$, $T_{opr} = -10\text{ to }85^{\circ}\text{C}$) (Note 5)

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Analog Reference Voltage	V _{AREF}		A _{VDD} – 0.9	–	A _{VDD}	V
Power Supply Voltage of Analog Control Circuit	A _{VDD}		V _{DD}			
Analog Reference Voltage Range (Note 4)	ΔV _{AREF}	1.8 V ≤ V _{DD} < 2.0 V	1.8	–	–	
		2.0 V ≤ V _{DD} < 2.7 V	2.0	–	–	
Analog Input Voltage	V _{AIN}		V _{SS}	–	V _{AREF}	
Power Supply Current of Analog Reference Voltage	I _{REF}	V _{DD} = A _{VDD} = V _{AREF} = 2.7 V V _{SS} = 0.0 V	–	0.3	0.5	mA
Non linearity Error		V _{DD} = A _{VDD} = 1.8 V V _{SS} = 0.0 V V _{AREF} = 1.8 V	–	–	±2	LSB
Zero Point Error			–	–	±2	
Full Scale Error			–	–	±2	
Total Error			–	–	±4	

Note 1: The total error includes all errors except a quantization error, and is defined as maximum deviation from the ideal conversion line.

Note 2: Conversion time is different in recommended value by power supply voltage.
About conversion time, refer to "8-bit AD Converter".

Note 3: Please use input voltage to AIN input Pin in limit of $V_{AREF} - V_{SS}$.

When voltage of range outside is input, conversion value becomes unsettled and gives affect to other channel conversion value.

Note 4: Analog Reference Voltage Range: $\Delta V_{AREF} = V_{AREF} - V_{SS}$

Note 5: When AD is used with $V_{DD} < 2.7$ V, the guaranteed temperature range varies with the operating voltage.

Note 6: The A_{VDD} pin should be fixed on the V_{DD} level even though AD convertor is not used.

Not Recommended
for New Design

16.5 AC Characteristics

($V_{SS} = 0\text{ V}$, $V_{DD} = 4.5\text{ to }5.5\text{ V}$, $T_{opr} = -40\text{ to }85^{\circ}\text{C}$)

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Machine Cycle Time	t _{cy}	NORMAL1, 2 mode	0.25	—	4	μs
		IDLE1, 2 mode				
		SLOW1, 2 mode	117.6	—	133.3	
		SLEEP1, 2 mode				
High Level Clock Pulse Width	t _{WCH}	For external clock operation (XIN input) fc = 16 MHz	—	31.25	—	ns
Low Level Clock Pulse Width	t _{WCL}					
High Level Clock Pulse Width	t _{WCH}	For external clock operation (XTIN input) fs = 32.768 kHz	—	15.26	—	μs
Low Level Clock Pulse Width	t _{WCL}					

($V_{SS} = 0\text{ V}$, $V_{DD} = 2.7\text{ to }4.5\text{ V}$, $T_{opr} = -40\text{ to }85^{\circ}\text{C}$)

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Machine Cycle Time	t _{cy}	NORMAL1, 2 mode	0.5	—	4	μs
		IDLE1, 2 mode				
		SLOW1, 2 mode	117.6	—	133.3	
		SLEEP1, 2 mode				
High Level Clock Pulse Width	t _{WCH}	For external clock operation (XIN input)	—	62.5	—	ns
Low Level Clock Pulse Width	t _{WCL}	f _c = 8 MHz				
High Level Clock Pulse Width	t _{WCH}	For external clock operation (XTIN input)	—	15.26	—	μs
Low Level Clock Pulse Width	t _{WCL}	f _s = 32.768 kHz				

($V_{SS} = 0\text{ V}$, $V_{DD} = 1.8\text{ to }2.7\text{ V}$, $T_{opr} = -40\text{ to }85^{\circ}\text{C}$)

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Machine Cycle Time	t _{cy}	NORMAL1, 2 mode	0.95	—	4	μs
		IDLE1, 2 mode				
		SLOW1, 2 mode	117.6	—	133.3	
		SLEEP1, 2 mode				
High Level Clock Pulse Width	t _{WCH}	For external clock operation (XIN input) f _c = 4.2 MHz	—	119.05	—	ns
Low Level Clock Pulse Width	t _{WCL}					
High Level Clock Pulse Width	t _{WCH}	For external clock operation (XTIN input) f _s = 32.768 kHz	—	15.26	—	μs
Low Level Clock Pulse Width	t _{WCL}					

16.6 Timer Counter 1 input (ECIN) Characteristics

($V_{SS} = 0\text{ V}$, $T_{opr} = -40\text{ to }85^{\circ}\text{C}$)

Parameter	Symbol	Condition		Min	Typ.	Max	Unit
TC1 input (ECIN input)	t _{TC1}	Frequency measurement mode $V_{DD} = 4.5\text{ to }5.5\text{ V}$	Single edge count	—	—	16	MHz
		Frequency measurement mode $V_{DD} = 2.7\text{ to }4.5\text{ V}$	Single edge count	—	—	8	
		Frequency measurement mode $V_{DD} = 1.8\text{ to }2.7\text{ V}$	Single edge count	—	—	4.2	

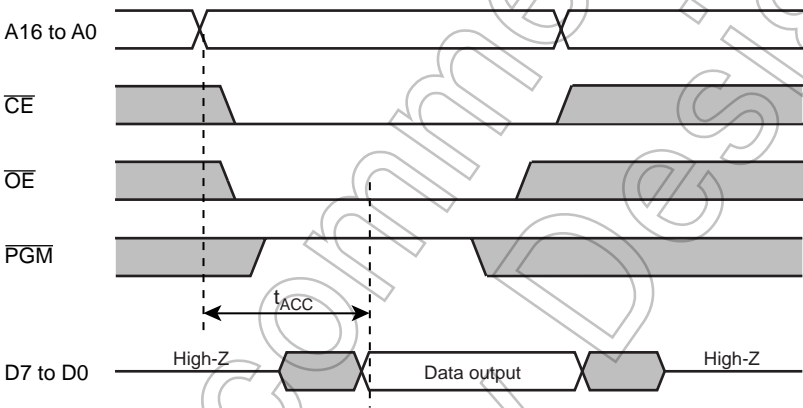
16.7 DC Characteristics, AC Characteristics (PROM mode)

16.7.1 Read operation in PROM mode

(V_{SS} = 0 V, T_{opr} = -40 to 85°C)

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
High level input voltage	V _{IH4}		2.2	—	V _{CC}	V
Low level input voltage	V _{IL4}		0	—	0.8	
Power supply	V _{CC}		4.75	5.0	5.25	
Program supply of program	V _{PP}					
Address access time	t _{ACC}	V _{CC} = 5.0 ± 0.25 V	—	1.5tcyc + 300	—	ns

Note: tcyc = 500 ns, f_{CLK} = 8 MHz

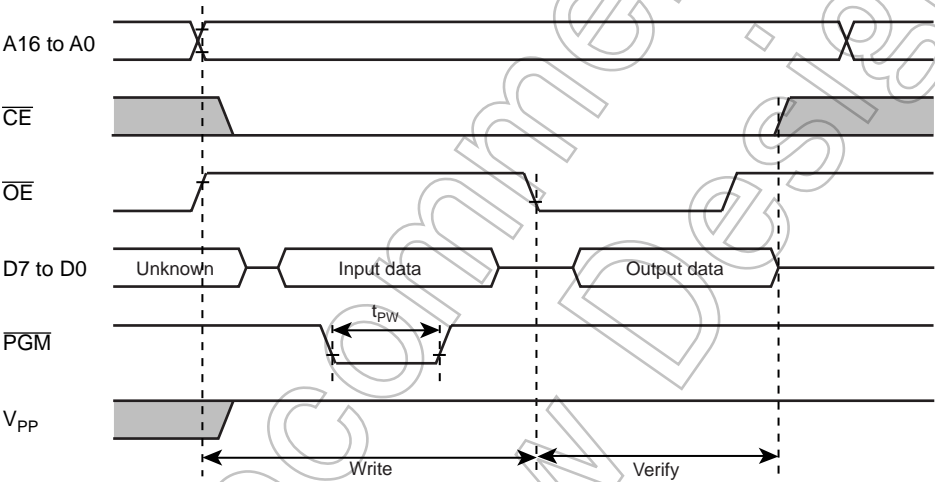


16.7.2 Program operation (High-speed) (Topr = 25 ± 5°C)

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
High level input voltage	V_{IH4}		2.2	–	V_{CC}	V
Low level input voltage	V_{IL4}		0	–	0.8	
Power supply	V_{CC}		6.0	6.25	6.5	
Program supply of program	V_{PP}		12.5	12.75	13.0	
Pulse width of initializing program	t_{PW}	$V_{CC} = 6.0\text{ V}$	0.095	0.1	0.105	ms

Note: $t_{cyc} = 500\text{ ns}$, $f_{CLK} = 8\text{ MHz}$

High-speed program writing

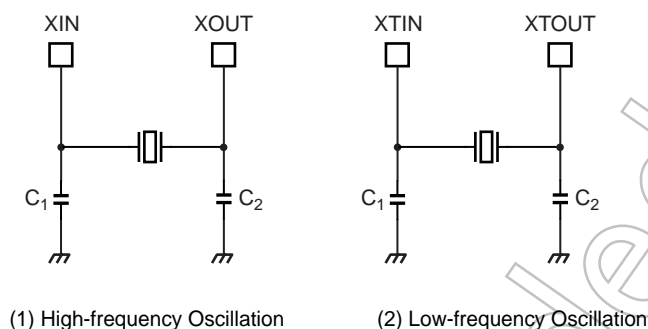


Note 1: The power supply of V_{PP} (12.75 V) must be set power-on at the same time or the later time for a power supply of V_{CC} and must be clear power-on at the same time or early time for a power supply of V_{CC} .

Note 2: The pulling up/down device causes a damage for the device while the voltage supplied to device (condition of especially $V_{PP} = 12.75\text{ V} \pm 0.25\text{ V}$). Do not pull up/down at programming.

Note 3: Use the recommended adapter and mode.
Using other than the above condition may cause the trouble of the writing.

16.8 Recommended Oscillating Conditions



Note 1: A quartz resonator can be used for high-frequency oscillation only when V_{DD} is 2.7 V or above. If V_{DD} is below 2.7 V, use a ceramic resonator.

Note 2: To ensure stable oscillation, the resonator position, load capacitance, etc. must be appropriate. Because these factors are greatly affected by board patterns, please be sure to evaluate operation on the board on which the device will actually be mounted.

Note 3: For the resonators to be used with Toshiba microcontrollers, we recommend ceramic resonators manufactured by Murata Manufacturing Co., Ltd.

For details, please visit the website of Murata at the following URL:
<http://www.murata.com>

16.9 Handling Precaution

- The solderability test conditions for lead-free products (indicated by the suffix G in product name) are shown below.

1. When using the Sn-37Pb solder bath
 - Solder bath temperature = 230 °C
 - Dipping time = 5 seconds
 - Number of times = once
 - R-type flux used
2. When using the Sn-3.0Ag-0.5Cu solder bath
 - Solder bath temperature = 245 °C
 - Dipping time = 5 seconds
 - Number of times = once
 - R-type flux used

Note: The pass criterion of the above test is as follows:

Solderability rate until forming $\geq 95\%$

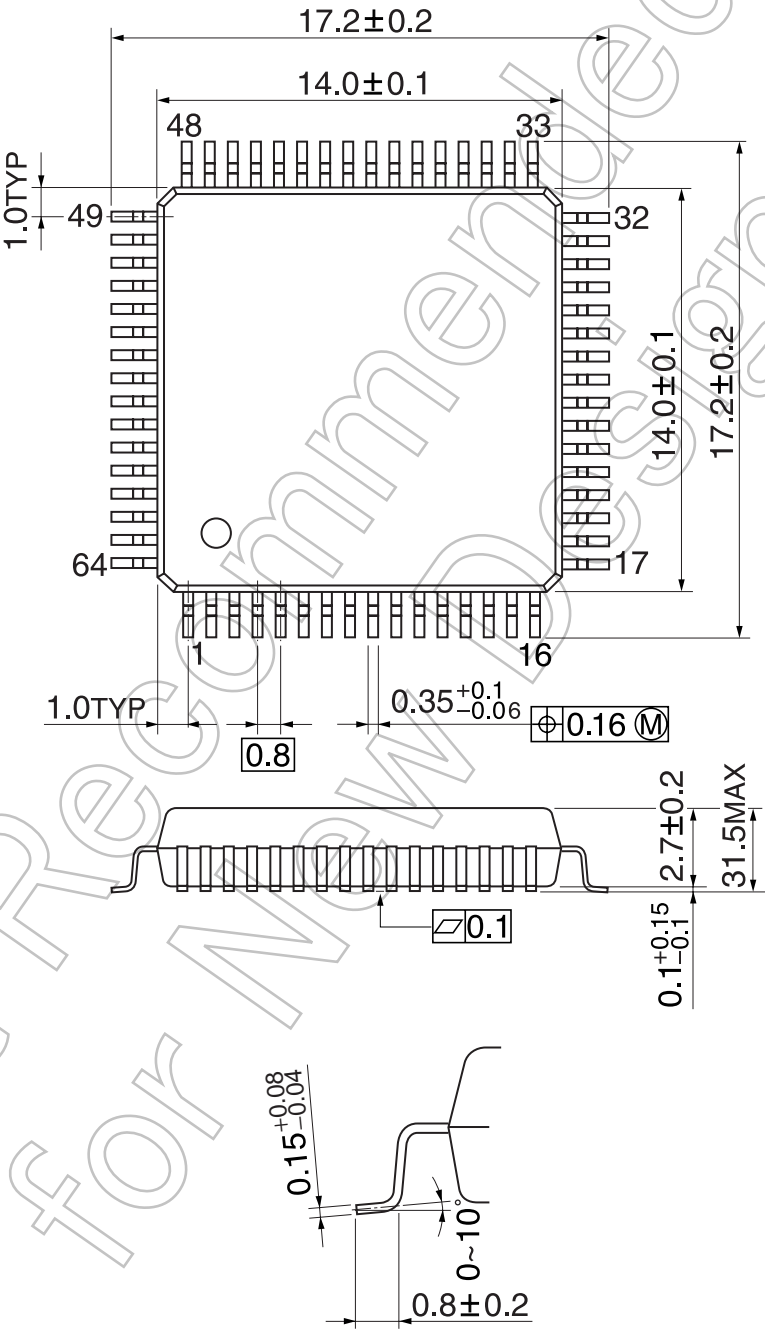
- When using the device (oscillator) in places exposed to high electric fields such as cathode-ray tubes, we recommend electrically shielding the package in order to maintain normal operating condition.

Not Recommended
for New Design

17. Package Dimensions

QFP64-P-1414-0.80C Rev 01

Unit: mm



Not Recommended
for New Design

This is a technical document that describes the operating functions and electrical specifications of the 8-bit microcontroller series TLCS-870/C (LSI).

Toshiba provides a variety of development tools and basic software to enable efficient software development.

These development tools have specifications that support advances in microcomputer hardware (LSI) and can be used extensively. Both the hardware and software are supported continuously with version updates.

The recent advances in CMOS LSI production technology have been phenomenal and microcomputer systems for LSI design are constantly being improved. The products described in this document may also be revised in the future. Be sure to check the latest specifications before using.

Toshiba is developing highly integrated, high-performance microcomputers using advanced MOS production technology and especially well proven CMOS technology.

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