

TB6569FG/FTG Usage considerations

Summary

The TB6569FG/FTG is a full-bridge driver IC for a DC motor. Output transistors adopt a MOS structure.

Efficient drive with lower heat generation is realized by adopting the low ON-resistance MOSFETs and a PWM control.

Furthermore, the TB6569FG/FTG has four selectable operations of forward (clockwise), reverse (counter-clockwise), short brake, and stop modes by setting two input signals of IN1 and IN2.

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1. Power Supply Voltage

(1) Operating Range of Power Supply Voltage

The absolute maximum voltage rating of the TB6569FG/FTG is 50 V. However, please use the IC within the range of 10 V to 45 V in actual usage.

(2) Power-ON/Power-OFF

Since the TB6569FG/FTG has a single power supply of VM and an undervoltage lockout circuit, it has no special procedures for turning on and off itself. However, unstable power supplies result in abnormal IC operations. Therefore, it is recommended to turn on the IC and then rotate the motor, after confirming that both IN1 and IN2 are in low states and that the power supply of VM is stable.

It is likewise recommended to turn off the TB6569FG/FTG after the motor movement is completely stopped.

2. Output Current

Note that the absolute maximum output current rating of the TB6569FG/FTG varies with the VM. OUT1 and OUT2 should be kept under 4.5 A when VM is 36 V or less. And they should be kept under 4.0 A when VM is more than 36 V.

Also, the usage conditions such as the ambient temperature, presence or absence of a heatsink, board layout and IC mount technique, have effect on increase and decrease of the available average output current. When T_j is less than 150°C, the TB6569FG/FTG should be used with the absolute maximum output current rating of 4.0 A, or with the average output current of 4.5 A or less.

3. Control Inputs

Even if there are pulse inputs to IN1, IN2, PWM, and VREF, they never seep into VM as long as when the VM power supply is turned off; therefore the TB6569FG/FTG will never be turned on.

To release the TSD and the ISD circuits, input low signal to IN1 and IN2 for 1 μ s or more.

4. PWM Frequency

The PWM input through the PWM pin controls the motor speed.
(The PWM control is also accomplished by the PWM input through the IN1 and IN2 pins instead of the PWM pin.)

During the PWM control, the normal mode and the short brake mode are repeated.
The TB6569FG/FTG internally generates the blank time so as to prevent the shoot-through current that occurs when the upper and lower power transistors in the output circuit turn on at the same time. This blank time is generated when the upper and lower power transistors switch.

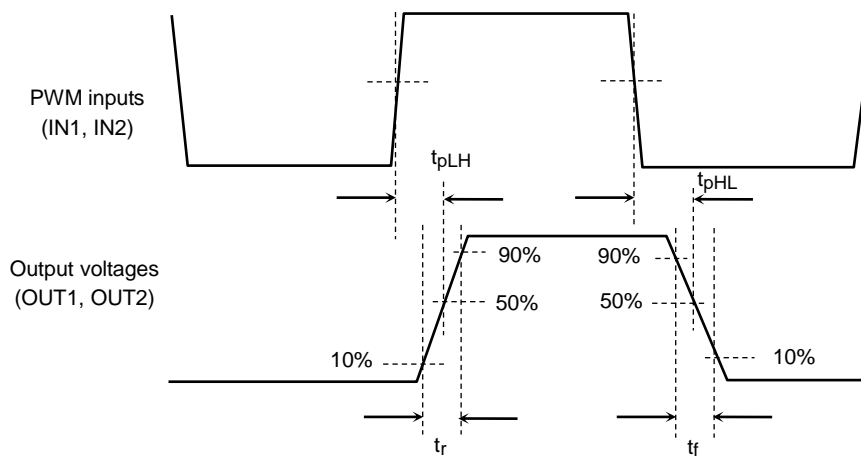
Therefore, the PWM control with the synchronous rectification is available without inputting the off time externally.

Though the reference PWM frequency in the operational range is stated as 100 kHz, in actual operations, the output voltage may be distorted to the input even when the IC operates within the operating range. (Refer to below figure of the switching characteristics)

The TB6569FG/FTG can support the frequency of over 100 kHz only as far as its output distortions to the inputs and the duty gaps are taken into account when it is used.

Note that the values of the following switching characteristics are given as typical values. The IC should be used with a sufficient safety margin because they vary with power supply voltages, temperatures, and IC variation.

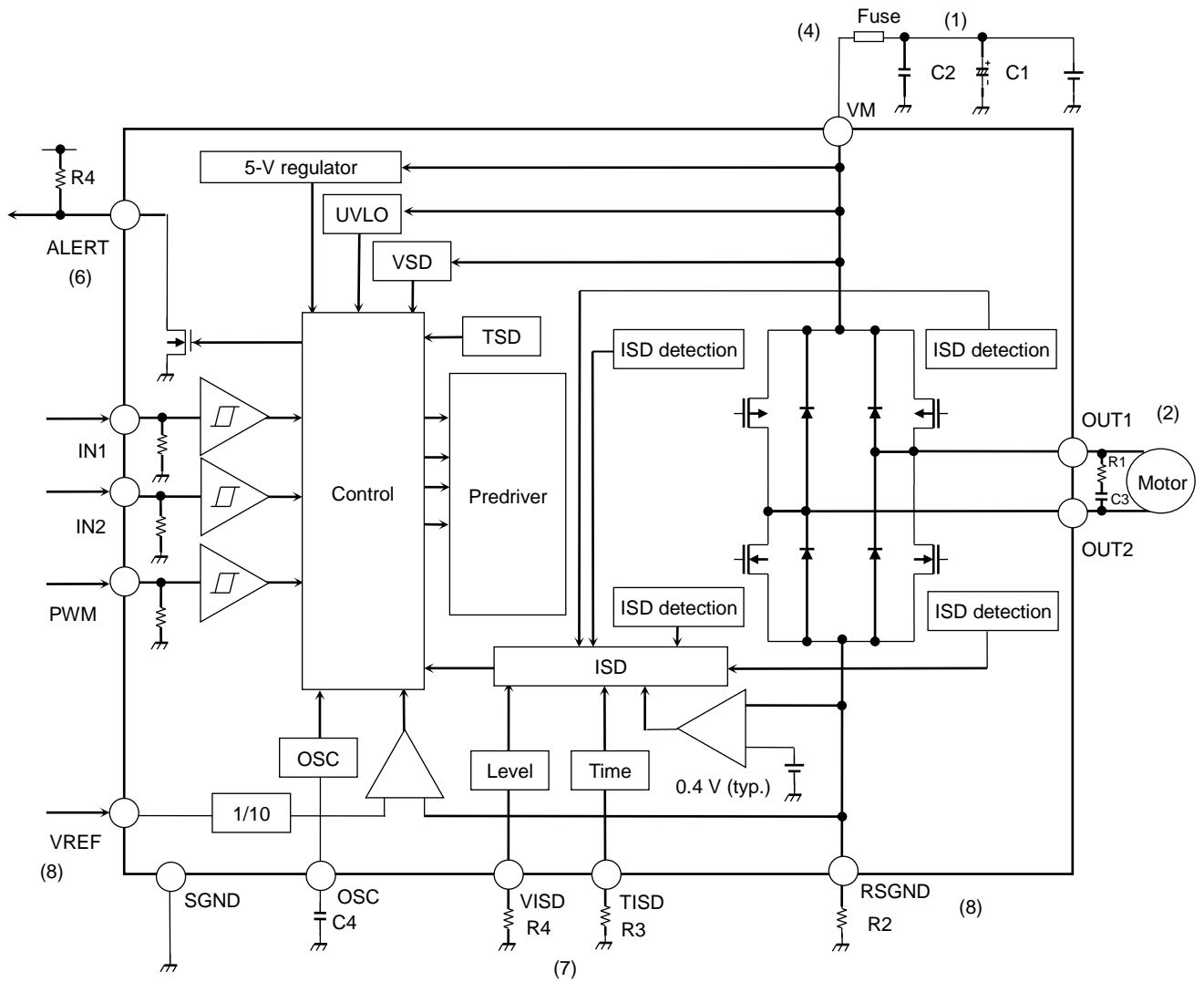
Switching Characteristics



VM = 24 V Ta = 25°C

Characteristics	Value	Unit
t_{pLH}	650 (typ.)	ns
t_{pHL}	450 (typ.)	
t_r	90 (typ.)	
t_f	130 (typ.)	

5. Application Circuit



(1) Capacitors Connected to the Power Supply Pin

Connect the capacitors between VM and GND as near the IC as possible.

Recommended Values

Characteristics	Symbol	Recommended Value	Remarks
VM – GND	C1	10 μ F to 100 μ F	Electrolytic capacitor
	C2	0.1 μ F to 1 μ F	Ceramic capacitor

(2) Capacitor and Resistor Between the Outputs

Connect the R1 resistor and the C3 capacitor only for removing the brush noise of the motor. If so, limit the current by using R1 because the outputs momentarily move to the short circuit mode in conduction if C3 is not charged.

(3) Routing of VM, OUT1, OUT2 and RSGND

The motor causes a large current flow through the TB6569FG/FTG. Therefore, sufficient space should be secured on designing the IC wiring pattern. Particularly for RSGND and SGND, an enough large space for their connections to GND should be secured so as not to be affected by wiring impedance.

(4) Fuse

For preventing a continuous flow of a large current due to overcurrent or IC damages, an appropriate fuse should be placed in the power supply of the TB6569FG/FTG.

The TB6569FG/FTG may fail because of illegal use such as exceeding the absolute maximum ratings, incorrect wirings and abnormal pulse noise induced by wirings and loads. As a result, a large current continuously flows into the TB6569FG/FTG leads to smoking and ignition. To make these negative impacts as small as possible, appropriate control of the capacitance and weld time of the fuse as well as positioning of the fuse in the circuit is required.

The TB6569FG/FTG incorporates an overcurrent detection circuit (ISD). However, it does not necessarily protect the TB6569FG/FTG in any case. On activation of the ISD circuit, overcurrent conditions should be removed immediately. Depending on the usage and the use environment of the TB6569FG/FTG, like using it with the absolute maximum ratings being exceeded, the ISD circuit may not operate correctly; or the TB6569FG/FTG may be broken before the ISD circuit is activated. Even after the activation of the ISD circuit, the TB6569FG/FTG may be destroyed due to the IC heating if overcurrent continues flowing too long.

There is a concern that a secondary destruction of the IC due to continuous overcurrent may occur. Another concern is that the ISD circuit may not run due to its blank time, interacting with the output load conditions. Toshiba, therefore, describes in the specification that the ISD circuit does not necessarily run in any case as one of the usage considerations.

For instance, if a current that neither reaches the absolute maximum output current rating nor infringes the lower limit of the operating voltage of the ISD circuit continues flowing, the DMOS transistors in the output stage will be degraded. On the other hand, if once a current exceeding the absolute maximum output current rating flows into the DMOS transistors in the output stage, they are degraded as well. Therefore, even though the TB6569FG/FTG is not broken after single overcurrent detection, it may be broken after two or three times of overcurrent detection because repeated detections will deepen the DMOS degradation.

Toshiba recommends the use of a fuse in the power supply to cope with such a secondary destruction.

(5) FIN

FIN plays a role as a heatsink. The heat protection should be considered when designing the board layout.

(FIN should be insulated or connected to GND because it is connected to the back of the chip electrically.)

(6) ALERT Pin

The ALERT pin behaves as an open-drain output. It has a pull-up resistor in an external power source to send out the High state. When the ALERT pin outputs Low, the TB6569FG/FTG operates normally. The High state (High-impedance) on the ALERT pin indicates some failure of the TB6569FG/FTG operation (UVLO, TSD, VSD and/or ISD is running).

It is recommended to use a pull-up resistor of 10 kΩ to 100 kΩ.

(7) Resistor Controls of the VISD and TISD Pins

Each ISD function is provided for each of the four output power transistors.

The resistor control of the VISD pin enables the current threshold tuning. If any one of the four ISD detection runs over the ISD detection time (mask time), it turns off all output power transistors (High-impedance).

The resistor control of the TISD pin enables the ISD detection time (mask time) tuning.

Configuring both the IN1 and IN2 pins low releases the ISD detection and returns the TB6569FG/FTG to the normal operation.

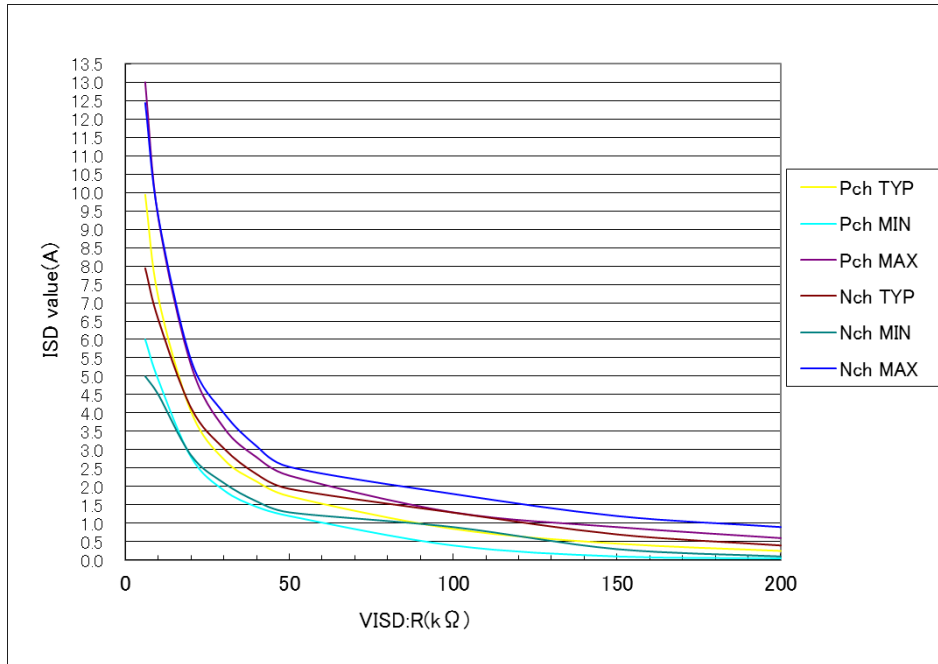
Relation of the ISD current thresholds and the external resistor values of the VISD pin, and that of the ISD mask times and the external resistor values of the TISD pin are shown in the following figures for your reference.

The mask time of the TISD pin should be configured to avoid malfunction by noise and IC destruction in overcurrent detection.

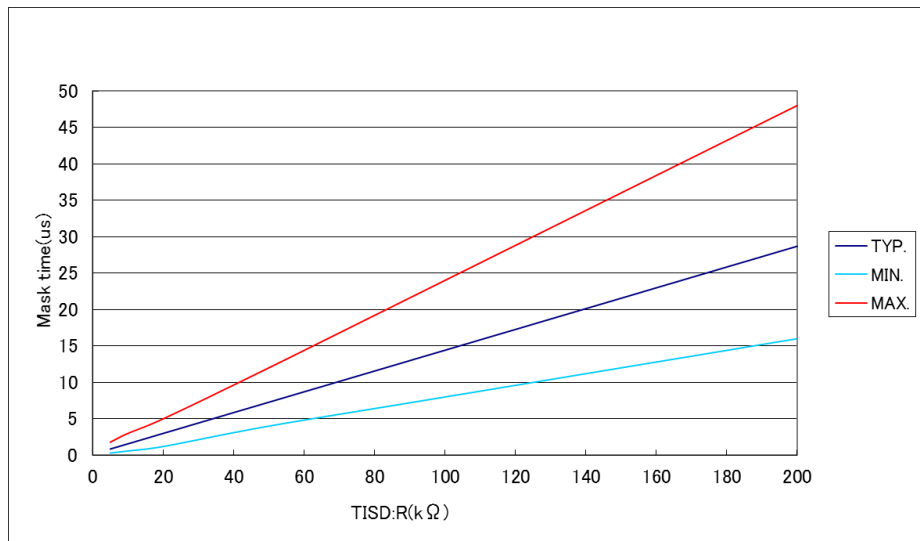
The resistors of less than 5 k Ω should not be connected to the VISD and TISD pins. However, when disabling the overcurrent detection, please connect the TISD pin to GND.

ISD behavior on load shorts

Irrespective of usage of the constant current PWM control (i.e. a detection resistor is connected between the RSGND and the SGND pins), pay attention that if wiring impedance exists between the RSGND and the SGND pins, the ISD function may not work even on a load short because the constant current PWM control operates before running of the ISD circuit.



External Resistor Value of the VISD Pin – ISD Current Thresholds Curves (for reference only)



External Resistor Value of the TISD Pin – ISD Mask Time Curves (for reference only)

Note: Range of external resistance: 5kΩ to 200kΩ

Note: The complementary output circuit consists of N-channel and P-channel DMOS transistors, thus the ISD current thresholds show slight differences.

(8) Calculations for Constant Current PWM Control of the RSGND and the VREF Pins

The frequency of the constant current PWM control is determined by the peak current in the constant current operations. This peak current is generated by the voltage across the VREF pin. The peak current values are calculated by the following equations:

$$I_O = V_{REF}/R_2 \times 1/10 \text{ [A]} \quad \text{For example, when } R_2 = 0.2 \Omega, \text{ and } V_{REF} = 2 \text{ V, then } I_O = 1 \text{ A}$$

The frequency of the constant current PWM is also configurable by controlling the capacitor of the OSC pin. The oscillation frequency is approximated by the following equation:

$$f_{osc} \text{ [Hz]} (\text{typ.}) = 0.42 / (C_{osc} \text{ [F]} \times 10^3) \quad \text{For example, when } C_4; C_{osc} = 1800 \text{ pF,}$$

then $f_{osc} = 233 \text{ kHz}$

The OSC frequency should be configured to 500 kHz or less. If it is configured higher than 500 kHz, the switching loss of the output stages controlled by the PWM frequency becomes larger. Note, however, that if the OSC frequency is too low, the PWM frequency may fall within the audible range.

Sufficient safety margin should be secured because the PWM frequency varies depending on the power supply voltage, temperature and IC variation.

The RSGND pin turns off all output transistors (High-impedance: Hi-Z) when the voltage exceeds 0.4 V (typ.) in order to avoid over voltage when the resistor for overvoltage detection is connected. This function is the same as that of the ISD circuit. At this time, also the ALERT pin outputs high. To return the TB6569FG/FTG to the normal operation by releasing the overvoltage detection, please configure both the IN1 and IN2 pins low.

For the RSGND pin, it is recommended to use an overvoltage detection resistor of 0.1 Ω or more.

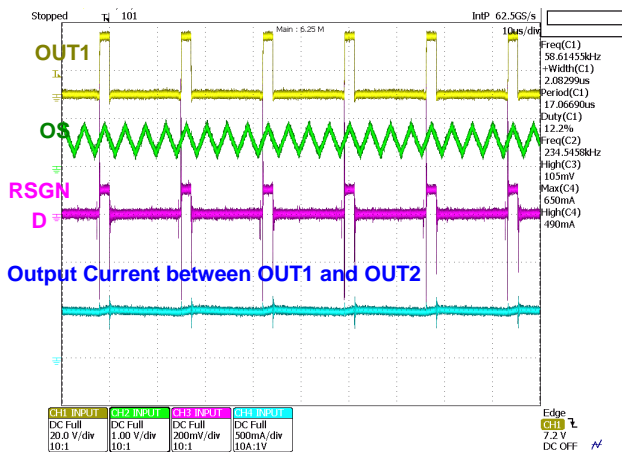
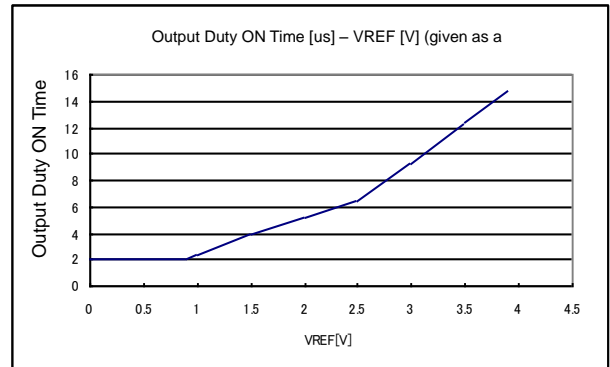
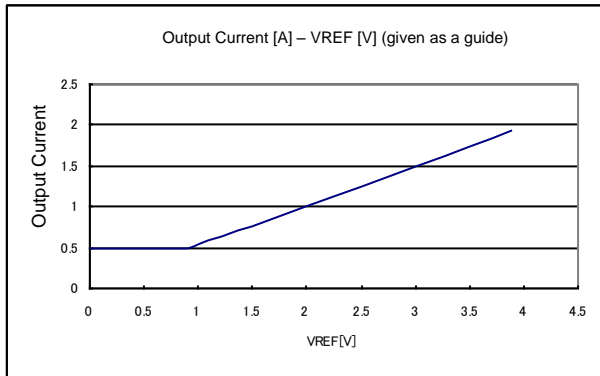
The voltage across the detection resistor is compared to the reference voltage across the SGND pin. Therefore, the overvoltage detection resistor of the RSGND should be placed near the RSGND pin and the SGND pin not to be affected by wiring impedance.

When the constant current PWM control is not used, the RSGND pin should be shorted to the SGND pin instead of connecting to the R2 resistor.

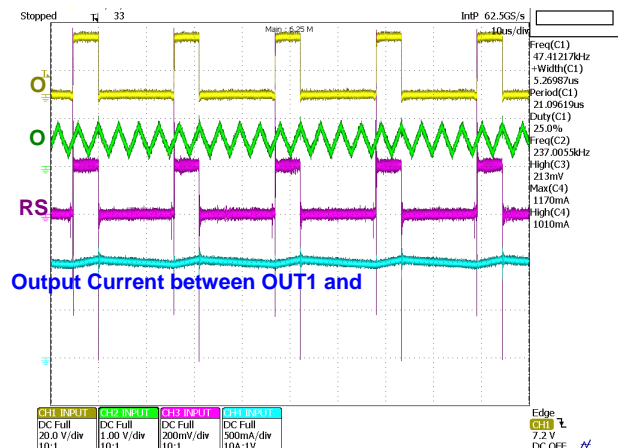
In this case, sufficient space should be secured between the RSGND pin and the SGND pin; otherwise wiring impedance generates between these pins, and it acts like a detection resistor. As a result, the motor moves as if its speed is controlled by a constant current PWM frequency.

As for the OSC pin, please connect the capacitor or connect GND directly.

The characteristics curves of VREF vs. output current and VREF vs. output duty ON time are shown below. And their active waveforms are also shown in below figures.
 For prevention of abnormal operations due to noise, the duty ON time of about 2 μ s is provided by default. Note that the offset current is provided at the output. Output offset currents vary with amount of loads. In case of the load conditions that is shown below (a resistor and a load of 5 Ω + 2 mH), the nominal offset is 0.5 A.



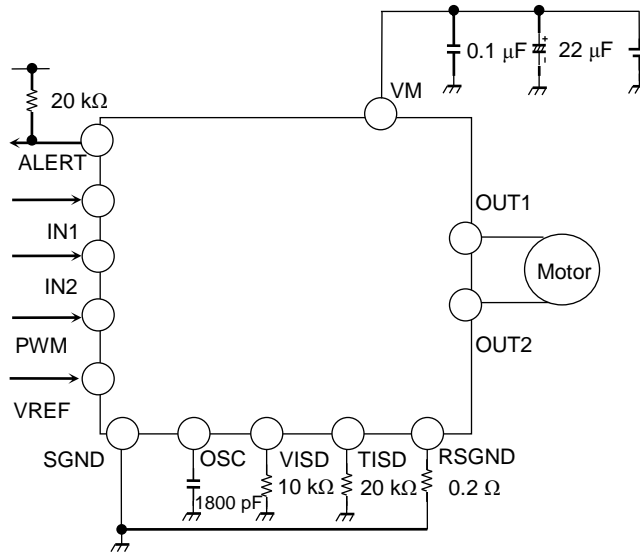
Reference waveforms when VREF = 0 V



Reference waveforms when VREF = 2.0 V

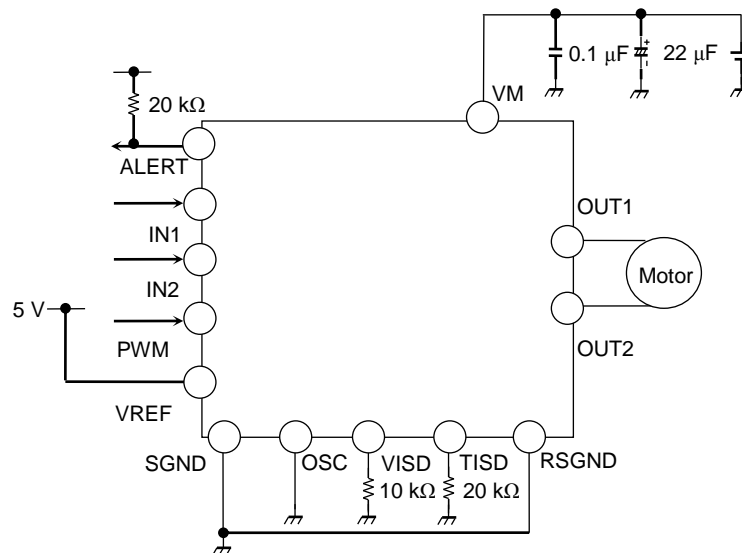
Conditions: VM = 24 V, OSC = 1800 pF, RSGND; 0.2 Ω , IN1 = H, IN2 = L,
 VREF = any value, Loads; 5 Ω + 2 mH

Application Circuit Example When Using Constant Current PWM Control



Application Circuit Example When NOT Using Constant Current PWM Control

The RSGND pin should be shorted to the SGND pin, and the VREF pin should be pulled High (5 V).



6. Power Dissipation

The power loss of the TB6569FG/FTG can be roughly estimated by the following equations:

(1) When PWM Duty = 100%

$$P = V_M \times I_{CC} + I_O^2 \times R_{ON} (U + L)$$

For example, when $V_M = 24 \text{ V}$ and the output current, $I_O = 0.5 \text{ A}$ (For I_{CC} and $R_{ON} (U + L)$, refer to the electrical characteristics on the data sheet.)

$$P (\text{typ.}) = 24 \text{ V} \times 2.5 \text{ mA (typ.)} + (0.5 \text{ A})^2 \times 0.55 \Omega (\text{typ.}) = 0.1975 \text{ W}$$

$$P (\text{max}) = 24 \text{ V} \times 8 \text{ mA (max)} + (0.5 \text{ A})^2 \times 0.9 \Omega (\text{max}) = 0.417 \text{ W}$$

(2) When using the PWM control

The power dissipation when using the PWM control can be roughly calculated as follows: (Switching loss occurring actually is not considered.)

$$P = \{V_M \times I_{CC} + I_O^2 \times R_{ON} (U + L)\} \times \text{duty of PWM}$$

7. Calculations of Heat Generation

Mutual relation of the ambient temperature, T_a , and the junction temperature, T_j , are roughly estimated by the following equation:

$$T_j = P \times R_{th(j-a)} + T_a$$

*: $R_{th(j-a)}$: Heat resistance between the junction and ambient temperatures

*: T_a : Ambient temperature (Stable ambient temperature avoiding the affect of any heat radiation)

For example, when $V_M = 24 \text{ V}$, output current (I_O) = 0.5 A, $T_a = 85^\circ\text{C}$, and $P (\text{max}) = 0.417 \text{ W}$, then

- (1) In mounting on the board of HSOP16-P-300-1.00 (refer to below figure)
When $R_{th(j-a)} = 89.3^\circ\text{C/W}$ and the board condition is 60mm × 30mm × 1.6mm,
 $T_j = 0.417 \text{ W} \times 89.3^\circ\text{C/W} + 85^\circ\text{C} = 122.2^\circ\text{C}$

Moreover, transient thermal resistance in 1s is as follows; $R_{th(j-a)} =$ approximately 16°C/W , then when $T_a = 85^\circ\text{C}$ and $P = 0.417 \text{ W}$,
 $T_j = 0.417 \text{ W} \times 16^\circ\text{C/W} + 85^\circ\text{C} = 91.7^\circ\text{C}$

- (2) In mounting on the board of P-VQFN32-0505-0.50-002 (refer to below figure)
When $R_{th(j-a)} = 35.2^\circ\text{C/W}$ and the board condition is based on JEDEC standard, 4-layer board of 76.2 mm × 114.3 mm × 1.6 mm,
 $T_j = 0.417 \text{ W} \times 35.2^\circ\text{C/W} + 85^\circ\text{C} = 99.7^\circ\text{C}$

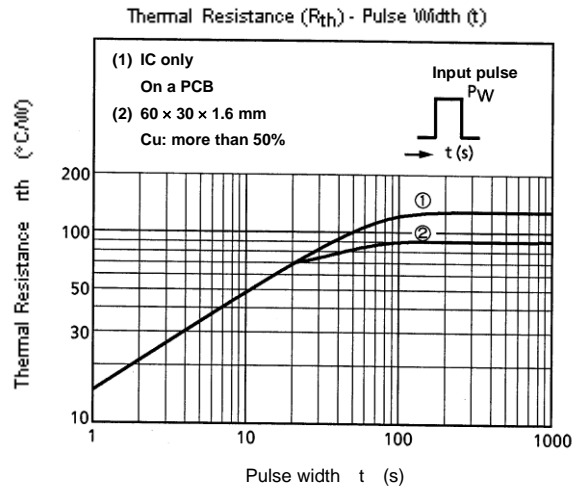
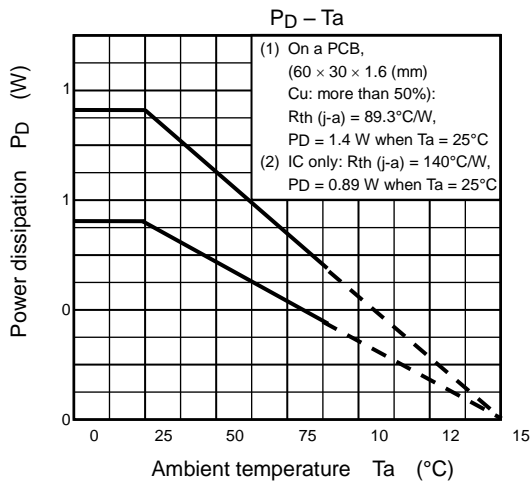
Moreover, transient thermal resistance in 1s is as follows; $R_{th(j-a)} =$ approximately 12°C/W , then when $T_a = 85^\circ\text{C}$ and $P = 0.417 \text{ W}$,
 $T_j = 0.417 \text{ W} \times 12^\circ\text{C/W} + 85^\circ\text{C} = 90^\circ\text{C}$

Pay attention that $R_{th(j-a)}$ depends on use conditions such as a board mounting method.

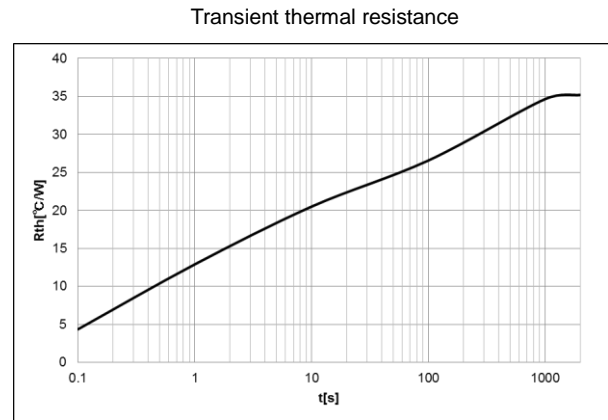
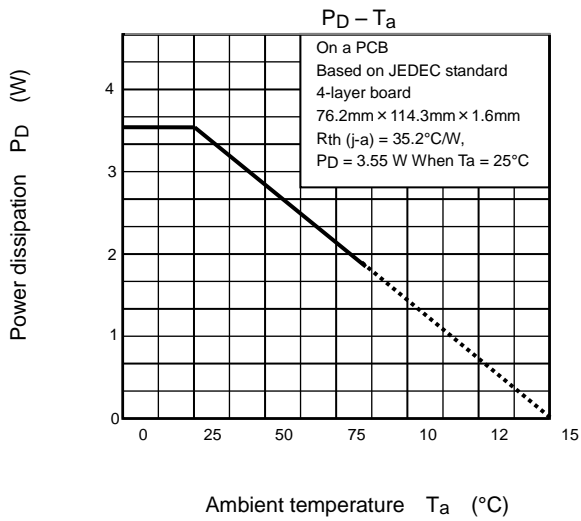
Higher the ambient temperature is, smaller the power dissipation becomes.

Note that the equations above are only the ways to find out rough estimation. A sufficient evaluation of the TB6569FG/FTG with the junction temperature less than 150°C is required for using the TB6569FG/FTG with a full safety margin.

HSOP16-P-300-1.00



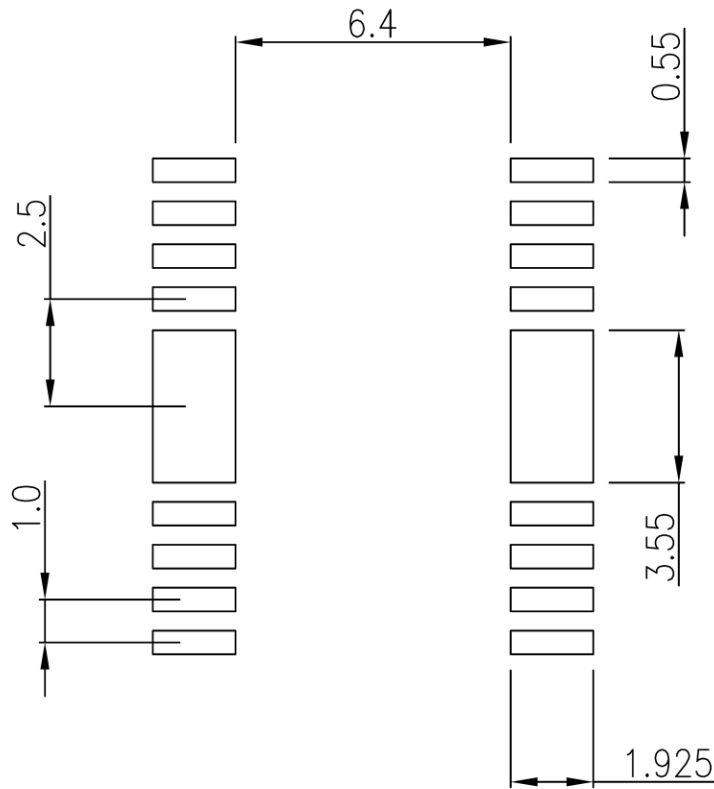
P-VQFN32-0505-0.50-002



8. Foot Pattern Example (for reference only)

HSOP16-P-300-1.00

Unit: mm



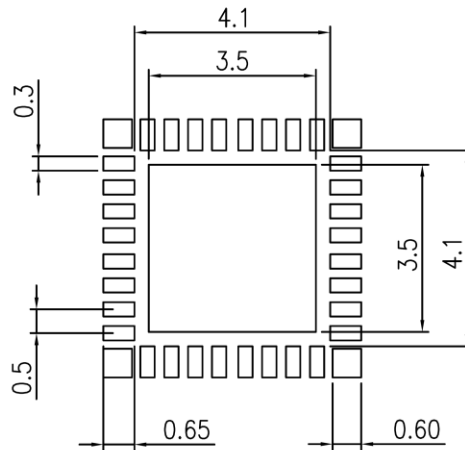
Notes

- All linear dimensions are given in millimeters unless otherwise specified.
- This drawing is based on JEITA ET-7501 Level3 and should be treated as a reference only.
TOSHIBA is not responsible for any incorrect or incomplete drawings and information.
- You are solely responsible for all aspects of your own land pattern, including but not limited to soldering processes.
- The drawing shown may not accurately represent the actual shape or dimensions.
- Before creating and producing designs and using, customers must also refer to and comply with the latest versions of all relevant TOSHIBA information and the instructions for the application that Product will be used with or for.

Note: Design the pattern in consideration of the heat design because the fin has the role of heat radiation.
(The fin should be insulated or connected to GND because it is connected to the back of the chip electrically.)

P-VQFN32-0505-0.50-002

Unit: mm



Notes

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- Before creating and producing designs and using, customers must also refer to and comply with the latest versions of all relevant TOSHIBA information and the instructions for the application that Product will be used with or for.

Note 1: Design the pattern in consideration of the heat design because the exposed metal portion of the back side has the role of heat radiation.
(The exposed metal portion of the back side should be insulated or connected to GND because it is connected to the back of the chip electrically.)

Note 2: Each OUT1, RSGND, OUT2, VM, and SGND pin has two pins. Please connect each same pin externally.

Notes on Contents

1. **Block Diagrams**
Some of the functional blocks, circuits, or constants in the block diagram may be omitted or simplified for explanatory purposes.
2. **Equivalent Circuits**
The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.
3. **Timing Charts**
Timing charts may be simplified for explanatory purposes.
4. **Application Circuits**
The application circuits shown in this document are provided for reference purposes only. Thorough evaluation is required, especially at the mass production design stage. Toshiba does not grant any license to any industrial property rights by providing these examples of application circuits.
5. **Test Circuits**
Components in the test circuits are used only to obtain and confirm the device characteristics. These components and circuits are not guaranteed to prevent malfunction or failure from occurring in the application equipment.

IC Usage Considerations**Notes on Handling of ICs**

- (1) The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings. Exceeding the rating(s) may cause the device breakdown, damage or deterioration, and may result injury by explosion or combustion.
- (2) Use an appropriate power supply fuse to ensure that a large current does not continuously flow in case of over current and/or IC failure. The IC will fully break down when used under conditions that exceed its absolute maximum ratings, when the wiring is routed improperly or when an abnormal pulse noise occurs from the wiring or load, causing a large current to continuously flow and the breakdown can lead smoke or ignition. To minimize the effects of the flow of a large current in case of breakdown, appropriate settings, such as fuse capacity, fusing time and insertion circuit location, are required.
- (3) If your design includes an inductive load such as a motor coil, incorporate a protection circuit into the design to prevent device malfunction or breakdown caused by the current resulting from the inrush current at power ON or the negative current resulting from the back electromotive force at power OFF. IC breakdown may cause injury, smoke or ignition. Use a stable power supply with ICs with built-in protection functions. If the power supply is unstable, the protection function may not operate, causing IC breakdown. IC breakdown may cause injury, smoke or ignition.
- (4) Do not insert devices in the wrong orientation or incorrectly. Make sure that the positive and negative terminals of power supplies are connected properly. Otherwise, the current or power consumption may exceed the absolute maximum rating, and exceeding the rating(s) may cause the device breakdown, damage or deterioration, and may result injury by explosion or combustion. In addition, do not use any device that is applied the current with inserting in the wrong orientation or incorrectly even just one time.

Points to Remember on Handling of ICs**(1) Over current Protection Circuit**

Over current protection circuits (referred to as current limiter circuits) do not necessarily protect ICs under all circumstances. If the Over current protection circuits operate against the over current, clear the over current status immediately.

Depending on the method of use and usage conditions, such as exceeding absolute maximum ratings can cause the over current protection circuit to not operate properly or IC breakdown before operation. In addition, depending on the method of use and usage conditions, if over current continues to flow for a long time after operation, the IC may generate heat resulting in breakdown.

(2) Thermal Shutdown Circuit

Thermal shutdown circuits do not necessarily protect ICs under all circumstances. If the thermal shutdown circuits operate against the over temperature, clear the heat generation status immediately.

Depending on the method of use and usage conditions, such as exceeding absolute maximum ratings can cause the thermal shutdown circuit to not operate properly or IC breakdown before operation.

(3) Heat Radiation Design

In using an IC with large current flow such as power amp, regulator or driver, please design the device so that heat is appropriately radiated, not to exceed the specified junction temperature (T_J) at any time and condition. These ICs generate heat even during normal use. An inadequate IC heat radiation design can lead to decrease in IC life, deterioration of IC characteristics or IC breakdown. In addition, please design the device taking into consideration the effect of IC heat radiation with peripheral components.

(4) Back-EMF

When a motor rotates in the reverse direction, stops or slows down abruptly, a current flow back to the motor's power supply due to the effect of back-EMF. If the current sink capability of the power supply is small, the device's motor power supply and output pins might be exposed to conditions beyond maximum ratings. To avoid this problem, take the effect of back-EMF into consideration in system design.

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