

TOSHIBA CMOS Linear Integrated Circuit Silicon Monolithic

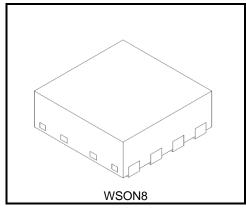
TCKE9 Series

23V, 4A eFuse IC with Adjustable Overcurrent Protection and High Speed Overvoltage clump.

The TCKE9 series is 23 V high input voltage Single Input-Single Output eFuse ICs. It can be used as a reusable fuse, and includes protection features like adjustable over current limit by an external resistor, short circuit protection, over voltage clamp, adjustable slew rete control by an external capacitance, adjustable under voltage protection by external resistor and thermal shutdown.

Switch ON resistance is only 34 m Ω (Typ.) and high output current is up to 4.0 A characteristics makes this series is suitable for protecting the power supply circuit of USB VBUS line, storage power supply circuit and battery charging application, etc.

This series is available in 0.5mm pitch small package WSON8 (2.0 mm (Typ.) x 2.0 mm (Typ.), t: 0.8 mm (MAX)). Thus this series is ideal for various applications such as portable electronics that require high-density soldering.



Weight: 7.99 mg (typ.)

Feature

- ◆ High input voltage: V_{IN} = 25.0 V (max)
- ullet Low ON resistance R_{ON}: 34 m Ω (Typ.)
- Adjustable overcurrent limit: 0.5 A to 4.0 A
- Fixed over voltage clamp

TCKE903: $V_{OVC} = 3.87 \text{ V (Typ.)}$

TCKE905: $V_{OVC} = 5.7 V (Typ.)$

TCKE912: $V_{OVC} = 13.7 V (Typ.)$

TCKE920: $V_{OVC} = 22.2 \text{ V (Typ.)}$

- Adjustable slew rate control by external capacitance for inrush current reduction
- Adjustable under voltage lockout by external resistor
- Thermal shutdown
- Small package: WSON8 (2.0 mm (Typ.) x 2.0 mm (Typ.), t: 0.8 mm (MAX)).
- IEC62368-1 (G9: Integrated circuit (IC) current limiters) Certified (Under planning)

Notice

This series is sensitive to electrostatic discharge.

Please ensure equipment and tools are adequately earthed when handling.

Start of commercial production 2024-03



1. Absolute Maximum Ratings (Note) (Ta = 25 °C)

Characteristics	Symbol		Unit							
Input voltage	VIN		V							
ILIM voltage	VILIM		V							
dV/dT voltage	V _d V/ _d T		V							
Control voltage	Ven/uvlo		V							
Output voltage	Vouт	-0	V							
Output voltage	Vout_pls	< 1µs Pulse	-1.2 (Min)	V						
FLAG voltage	VFLAG	-0.3 to 6.0		V						
FLAG sink current	IFLAG	0 to 10		mA						
Power dissipation	PD	2.0 (Note 1)		W						
Junction temperature	Tj	150		150		150		150		°C
Storage temperature	T _{stg}		-55 to 150	°C						

Note: Note: Using continuously under heavy loads (e.g. the application of high temperature/current/voltage and the significant change in temperature, etc.) may cause this product to decrease in the reliability significantly even if the operating conditions (i.e. operating temperature/current/voltage, etc.) are within the absolute maximum ratings. Please design the appropriate reliability upon reviewing the Toshiba Semiconductor Reliability Handbook ("Handling Precautions"/"Derating Concept and Methods") and individual reliability data (i.e. reliability test report and estimated failure rate, etc).

Note 1: Rating at mounting on a board: FR4 board. (76.2 mm x 114.3 mm x 1.6 mm, 4 layer)

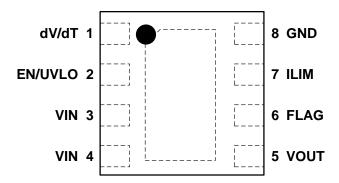
2. Operating Ranges

Characteristics	Symbol	Rar	Unit			
Input voltage	V _{IN}	2.7	V			
Output current	lout	DC 0 to 4.0		A		
ILIM External resistance	RILIM	487 to	Ω			
Control voltage	V _{EN/U} VLO	0 to	V			
FLAG voltage voltage	V _{FLAG}	0 1	0 to 6			
FLAG sink current	I _{FLAG}	0 to	mA			
Operation junction temperature	T _{j_opr}	-40 t	°C			
External capacitance	C _{dV/dT}	330	00 to	pF		



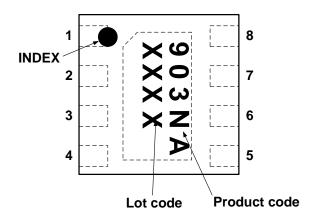
3. Pin Assignment (Top view)

WSON8



4. Top Marking (Top view)

Example: TCKE903NA

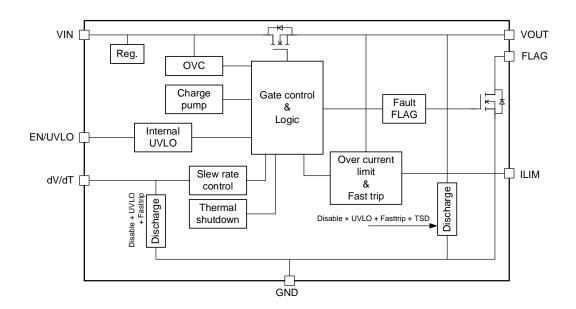


4.1. Product list

Part number	Over voltage Clamp	V _{EN/UVLO} function	Fault Response	Top Marking	Package
TCKE903NA	3.87 V (Typ.)	Active High	Auto-retry	903NA	
TCKE903NL	3.87 V (Typ.)	Active High	Latched	903NL	
TCKE905ANA	5.7 V (Typ.)	Active High	Auto-retry	905ANA	
TCKE905NL	5.7 V (Typ.)	Active High	Latched	905NL	WSON8
TCKE912NA	13.7 V (Typ.)	Active High	Auto-retry	912NA	WSONO
TCKE912NL	13.7 V (Typ.)	Active High	Latched	912NL	
TCKE920NA	22.2 V (Typ.)	Active High	Auto-retry	920NA	
TCKE920NL	22.2 V (Typ.)	Active High	Latched	920NL	



5. Block Diagram





6. PIN Description

PIN Name	Description
EN/UVLO	This pin has two functions. One function turns on the output voltage of the internal
	MOSFET as an enable signal. Another function can be used as a UVLO trip point with
	external resistors.
ILIM	Current limit set input. A resistor between ILIM pin and GND sets the current limit.
dV/dT	Rise time set input. A capacitor between dV/dT pin and GND sets the slew rate of VOUT
av/a1	when the device turns on.
FLAG	FRAG Open-Drain output pin.
FLAG	Fault event indicator which is pulled low when a fault is detected.
VIN	Supply Input. Input to the power switch and the supply voltage for the device.
GND	Ground.
VOUT	Output. Output of the power switch.

7. Operation Logic Table

	EN/UVLO = Low level	EN/UVLO = High level
Output	OFF	ON



8. Electrical Characteristics

8.1. DC Characteristics 1

(Unless otherwise specified, T_a = -40 to 125 °C, V_{IN} = 3.3 V (TCKE903), V_{IN} = 5 V (TCKE905), V_{IN} = 12 V (TCKE912), V_{IN} = 20 V (TCKE920), R_{ILIM} = 487 Ω)

Characteristics	Symbol Test Condition		T _a = 25 °C			T _a = -40 to 125 °C (Note2)		Unit	
				Min	Тур.	Max	Min	Max	
Basic operation									
VIN under voltage lockout	VIN_UVLOR1	Rising		_	2.51	_	2.35	2.68	V
(UVLO) threshold	V _{IN_UVLOF1}	Falling		_	2.41		2.25	2.53	V
EN/UVLO threshold voltage rising	VENUVR	-	_	_	1.20	_	1.13	1.27	V
EN/UVLO threshold voltage falling	VENUVF	_		_	1.10	_	1.03	1.17	V
EN/UVLO leak current	IENUVL	V _{IN} = 25 V, EN = 0 V		_	0	_	-0.1	0.1	μΑ
On registance	Pou	IOUT = 1.5 A, V _{IN} ≥ 4 V		_	34	37	_	52	mΩ
On resistance	Ron	I _{OUT} = 1.5 A, 2.7 V ≤ V _{IN} < 4 V		_	36.0	40.0	_	55.3	mΩ
		$V_{EN} = 3 \text{ V},$ $R_{ILIM} = 487 \ \Omega,$ $I_{OUT} = 0 \ A$	TCKE903	_	180	_	_	250	μΑ
			TCKE905	_	180	_	_	250	μΑ
Quiescent current (ON state)	IQ		TCKE912	_	185	_	_	250	μA
			TCKE920	_	190	_	_	250	μA
		V _{EN} = 0 V, V _{IN}	V _{EN} = 0 V, V _{IN} ≤ 5 V (Note 3)		0.07	_	_	_	μΑ
Quiescent current (OFF state)	I _{Q(OFF)}	V _{EN} = 0 V, 5 V	< V _{IN} ≤ 18 V	_	1.20	_	_	2.55	μΑ
Quescent current (OTT state)	<i>α(σ</i> , τ)	V _{EN} = 0 V, 18 V < V _{IN} ≤ 21 V TCKE920NA / NL only		_	1.20	_	_	2.55	μΑ
FLAG	•						•	•	
FLAG pin resistance	RFLAG	Low Level (No	ote 3)	_	12	_	_	_	Ω
FLAG Leak current	IFLAG	VFLAG = 6 V		_	_	_	_	0.1	μΑ

Note 2: This parameter is warranted by design.

Note 3: This parameter is reference only.



8.2. DC Characteristics 2

(Unless otherwise specified, T_a = -40 to 125 °C, V_{IN} = 3.3 V (TCKE903), V_{IN} = 5 V (TCKE905), V_{IN} = 12 V (TCKE912), V_{IN} = 20 V (TCKE920), R_{ILIM} = 487 Ω)

Characteristics	Symbol	Symbol Test Condition		Т	a = 25 °C		T _a = -40 to 125 °C (Note.2)		単位
				Min	Тур.	Max	Min	Max	
Over-voltage Protection									
		TCKE903		_	3.87	_	3.64	4.1	V
Over-voltage clamp (OVC)	Veve	TCKE905	Dour = 10 k0	_	5.7		5.4	6.0	V
ver-voltage Protection ver-voltage clamp (OVC) hreshold voltage utput voltage at Over-voltage amped VCL VC response time (Note 3) ver-current Protection vercurrent limit (Note 4) urrent monitor gain as easured on ILIM pin (ILIM/IouT) Note5) ast trip comparator level Note 3) IM pin short resistor detect hreshold (Note 3) hort circuit response time Note 3) urrent limit response time Note 3) urrent limit response time Note 3) hermal shutdown Protection (Note 3) hermal shutdown Threshold Talentage Total American Street Stre	Vovc	TCKE912	R _{OUT} = 10 kΩ	_	13.7	_	13.2	14.3	V
		TCKE920		_	22.2	_	21	23.3	V
		TCKE903		_	3.6	_	3.45	3.87	V
Output voltage at Over-voltage	M	TCKE905	10 1	_	5.55	_	5.2	5.8	V
clamped	VCLAMP	TCKE912	I _{OUT} = 10 mA	_	13.55	_	13	14.1.	V
		TCKE920		_	22.1	_	21.0	22.8	V
0)(0	_	IOUT = 4 A	•	_	6.0	_	_	_	μS
OVC response time (Note 3)	tovc	IOUT = 100 mA		_	9.0	_	_	_	μS
Over-current Protection		l				I	I.		
		R _{ILIM} = 487 Ω		_	4.09	_	3.40	4.68	
	I _{LIM} (IOUT_CL)	R _{ILIM} = 1780 Ω		_	1.11		0.95	1.36	A A
Overcurrent limit (Note 4)		R _{ILIM} = 4420 Ω		_	0.49	_	0.33	0.65	
		$R_{ILIM} = 0 \Omega \text{ (Note 3)}$		_	2	_	_	_	
		R _{ILIM} = OPEN (Note 3)		_	0	_	_	_	
		I _{OUT} = 4 A, V _{IN} ≥ 4 V		_	290	_	255	326	μΑ/Α
Current monitor gain as		I _{OUT} = 4 A, 2.7 V ≤ V _{IN} < 4 V		_	285	_	252	316	μΑ/Α
(Note5)	GIMON	I _{OUT} = 1 A, V _{IN} ≥ 4 V		_		_	247	325	μΑ/Α
,		I _{OUT} = 1 A, 2.7 V	≤ V _{IN} < 4 V	_	_	_	245	320	μΑ/Α
Fast trip comparator level (Note 3)	ISHORT	_	-	_	I _{LIM} × 1.6	_	_	_	Α
ILIM pin short resistor detect Threshold (Note 3)	RSHORTLIM	IOUT > ISHORT to	ILIM	_	100	_	_	_	Ω
Short circuit response time (Note 3)	tshort	I _{OUT} > I _{LIM} + 20%	5 to I _{OUT} ≤ I _{LIM}	_	2.0	_	_	_	μS
Current limit response time (Note 3)	tLIM	R _{ILIM} = 487 Ω	R _{ILIM} = 487 Ω		80	_	_	_	μS
Thermal shutdown Protection	(Note 3)	•			•	•		•	•
Thermal shutdown Threshold	T _{SD}	Tj		_	155	_	_	_	°C
Thermal shutdown Hysteresis	T _{SDH}	Tj		_	20	_	_		°C
Thermal shutdown Auto-retry Interval	ttsd,rst	_	-	_	55	_	_	_	ms

Note 2: This parameter is warranted by design.

Note 3: This parameter is reference only.

Note 4: Pulsed testing techniques used during this test maintain junction temperature approximately equal to ambient temperature.

Note 5: Without OVC operation.



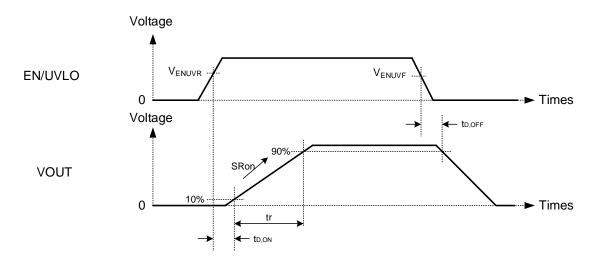
8.3. AC Characteristics

(Unless otherwise specified, $T_a = 25$ °C, $R_{ILIM} = 487 \Omega$, $R_{LOAD} = 100 \Omega$, $C_{IN} = C_{OUT} = 1 \mu F$)

Characteristics	Symbol	Test Condition	on	Min	Тур.	Max	Unit
			$V_{IN} = 3.3 \text{ V}$	_	181	_	μS
		$C_{dV/dT} = OPEN (Note 6)$	V _{IN} = 5 V	_	— 181 — μs — 197 — μs — 197 — μs — 240 — μs — 250 — μs — 310 — μs — 9.0 — μs — 8.9 — μs — 8.7 — μs — 9.2 — μs — 9.0 — μs — 9.2 — μs — 142 — μs — 180 — μs — 180 — μs — 353 — μs — 238 — μs — 366 — μs — 915 — μs — 18.6 — V/ms — 27.1 — V/ms — 11.0 — V/ms	μS	
Turn on delay	to 0		V _{IN} = 12 V	_	197	-	μS
(Note 3)	U,ON		$V_{IN} = 3.3 \text{ V}$	_	240		μS
CdV/dT C		$C_{dV/dT} = 3300 pF$	$V_{IN} = 5 V$	_	250		μS
		V _{IN} = 12 V	_	310		μS	
			$V_{IN} = 3.3 \text{ V}$	_	9.0		μS
		$C_{dV/dT} = OPEN (Note 6)$	$V_{IN} = 5 V$	_	8.9		μS
Turn off delay	to orr		V _{IN} = 12 V	_	8.7		μS
(Note 3)	ID,OFF	$C_{dV/dT} = 3300 \text{ pF}$ $V_{IN} = 3.3 \text{ V}$ $V_{IN} = 5 \text{ V}$ $V_{IN} = 12 \text{ V}$ $V_{IN} = 3.3 \text{ V}$	$V_{IN} = 3.3 \text{ V}$	_	9.2		μS
			$V_{IN} = 5 V$	_	9.0		μS
			8.8		μS		
	$C_{dV/dT} = 3300 \text{ pF}$ $C_{dV/dT} = OPEN \text{ (N)}$ $C_{dV/dT} = 3300 \text{ pF}$ $C_{dV/dT} = OPEN \text{ (N)}$ $C_{dV/dT} = 3300 \text{ pF}$ $C_{dV/dT} = 3300 \text{ pF}$ $C_{dV/dT} = OPEN \text{ (N)}$ $C_{dV/dT} = OPEN \text{ (N)}$ $C_{dV/dT} = OPEN \text{ (N)}$		$V_{IN} = 3.3 \text{ V}$	_	142		μS
		$C_{dV/dT} = OPEN (Note 6)$	$V_{IN} = 5 V$	_	180		μS
VOUT rise time		$t_{D,ON} = \frac{V_{IN} = 12 V}{V_{IN} = 3.3 V} - \frac{240}{-} - \frac{197}{-} - $	μS				
(Note 3)	$t_{D,OFF} = C_{dV/dT} = OPEN (Note 6) = V_{IN} = 3.3 \text{ V} \qquad$	238	_	μS			
		$C_{dV/dT} = 3300 pF$	V _{IN} = 5 V	_	366	_	μS
			V _{IN} = 12 V	_	915	_	μS
			V _{IN} = 3.3 V	_	18.6	_	V/ms
		$C_{dV/dT} = OPEN (Note 6)$	V _{IN} = 5 V	_	22.2	_	V/ms
	CDon		V _{IN} = 12 V	_	27.1	_	V/ms
(Note 3)	SKON		V _{IN} = 3.3 V	_	11.0	_	V/ms
Turn off delay (Note 3) VOUT rise time (Note 3) Output Rising Slew Rate		$C_{dV/dT} = 3300 pF$	V _{IN} = 5 V	_	11.0	_	V/ms
			V _{IN} = 12 V	_	11.0	_	V/ms

Note 3: This parameter is reference only.

Note 6: If $C_{dV/dT} = OPEN$ and there is a load at boot time, booting may not be completed.





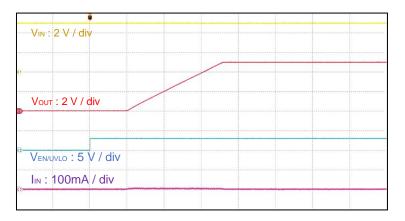
9. Function Explanation

9.1. Inrush current reduction

When the output is turned on, an inrush current flows to charge the capacitor connected to the load side. If this current is too large, the overcurrent protection circuit may malfunction, making it impossible to start up, or the output voltage may overshoot.

To prevent this, this function controls the slew rate when the output voltage rises by limiting the inrush current.

The following figure shows the rise of the output voltage (V_{OUT}) and the inrush current when the inrush current is limited by this function.

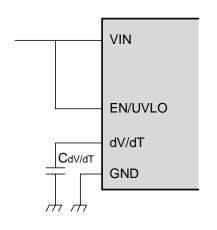


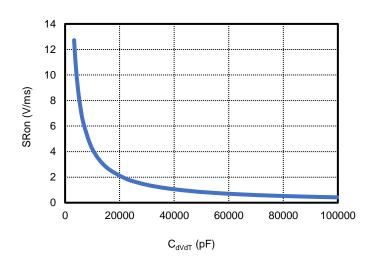
9.1.1. Setting of slew rate control for inrush current reduction

Toshiba eFuse IC has a variable inrush current function. The external capacitor at the dV/dT pin can be used to appropriately set the rise time Slew rate (SRon) of the output voltage. The formula for the rise time is as follows.

SRon
$$\left(\frac{V}{ms}\right) = \frac{I_{INRUSH}(mA)}{C_{OUT}(\mu F)}$$

$$C_{\text{dV/dT}} \text{ (pF)} = \frac{42000}{\text{SRon} \left(\frac{\text{V}}{\text{ms}}\right)}$$





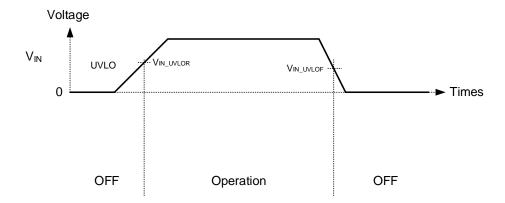
External Circuits around dV/dT pin



9.2. VIN Under voltage lockout function (VIN UVLO)

This function stops eFuse IC operation at low input voltage to prevent load malfunction.

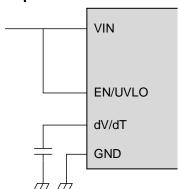
The TCKE9 series starts operation when the input voltage (V_{IN}) exceeds a certain voltage. This voltage has hysteresis at the rising and falling voltages, and operation stops when the voltage drops below a certain level.



9.3. EN/UVLO Operation

The TCKE9 series is equipped with EN/UVLO pin, and this pin can be used to control the operation of the whole eFuse IC. It is also possible to set the operating voltage of the UVLO function to the optimum value by externally attaching a resistor. The following are examples of uses for this pin.

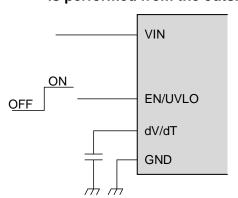
9.3.1. When the operation voltage of the UVLO function is not changed or the operation control is not performed.



Connect the EN/UVLO pin directly to the VIN pin. This eliminates the need for pull-up resistors in the TCKE9 series. The EN/UVLO pin is designed to be breakdown-voltage 25 V, and the VIN pin and the EN/UVLO pin can be directly connected. This helps reduce the number of parts.

Connections of EN/UVLO pin (Direct VIN connection)

9.3.2. When the operation voltage of the UVLO function is not changed and the operation control is performed from the outside.



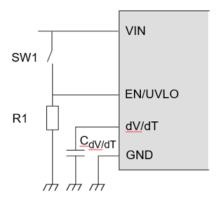
Connect external control signals directly into the EN/UVLO pin. Since the on/off threshold voltages of the EN/UVLO pin s are hysteretic, set the High level of the control signal to be 1.2 V (Typ.) or higher and the Low level of the control signal to be 1.1 V (Typ.) or lower.

If the EN/UVLO pin is open (indefinite), the eFuse IC operation may become abnormal. Be careful not to open this pin even when it is at the Low level.

Connections of EN/UVLO pin (External control signals)



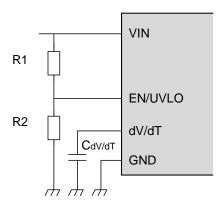
9.3.3. When the operation voltage of the UVLO function is not changed and the operation is controlled by the short-circuit switch with the VIN pin



Switches can be directly connected to the VIN pin so that the operation can be controlled. A pull-down resistor is required to prevent the EN/UVLO pin from being opened when the SW1 is opened. The value of the pull-down resistor may be any value that does not cause the EN/UVLO pin to become indefinite. However, when the SW1 is conducting, consider the current flowing through R1, and check the value with the actual device to determine the value of the pull-down resistor.

Connection examples of EN/UVLO pin (connected by VIN and switches)

9.3.4. To change the operating voltage of the UVLO function



a) When operation is not controlled

By adding an external resistor to the EN/UVLO pin, the operating voltage of the UVLO function can be changed to an optimum value. An example of the circuit is shown in the figure below.

a) When operation is not controlled by the EN/UVLO pin

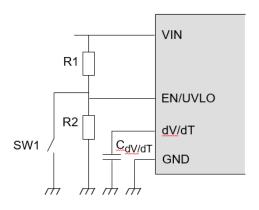
Operation is stopped when the input voltage drops by controlling the operation of the EN/UVLO pin with the voltage obtained by dividing the input voltage by an external resistor. The operating voltage of the UVLO function can be set to the optimum value by properly selecting the external resistance. However, the VIN voltage cannot be set to VIN UVLO threshold or less.

The equation for setting $V_{\text{IN/UVLO}(fall)}$ by controlling the external resistors R1 and R2 of the EN/UVLO pin is as follows.

$$V_{IN/UVLO(fall)}$$
 (V)= $\frac{R1+R2}{R2}$ × $V_{EN/UVLOF}$ (V)

V_{EN/UVF}: EN threshold (falling) 1.1 V (Typ.)

As described above, the control voltage of the EN/UVLO pin is hysteretic, and therefore the voltage to be activated at the time of rising changes. The start-up voltage $V_{IN/UVLO(rise)}$ is calculated by the following equation



b) When operation is controlled

$$V_{IN/UVLO(rise)}$$
 (V)= $\frac{R1+R2}{R2}$ × $V_{EN/UVLOR}$ (V)

V_{EN/UVR}: EN threshold (rising) 1.2 V (Typ.)

b) The circuit when the operation control is performed.

The switch can be connected in parallel with R2 to control the operation. contrary to the case of 9.3.3, the eFuse IC stops operating at the time of SW1 conduction. At this point, R1 is the current limiting resistor. Be careful when selecting the resistors for R1 and R2.



9.4. FLAG Output

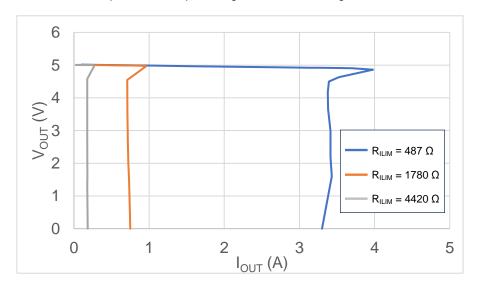
The TCKE9 series outputs L signal to FLAG pin in case of detecting abnormal conditions.

Failure mode	Symbol	IC Operation	FLAG output
Over temperature	TSD	Shutdown	Yes
V _{IN} Over voltage	OVC	V _{оит} voltage clamp	_
V _{IN} Under voltage	UVLO	Shutdown	_
Iou⊤ Over current	OCL	I _{OUT} current limiting	_
Vou⊤ Short circuit	OCL	I _{OUT} current limiting	_
ILM pin open	_	Shutdown	_
ILM pin short	_	Shutdown	Yes
(I _{ОUТ} > 3 A Тур.)			

If a large current continues in the OVC state or if the OCL state continues, the junction temperature of the IC will rise. If the junction temperature rises and TSD is detected, it shuts down and outputs FLAG.

9.5. Overcurrent limit function (OCL)

The overcurrent limit function prevents damage to the IC and load by suppressing power consumption in the event of an error. If the output current exceeds the limit current (I_{LIM}) due to a load error or short circuit, the output voltage and output current also decrease, thereby limiting the power consumed by the IC and the load. In addition to the short-circuit protection function, which will be described later, it is double-protected against overcurrent, which greatly contributes to the prevention of ignition and smoke. The figure below shows the relationship between output voltage and current during overcurrent limit function.

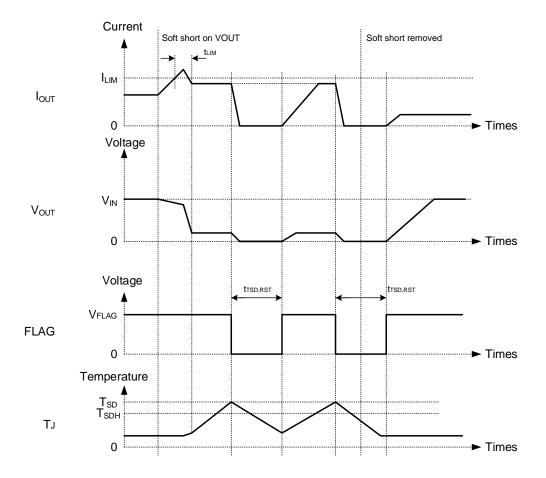


Output voltage – output current Characteristics during Overcurrent limit function



9.5.1. Auto-retry type overcurrent protection clamping operation

When the output current value reaches the I_{LIM} and an overcurrent is detected, the current is limited so that no more current flows than the I_{LIM} . Currently, the output voltage decreases according to the relationship between output voltage and current. If the overcurrent is not resolved at this stage, the IC temperature rises, and if the temperature reaches the temperature for overheat protection, the output stops and the IC goes into shutdown state. After a certain term, operation starts again, but if the overcurrent has not been resolved, the current is again limited. Therefore, the recovery attempt is repeated by the following cycle. Current limit \rightarrow temperature rise \rightarrow overheat protection \rightarrow shutdown \rightarrow temperature drop \rightarrow restart \rightarrow current limit



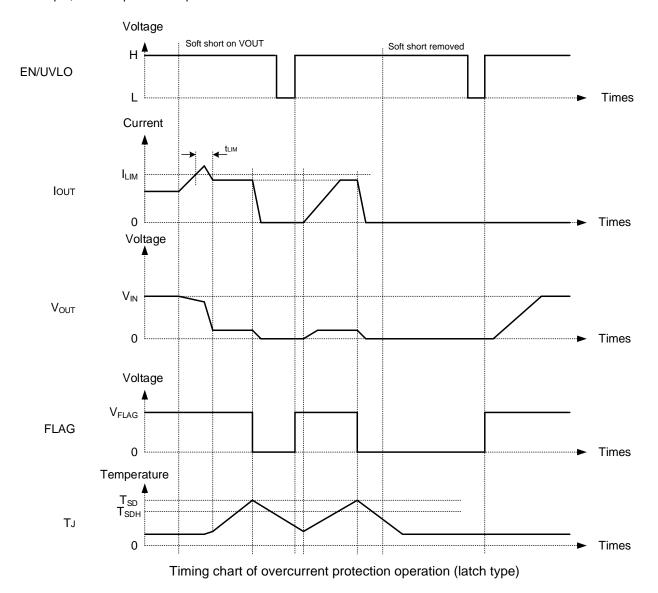
Timing chart of overcurrent protection operation (auto-retry type)

Rev.1.0



9.5.2. Latch type overcurrent protection clamping operation

In case of latch type, overheat protection operation is latched, To recover it, it must be restarted by a control signal on the EN/UVLO pin, etc. The protection operation continues until it is restarted.





9.5.3. Setting the overcurrent protection function

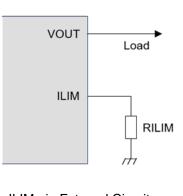
Toshiba eFuse IC has a variable current limit. By selecting the external resistor R_{ILIM} of the ILIM pin appropriately, the current limit can be set to the optimum value for each application. The I_{LIM} calculation method is common to the TCKE9 series and is shown in the figure below. However, the deviation between the theoretical value and the measured value is large when the current is 1 A or lower. Be sure to check the resistance value with the actual machine when selecting the resistance value.

Also, please make sure that R_{ILIM} does not fall below 487 Ω .

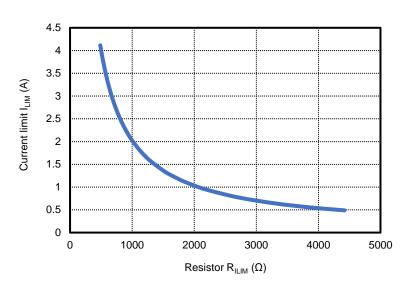
$$R_{ILIM}(\Omega) = \frac{1985}{I_{LIM}(A)-0.04}$$

 R_{ILIM} : ILIM pin external resistor (Ω)

The following is a diagram of the peripheral circuitry of the ILIM pin and the relation between RILIM and ILIM.







RILIM - ILIM Characteristics

9.5.4. R_{ILIM} pin open protection

When the ILIM pin is open, even if EN = High level is input, the device is VOUT keeps OFF.

9.5.5. R_{ILIM} Resistor short circuit protection

If the ILIM pin voltage does not rise for a certain term from EN = High level, the IC judges that the ILIM pin is shorted to GND and shifts to the internally set current limit operation state. The internal set current limit is 2A (Typ.).



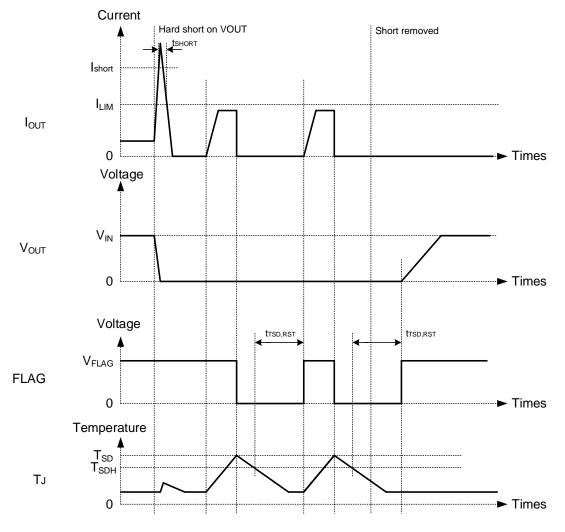
9.6. VOUT Short circuit protection (Fast trip)

The short-circuit protection function (fast trip) prevents excessive current from flowing by stopping operation when the power supply line or load is short-circuited due to some kind of abnormality. If the output current is 1.6 times the current limit (ILIM) for a very short term, the output is judged to be short-circuited and this function operates.

9.6.1. Auto-retry type VOUT short-circuit protection

This function judges that VOUT is short-circuited and stops the output when the output current reaches 1.6 times the overcurrent limit current (I_{ILIM}). The IC then starts operating with soft-start operation, but if the short-circuit continues, the IC temperature rises because the over-current limit current (OCL) operation continues. After that, the overheat protection function operates and the IC goes into shutdown state. After a certain term, the IC starts operating again, but if the VOUT short circuit has not been resolved, the current is limited again. Therefore, the recovery attempt is repeated by the following cycle.

Current limit → temperature rise → overheat protection → shutdown → temperature drop → restart → current limit

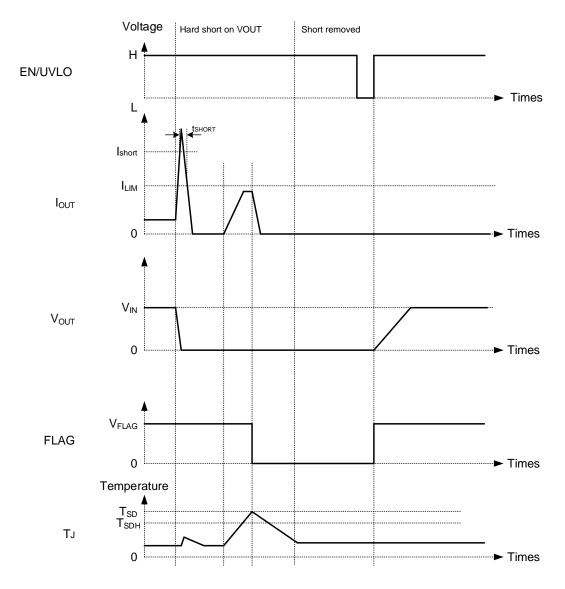


Timing chart of VOUT Short circuit protection (auto-retry type)



9.6.2. Latch type VOUT short-circuit protection

This function judges that VOUT pin is short-circuited and stops output when the output current reaches 1.6 times the overcurrent limit current (I_{LIM}). The IC then starts operating with soft-start operation, but if the short-circuit continues, the IC temperature rises as the operation continues at the overcurrent limit current (OCL). After that, the overheat protection function operates and the IC goes into shutdown state. To recover it, it must be restarted by a control signal on the EN/UVLO pin, etc. The protection operation continues until it is restarted.



Timing chart of VOUT Short circuit protection (latch type)



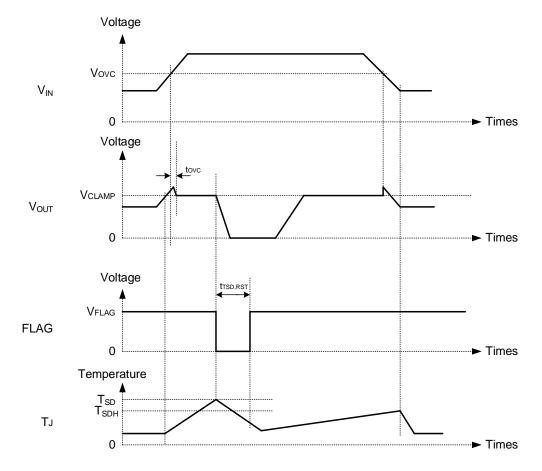
9.7. Overvoltage clamp protection function (OVC)

The overvoltage clamp function clamps the output voltage with a limited voltage and prevents overvoltage from being applied to the load without outputting any more voltage. The limit voltages are set as following table.

Characteristics	Symbol	Test Condition		Ta = 25°C			Ta = -40 to 125°C (Note2)		Unit
				Min	Тур.	Max	Min	Max	
Over-voltage Protection									
Over-voltage clamp (OVC)	Vovc	TCKE903		_	3.87	_	3.64	4.1	V
		TCKE905	R _{OUT} = 10 kΩ	_	5.7	_	5.4	6.0	V
Threshold voltage		TCKE912		_	13.7	_	13.2	14.3	V
- The second of		TCKE920		_	22.2	_	21	23.3	V
	.,	TCKE903		_	3.6	_	3.45	3.87	V
Over-voltage clamp (OVC)		TCKE905		_	5.55	_	5.2	5.8	V
Accuracy	VCLAMP	TCKE912	$I_{OUT} = 10 \text{ mA}$	_	13.55	_	13	14.1.	V
		TCKE920		_	22.1	_	21.0	22.8	V
OVC reasons time	tovo	I _{OUT} = 4 A	I _{OUT} = 4 A		6.0	_	_	_	μS
	tovc	I _{OUT} = 100 mA		_	9.0	_	_	_	μS

Note 2: This parameter is warranted by design.

Similar to the overcurrent protection and short-circuit protection, the auto-retry type will attempt to recover from the overvoltage clamp, but the latch type will retain this state until it is restarted.



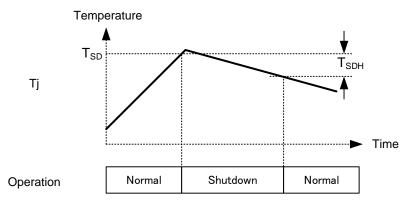
Timing chart of overvoltage clamp protection operation (auto-retry type)



9.8. Thermal shutdown protection

The overheat protection function (thermal shutdown) stops IC operation and transitions to the shutdown state when the junction temperature of the eFuse IC exceeds the set value due to continuous high current output.

The thermal shutdown protection has hysteresis between detect temperature and recover temperature.



9.8.1. Auto-retry type Thermal shutdown

After going into a shutdown state due to overheat protection, the unit restarts after a certain term.

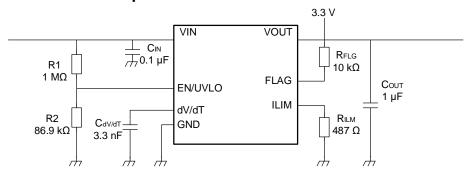
9.8.2. Latch type Thermal shutdown

After the overheat protection function is operated and the device goes into a shutdown state, it must be restarted by a control signal on the EN/UVLO pin, etc. to recover. The protection operation continues until restarted.



10. Application note

10.1. Application circuit example



Peripheral circuits Connect the power supply to the VIN pin. During normal operation, almost the same voltage as the VIN voltage is output from the VOUT pin through the internal MOSFET. If the current suddenly decreases, for example, when short-circuiting or overcurrent is protected, high-spike voltages may be generated due to back electromotive force of inductance components such as wirings connected to the input/output pins of the eFuse IC, causing damage to the eFuse IC and resulting destruction. In this case, a positive spike voltage is generated on the input side and a negative spike voltage is generated on the output side. When designing boards, design patterns so that the length of the wires on the input-side and output-side of the eFuse IC is as short as possible. Also, the GND wiring area should be as wide as possible to reduce the impedance. C_{IN} functions to suppress the peak value against the positive spike voltage generated by the inputs. The peak value V_{SPIKE} of the spike voltage and the capacitance value of the C_{IN} have the following relationship. It can be understood that the spike voltage can be reduced by increasing the C_{IN}.

$$V_{\text{SPIKE}}(V) = V_{\text{IN}} + I_{\text{OUT}} \times \sqrt{\frac{L_{\text{IN}}}{C_{\text{IN}}}}$$

 L_{IN} : effective inductance component of the input pin (H), I_{OUT} : output current (A)

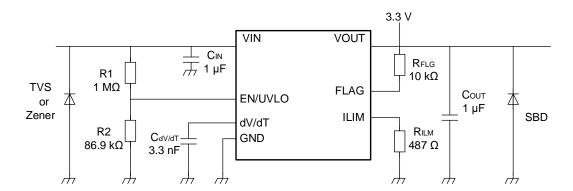
V_{SPIKE}: peak value of spiked voltage generated (V), V_{IN}: power supply voltage during normal operation (V)

If V_{IN} exceeds 10 V, C_{IN} is recommended to be 1 μ F or more. Be sure to measure it on the actual PCB board. If the V_{IN} voltage is high, the V_{OUT} voltage is also high, and the current change during short-circuit or overcurrent protection is large, causing unstable operation without stable V_{IN} and V_{OUT} and possibly leading to IC destruction. If transient voltages exceeding the absolute maximum ratings are applied to the VIN pin, connect a TVS diode (ESD protection diode) or a Zener diode between the input pins and GND.

For negative spike voltages generated on the output side, such as during short-circuit protection operation, connect an SBD (Schottky barrier diode) between the output pin and GND. This will prevent the output potential from dropping significantly below GND.

SBD is effective not only for protecting eFuse ICs, but also for protecting ICs and devices connected to the load side.

As noted above, TVS diode(Zener diode) and SBD are recommended for eFuse IC because they can provide more robust protective features. The diagram below shows the peripheral circuit diagram.





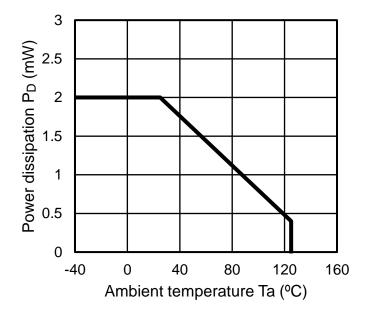
10.2. Power Dissipation

Both unit and board-mounted power dissipation ratings for TCKE9 series are available in the Absolute Maximum Ratings table.

Power dissipation is measured on the board shown below.

[The Board conditions]

FR4 board dimension: 76.2 mm x 114.3 mm x 1.6 mm, 4 layer





10.3. Attention in Use

Input/Output capacitors

Ceramic capacitors can be used with the IC, however some type capacitors may have very large temperature dependence. Please consider usage environment condition carefully to select the capacitors.

Mounting

The long distance between IC and output capacitor might affect phase assurance by impedance in wire and inductor. For stable power supply, output capacitor need to mount near IC as much as possible. Also VIN and GND pattern need to be large and make the wire impedance small as possible.

Permissible Loss

Please have enough design patterns for expected maximum permissible loss. And under consideration of surrounding temperature, input voltage, and output current etc., we recommend proper dissipation ratings for maximum permissible loss; in general maximum dissipation rating is 70 to 80 %.

Protection circuit

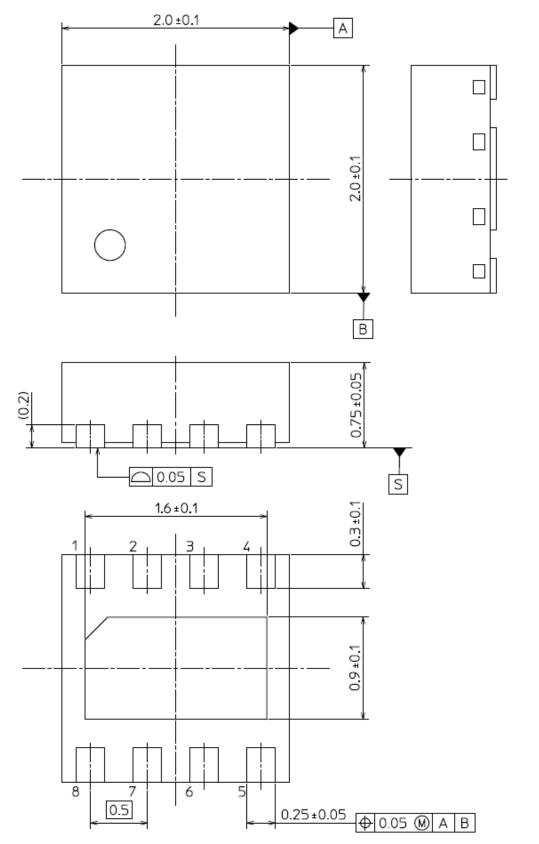
Over current protection, Over voltage protection and Thermal shut down function are designed in these products, but these are not designed to constantly ensure the suppression of the device within operation limits. Depending on the condition during actual usage, it could affect the electrical characteristic specification and reliability. Also note that if output pin pins and GND pin s are not completely shorted out, these products might be break down.

When using these products, please read through and understand the concept of dissipation for absolute maximum ratings from the above mention or our 'Semiconductor Reliability Handbook'. Then use these products under absolute maximum ratings in any condition. Furthermore, Toshiba recommend inserting failsafe system into the design.



11. Package Information

WSON8 Unit: mm



質量:7.99mg (Typ.)



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