

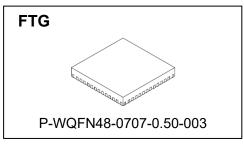
TOSHIBA BiCD Integrated Circuit Silicon Monolithic

TB67S261FTG

CLOCK-in controlled Bipolar Stepping Motor Driver

1. Description

The TB67S261 is a two-phase bipolar stepping motor driver using a PWM chopper. An interface is PHASE in control. Fabricated with the BiCD process, rating is 50 V/2.0 A.



Weight: 0.10 g (Typ.)

2. Features

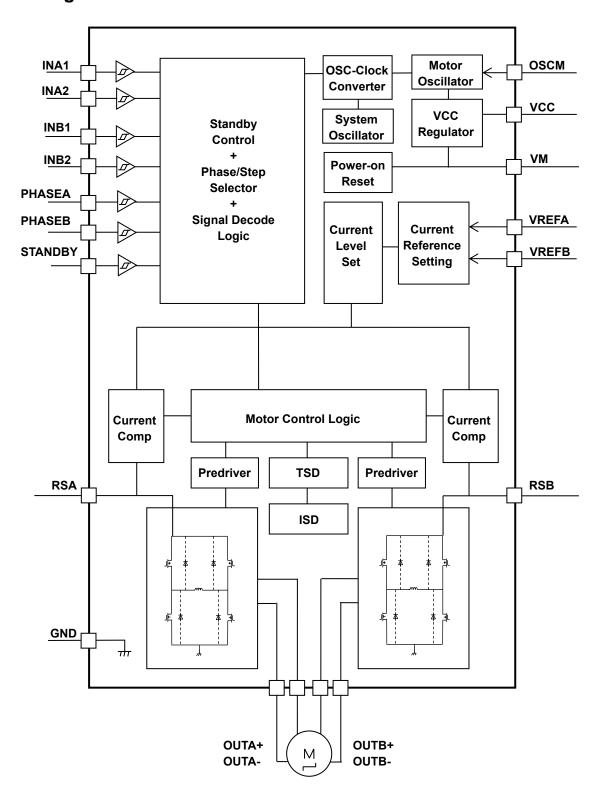
- BiCD process integrated monolithic IC.
- Capable of controlling 1 bipolar stepping motor.
- PWM controlled constant-current drive.
- Allows full, half, quarter step operation.
- Low on-resistance (High + Low side=0.8 Ω(Typ.)) MOSFET output stage.
- High efficiency motor current control mechanism (Advanced Dynamic Mixed Decay)
- High voltage and current (For specification, please refer to absolute maximum ratings and operation ranges)
- Error detection (TSD/ISD) signal output function
- Built-in error detection circuits (Thermal shutdown (TSD), over-current shutdown (ISD), and power-on reset (POR))
- Built-in VCC regulator for internal circuit use.
- Chopping frequency of a motor can be customized by external resistance and capacitor.
- package

TB67S261FTG: P-WQFN48-0707-0.50-003

Note: Please be careful about thermal conditions during use.



3. Block Diagram



Block Diagram Figure 3.1

Note: Functional blocks/circuits/constants in the block chart etc. may be omitted or simplified for explanatory purposes.



Note: All the grounding wires of the TB67S101A must run on the solder mask on the PCB and be externally terminated at only one point. Also, a grounding method should be considered for efficient heat

Careful attention should be paid to the layout of the output, VDD(VM) and GND traces, to avoid short circuits across output pins or to the power supply or ground. If such a short circuit occurs, the device may be permanently damaged.

Also, the utmost care should be taken for pattern designing and implementation of the device since it has power supply pins (VM, RS, OUT, GND) through which a particularly large current may run. If these pins are wired incorrectly, an operation error may occur or the device may be destroyed.

The logic input pins must also be wired correctly. Otherwise, the device may be damaged owing to a current running through the IC that is larger than the specified current.



4. Pin Assignments

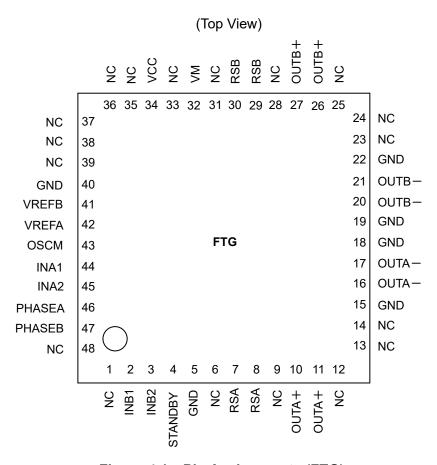


Figure 4.1 Pin Assignments (FTG)

Note: Please mount the four corner pins of the QFN package and the exposed pad to the GND area of the PCB.



5. Pin Description

5.1. TB67S261FTG (QFN48)

Table 5.1 Pin Description(1 to 28)

Pin No.	Pin Name	Function
1	NC	Non-connection pin
2	INB1	Motor Bch excitation control input 1
3	INB2	Motor Bch excitation control input 2
4	STANDBY	All-function-initializing and Low power dissipation mode
5	GND	Ground pin
6	NC	Non-connection pin
7	RSA(Note)	Motor Ach current sense pin
8	RSA(Note)	Motor Ach current sense pin
9	NC	Non-connection pin
10	OUTA+(Note)	Motor Ach (+) output pin
11	OUTA+(Note)	Motor Ach (+) output pin
12	NC	Non-connection pin
13	NC	Non-connection pin
14	NC	Non-connection pin
15	GND	Ground pin
16	OUTA-(Note)	Motor Ach (-) output pin
17	OUTA-(Note)	Motor Ach (-) output pin
18	GND	Ground pin
19	GND	Ground pin
20	OUTB-(Note)	Motor Bch (-) output pin
21	OUTB-(Note)	Motor Bch (-) output pin
22	GND	Ground pin
23	NC	Non-connection pin
24	NC	Non-connection pin
25	NC	Non-connection pin
26	OUTB+(Note)	Motor Bch (+) output pin
27	OUTB+(Note)	Motor Bch (+) output pin
28	NC	Non-connection pin



Table 5.2 Pin Description(29 to 48)

Pin No.	Pin Name	Function
29	RSB(Note)	Motor Bch current sense pin
30	RSB(Note)	Motor Bch current sense pin
31	NC	Non-connection pin
32	VM	Motor power supply pin
33	NC	Non-connection pin
34	VCC	Internal VCC regulator monitor pin
35	NC	Non-connection pin
36	NC	Non-connection pin
37	NC	Non-connection pin
38	NC	Non-connection pin
39	NC	Non-connection pin
40	GND	Ground pin
41	VREFB	Motor Bch output set pin
42	VREFA	Motor Ach output set pin
43	OSCM	Oscillating circuit frequency for chopping set pin
44	INA1	Motor Ach excitation control input 1
45	INA2	Motor Ach excitation control input 2
46	PHASEA	Current direction signal input for motor Ach
47	PHASEB	Current direction signal input for motor Bch
48	NC	Non-connection pin

Note: Please do not run patterns under NC pins.

Please connect the pins with the same pin name, while using the TB67S261.



5.2. Input/Output equivalent circuit

Table 5.6 Input/Output equivalent circuit

Pin name	IN/OUT signal	Equivalent circuit
INA1 INA2 PHASEA INB1 INB2 PHASEB STANDBY	Digital input (VIH/VIL) VIH: 2.0 V (Min) to 5.5 V (Max) VIL: 0 V (Min) to 0.8 V (Max)	Logic Input Pin GND GND
VCC VREFA VREFB	VCC voltage range 4.75 V (Min) to 5.0 V (Typ.) to 5.25 V (Max) VREF voltage range 0 V to 3.6 V	VCC Δ VREF Δ 1 KΩ VREF Δ 1 KΩ
OSCM	OSCM frequency setting range 0.64 MHz (Min) to 1.12 MHz (Typ.) to 2.4 MHz (Max)	OSCM T KΩ > COSCM
OUTA+ OUTA- OUTB+ OUTB- RSA RSB	VM power supply voltage range 10 V (Min) to 47 V (Max) OUT pin voltage 10 V (Min) to 47 V (Max)	RS DOUT-OUT-GND

Note: The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.



6. Functional Description (Stepping motor)

Motor output current (lout): The flow from OUT+ to OUT- is plus current. The flow from OUT- to OUT+ is minus current.

6.1. Step resolution select function

Table 6.1 Full step resolution

Ach				Bch			
Input			Output	Input			Output
PHASEA	INA1	INA2	lout(A)	PHASEB	INB1	INB2	lout(B)
Н	Н	Н	+100 %	Н	Н	Н	+100 %
L	Н	Н	-100 %	Н	Н	Н	+100 %
L	Н	Н	-100 %	L	Н	Н	-100 %
Н	Н	Н	+100 %	L	Н	Н	-100 %

Note: Please set INA1, INA2, INB1, and INB2 to Low until VM power supply reaches the proper operating range.

Table 6.2 Half step resolution

Ach				Bch			
	Input		Output		Input		Output
PHASEA	INA1	INA2	lout(A)	PHASEB	INB1	INB2	lout(B)
Н	Н	Н	+100 %	Н	Н	Н	+100 %
-	L	L	0 %	Н	Н	Н	+100 %
L	Н	Н	-100 %	Н	Н	Н	+100 %
L	Н	Н	-100 %	-	L	L	0 %
L	Н	Н	-100 %	L	Н	Н	-100 %
-	L	L	0 %	L	Н	Н	-100 %
Н	Н	Н	+100 %	L	Н	Н	-100 %
Н	Н	Н	+100 %	-	L	Ĺ	0 %

^{- :} Don't care



Table 6.3 Quarter step resolution

Ach				Bch				
Input		Output	Input			Output		
PHASEA	INA1	INA2	lout(A)	PHASEB	INB1	INB2	lout(B)	
Н	Н	L	+71 %	Н	Н	L	+71 %	
Н	L	Н	+38 %	Н	Н	Н	+100 %	
Х	L	L	0 %	Н	Н	Н	+100 %	
L	L	Н	-38 %	Н	Н	Н	+100 %	
L	Н	L	-71 %	Н	Н	L	+71 %	
L	Н	Н	-100 %	Н	L	Н	+38 %	
L	Н	Н	-100 %	Х	L	L	0 %	
L	Н	Н	-100 %	L	L	Н	-38 %	
L	Н	L	-71 %	L	Н	L	-71 %	
L	L	Н	-38 %	L	Н	Н	-100 %	
Х	L	L	0 %	L	Н	Н	-100 %	
Н	L	Н	+38 %	L	Н	Н	-100 %	
Н	Н	L	+71 %	L	Н	L	-71 %	
Н	Н	Н	+100 %	L	L	Н	-38 %	
Н	Н	Н	+100 %	Х	L	L	0 %	
Н	Н	Н	+100 %	Н	L	Н	+38 %	

X : Don't care

Table 6.4 Others

Pin Name	Н	L	Notes
INA1, INA2 INB1, INB2	The current value of e input 4	ach ch is set up with 2 value.	Please refer to the above-mentioned current value setting table.
PHASEA PHASEB	OUT+: H OUT-: L	OUT+: L OUT-: H	In PHASE=H, Charge current flows in the direction of OUT- from OUT+.
STANDBY	Standby release	Standby mode	In STANDBY= L, an internal oscillating circuit and a motor output part are stopped. (The drive of a motor cannot be performed.)



6.2. Current phasor

6.2.1. Full step resolution

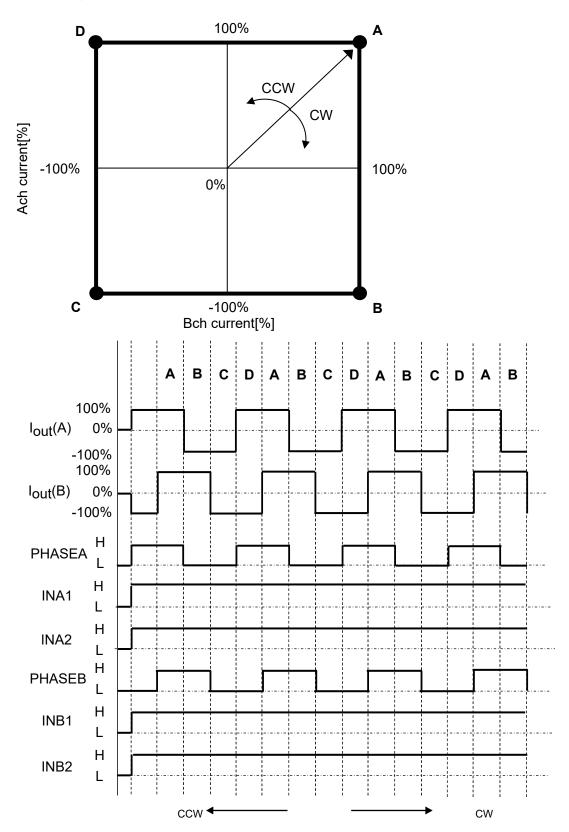
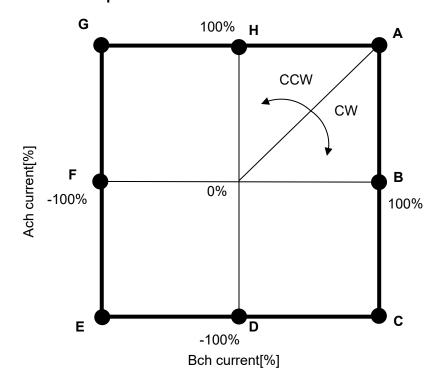


Figure 6.1 Full step resolution

Note: Timing charts may be simplified for explanatory purpose. Please set INA1, INA2, INB1, and INB2 to Low until VM power supply reaches the proper operating range.



6.2.2. Half step resolution



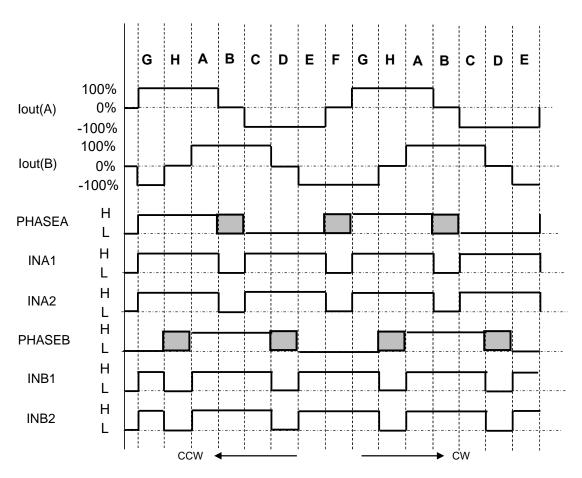
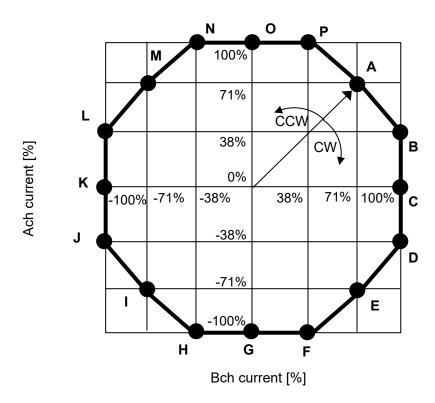


Figure 6.2 Half step resolution

Note: Timing charts may be simplified for explanatory purpose. Please set INA1, INA2, INB1, and INB2 to Low until VM power supply reaches the proper operating range.



6.2.3. Quarter step resolution



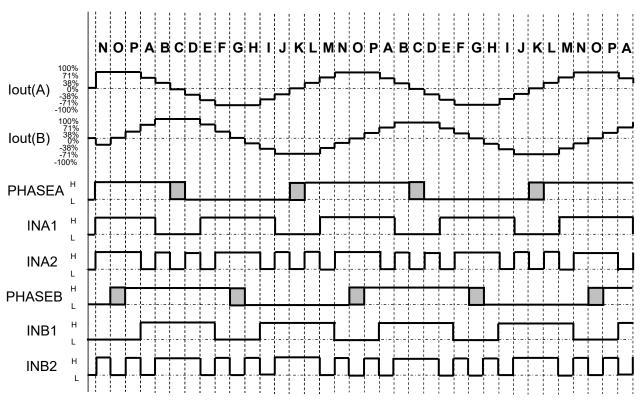


Figure 6.3 Quarter step resolution

Note: Timing charts may be simplified for explanatory purpose. Please set INA1, INA2, INB1, and INB2 to Low until VM power supply reaches the proper operating range.



6.3. Decay function

6.3.1. ADMD(Advanced Dynamic Mixed Decay) constant current control

The Advanced Dynamic Mixed Decay threshold, which determines the current ripple level during current feedback control, is a unique value.

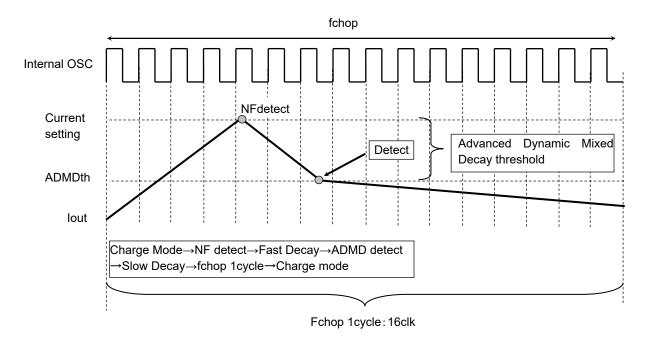


Figure 6.4 ADMD constant current control

Auto Decay Mode current waveform

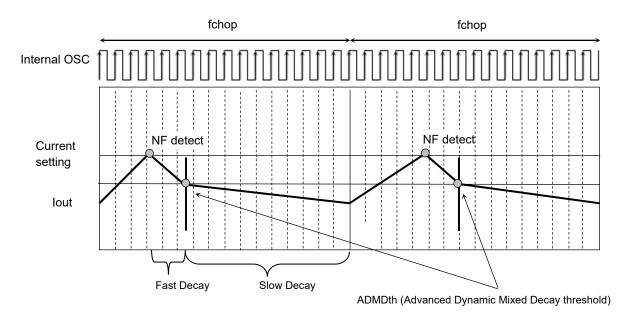


Figure 6.5 Auto Decay Mode current waveform



6.3.2. ADMD current waveform

6.3.2.1. When the next current step is higher:

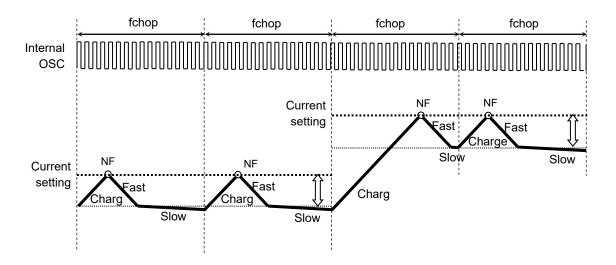
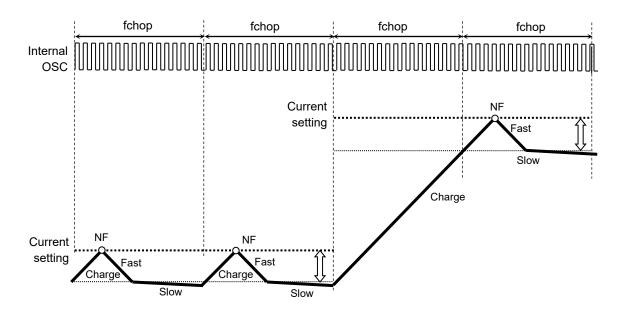


Figure 6.6 When the next current step is higher

6.3.2.2. When Charge period is more than 1 fchop cycle:



When Charge period is more than 1 fchop cycle Figure 6.7

When the Charge period is longer than fchop cycle, the Charge period will be extended until the motor current reaches the NF threshold. Once the current reaches the next current step, then the sequence will go on to decay mode.



6.3.2.3. When the next current step is lower:

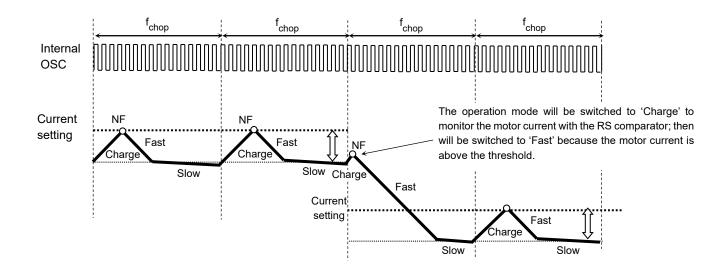


Figure 6.8 When the next current step is lower

6.3.2.4. When the Fast continues past 1 fchop cycle (the motor current not reaching the ADMD threshold during 1 fchop cycle)

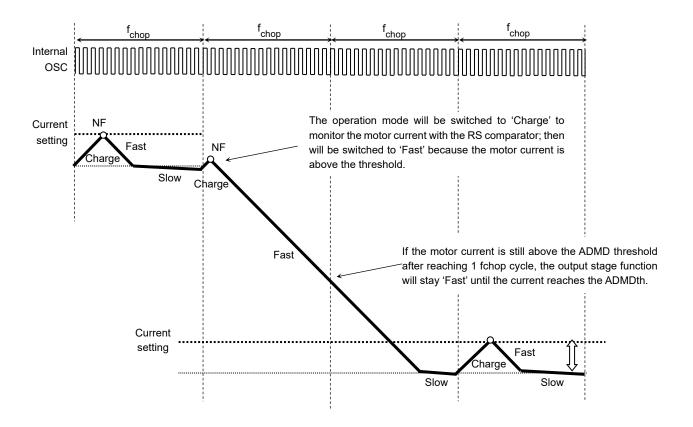


Figure 6.9 When the Fast continues past 1 fchop cycle



6.4. Output Transistor Operation Mode

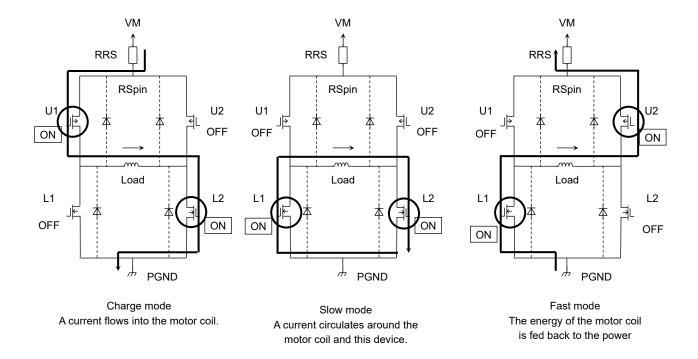


Figure 6.10 Output transistor function mode

Output transistor function

Table 6.5 Output MOSFET function

CLK	U1	U2	L1	L2
CHARGE	ON	OFF	OFF	ON
SLOW	OFF	OFF	ON	ON
FAST	OFF	ON	ON	OFF

Note: This table shows an example of when the current flows as indicated by the arrows in the figures shown above. If the current flows in the opposite direction, refer to the following table.

CLK	U1	U2	L1	L2
CHARGE	OFF	ON	ON	OFF
SLOW	OFF	OFF	ON	ON
FAST	ON	OFF	OFF	ON

This IC controls the motor current to be constant by 3 modes listed above.

Note: The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.



6.5. Calculation of the Predefined Output Current

For PWM constant-current control, this IC uses a clock generated by the OSCM oscillator. The peak output current (Setting current value) can be set via the current-sensing resistor (RS) and the reference voltage (Vref), as follows:

$$Iout(Max) = Vref(gain) \times \frac{Vref(V)}{RRS}$$

Vref(gain): the Vref decay rate is 1 / 5.0 (Typ.)

For example: In the case of a 100 % setup

when Vref = 3.0 V, Torque = 100 %, RRS = 0.51 Ω , the motor constant current (Setting current value) will be calculated as:

lout = $3.0 \text{ V} / 5.0 / 0.51 \Omega = 1.18 \text{ A}$

6.6. Calculation of the OSCM oscillation frequency (chopper reference frequency)

An approximation of the OSCM oscillation frequency (fOSCM) and chopper frequency (fchop) can be calculated by the following expressions.

$$\begin{aligned} &\text{fOSCM} = 1/[0.56 \text{ x } \{\text{Cx}(\text{R1} + 500)\}] \\ &\text{C, R1: External components for OSCM } (\text{C} = 270 \text{ pF} \text{ , R1} = 5.1 \text{ k}\Omega => \text{fOSCM} = \text{About 1.12 MHz } (\text{Typ.})) \end{aligned}$$

$$\begin{aligned} &\text{fchop} = \text{fOSCM} \text{ / 16} \\ &\text{fOSCM} = 1.12 \text{ MHz} => \text{fchop} = \text{Arround 70 kHz} \end{aligned}$$

If chopping frequency is raised. Ripple of current will become small and wave-like reproducibility will improve. However, the gate loss inside IC goes up and generation of heat becomes large.

By lowering chopping frequency, reduction in generation of heat is expectable. However, Ripple of current may become large. It is a standard about about 70 kHz. A setup in the range of 50 to 100 kHz is recommended.



7. Absolute maximum ratings

Table 7.1 Absolute Maximum Ratings ($T_a = 25$ °C)

	Symbol	Rating	Unit	Remarks	
	Motor power supply	VM	50	V	-
1	Motor output voltage	Vout	50	V	-
1	Motor output current	lout	2.0	Α	(Note 1)
Inte	rnal Logic power supply	VCC	6.0	V	When externally applied.
				V	-
	Logic input voltage	VIN(L)	-0.4	V	-
V	ref reference voltage	Vref	5.0	V	-
Power dissipation	Device alone	PD	1.3	W	(Note2)
Power dissipation	When mounted on a PCB	רם	4.1	W	(Note3)
0	perating temperature	TOPR	-20 to 85	°C	-
Ş	Storage temperature		-55 to 150	°C	-
J	unction temperature	Tj(Max)	150	°C	-

Note1: Usually, the maximum current value at the time should use 70 % or less of the absolute maximum ratings for a standard on thermal rating. The maximum output current may be further limited in view of thermal considerations, depending on ambient temperature and board conditions.

Note 2: Device alone. (Ta =25 °C)

If the ambient temperature is above 25 °C, the power dissipation must be de-rated by 10.4 mW/°C.

Note 3: 4 layer glass epoxy board (Board size: 100 mm × 110 mm × 1.6 mm, Ta =25 °C)

If the ambient temperature is above 25 °C, the power dissipation must be de-rated by 32.8 mW/°C.

Ta: Ambient temperature

Topr: Ambient temperature while the IC is active

Tj: Junction temperature while the IC is active. The maximum junction temperature is limited by the thermal shutdown (TSD) circuitry. It is advisable to keep the maximum current below a certain level so that the maximum junction temperature, Tj (MAX), will not exceed 120 °C.

Caution: Absolute maximum ratings

The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings.

Exceeding the rating (s) may cause device breakdown, damage or deterioration, and may result in injury by explosion or combustion.

The value of even one parameter of the absolute maximum ratings should not be exceeded under any circumstances. The TB67S101ANG does not have overvoltage detection circuit. Therefore, the device is damaged if a voltage exceeding its rated maximum is applied.

All voltage ratings, including supply voltages, must always be followed. The other notes and considerations described later should also be referred to.



(For reference) PD-Ta graph

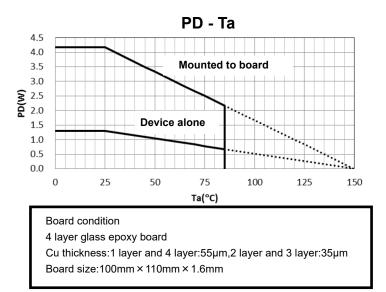


Figure7 (For reference) PD-Ta graph



8. Operation Ranges

Table 8.1 Operation Ranges (Ta = -20 to 85 °C)

Characteristics	Symbol	Min	Тур.	Max	Unit	Remarks
Motor power supply	VM	10	24	47	V	-
Motor output current	lout	-	1.4	2.0	Α	(Note 1)
Logio input voltago	VIN(H)	2.0	ı	5.5	>	Logic input High Level
Logic input voltage	VIN(L)	0	-	0.8	V	Logic input Low Level
Phase input frequency	fPHASE	-	-	400	kHz	-
Chopper frequency	fchop(range)	40	70	150	kHz	-
Vref input voltage	Vref	GND	2.0	3.6	V	-

Note1: Maximum current for actual usage may be limited by the operating circumstances such as operating conditions (exciting mode, operating time, and so on), ambient temperature, and heat conditions (board condition and so on).



9. Electrical Specifications

9.1. Electrical Characteristics 1 (Ta = 25 °C, VM = 24 V, unless otherwise specified)

Table 9.1 Electrical Specifications 1

Characteristics		Symbol	Test condition	Min	Тур.	Max	Unit
Logio input voltago	HIGH	VIN(H)	Logic input (Note)	2.0	-	5.5	V
Logic input voltage	LOW	VIN(L)	Logic input (Note)	0	-	0.8	V
Logic input hysteresis volta	ge	VIN(HYS)	Logic input (Note)	100	-	300	mV
	HIGH	IIN(H)	VIN(H)=3.3 V	-	33	-	μΑ
Logic input current	LOW	IIN(L)	VIN(L)=0 V	-	-	1	μΑ
		IM1	Output pins=open STANDBY=L	-	2	3.5	mA
Power consumption		IM2	Output pins=open STANDBY=H	-	3.5	5.5	mA
		IM3	Output pins=open Full step resolution	-	5.5	7	mA
Outrout leakers aument	High-side	IOH	VRS=VM=50 V,Vout=0 V	-	-	1	μA
Output leakage current	Low-side	IOL	VRS=VM=Vout=50 V	1	-	-	μA
Motor current channel diffe	rential	∆lout1	Current differential between Ch	-5	0	5	%
Motor current setting accur	асу	∆lout2	lout=1.5 A	-5	0	5	%
RS pin current	RS pin current		VRS=VM=24 V	0	-	10	μA
Motor output ON-resistance (High-side+Low-side)	Э	Ron(H+L)	Tj=25 °C, Forward direction (High-side+Low-side)	_	0.8	0.9	Ω

Note: VIN (H) is defined as the VIN voltage that causes the outputs (OUTA,OUTB) to change when a pin under test is gradually raised from 0 V. VIN (L) is defined as the VIN voltage that causes the outputs (OUTA, OUTB) to change when the pin is then gradually lowered from 5V. The difference between VIN (H) and VIN (L) is defined as the VIN (HYS).

Note: When the logic signal is applied to the device whilst the VM power supply is not asserted; the device is designed not to function, but for safe usage, please apply the logic signal after the VM power supply is asserted and the VM voltage reaches the proper operating range.



9.2. Electrical Specifications 2 (Ta =25 °C, VM = 24 V, unless otherwise specified)

Table 9.2 Electrical Specifications 2

Characteristics	Symbol	Test condition	Min	Тур.	Max	Unit
Vref input current	Iref	VREF=2.0 V	-	0	1	μΑ
VCC voltage	VCC	ICC=5.0 mA	4.75	5.0	5.25	V
VCC current	ICC	VCC=5.0 V	-	2.5	5	mA
Vref gain rate	Vref(gain)	Vref=2.0 V	1/5.2	1/5.0	1/4.8	_
Thermal shutdown(TSD) threshold (Note1)	TjTSD	_	145	160	175	°C
VM recovery voltage	VMR	_	7.0	8.0	9.0	V
Over-current detection (ISD) threshold (Note2)	ISD	_	2.1	3.0	4.0	Α

Note1: About TSD

When the junction temperature of the device reached the TSD threshold, the TSD circuit is triggered; the internal reset circuit then turns off the output transistors. Noise rejection blanking time is built-in to avoid misdetection. Once the TSD circuit is triggered, the device will be set to standby mode, and can be cleared by reasserting the VM power source, or setting the DMODE pins to standby mode. The TSD circuit is a backup function to detect a thermal error, therefore is not recommended to be used aggressively.

Note2: About ISD

When the output current reaches the threshold, the ISD circuit is triggered; the internal reset circuit then turns off the output transistors. Once the ISD circuit is triggered, the device keeps the output off until power-on reset (POR), is reasserted or the device is set to standby mode by DMODE pins. For fail-safe, please insert a fuse to avoid secondary trouble.

Back-EMF

While a motor is rotating, there is a timing at which power is fed back to the power supply. At that timing, the motor current recirculates back to the power supply due to the effect of the motor back-EMF.

If the power supply does not have enough sink capability, the power supply and output pins of the device might

rise above the rated voltages. The magnitude of the motor back-EMF varies with usage conditions and motor characteristics. It must be fully verified that there is no risk that the TB67S101A or other components will be damaged or fail due to the motor back-EMF.

Cautions on Overcurrent Shutdown (ISD) and Thermal Shutdown (TSD)

The ISD and TSD circuits are only intended to provide temporary protection against irregular conditions such as an output short-circuit; they do not necessarily guarantee the complete IC safety.

If the device is used beyond the specified operating ranges, these circuits may not operate properly: then the device may be damaged due to an output short-circuit.

The ISD circuit is only intended to provide a temporary protection against an output short-circuit. If such a condition persists for a long time, the device may be damaged due to overstress. Overcurrent conditions must be removed immediately by external hardware.

IC Mounting

Do not insert devices incorrectly or in the wrong orientation. Otherwise, it may cause breakdown, damage and/or deterioration of the device.



Characteristics	Symbol	Test condition	Min	Тур.	Max	Unit
Minimum PHASE pulse width	fPHASE(Min)	_	100	_	_	ns
	twp	_	50	-	-	
	twn	_	50	-	-	
Output transistor switching specific	tr	_	30	80	130	ns
	tf	_	40	90	140	
	tpLH(PHASE)	PHASE - Output	250	-	1200	
	tpHL(PHASE)	PHASE - Output	250	-	1200	
Analog noise blanking time	AtBLK	VM=24 V,lout=1.5 A Analog tblank	250	400	550	ns
Oscillator frequency accuracy	⊿fOSCM	COSC=270 pF, ROSC=5.1 kΩ	-15	-	+15	%
Oscillator reference frequency	fOSCM	COSC=270 pF, ROSC=5.1 kΩ	952	1120	1288	kHz
Chopping frequency	fchop	Output:Active(IOUT =1.5 A), fOSCM = 1120 kHz	-	70	-	kHz

AC Electrical Specification Timing chart

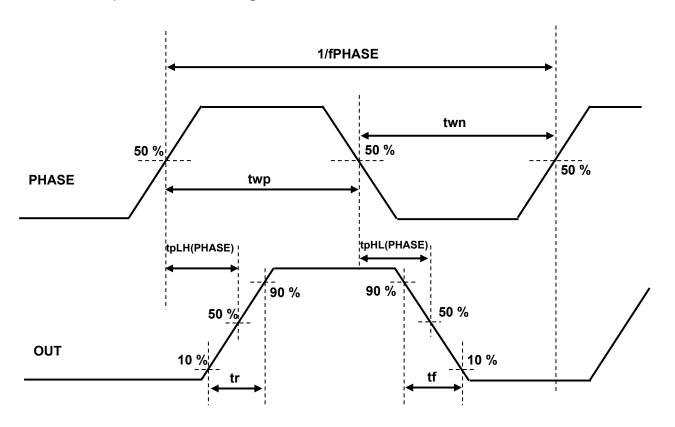
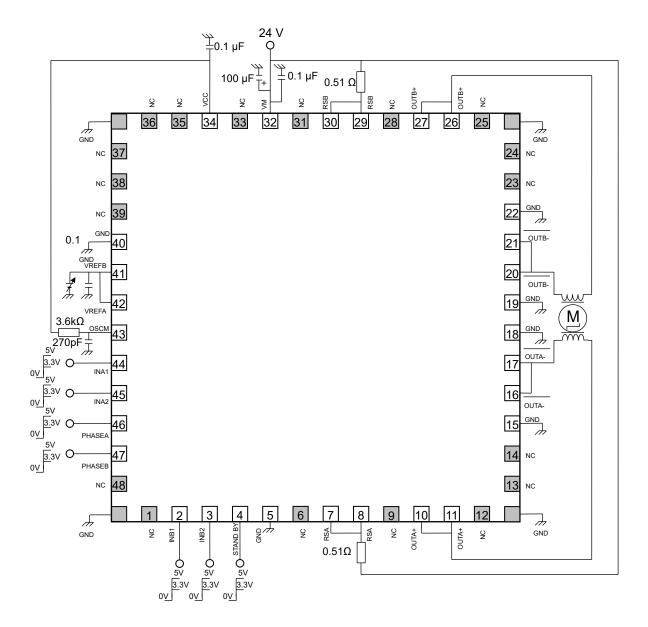


Figure 9.1 AC Electrical Specification Timing chart



Example Application Circuits

The values shown in the following figure are typical values. For input conditions, see the Operating Ranges.



Note: I will recommend the addition of a capacitor if necessary. The GND wiring must become one point as much as possible-earth.

The example of an applied circuit is for reference, and enough evaluation should be done before the massproduction design.

Moreover, it is not the one to permit the use of the industrial property.



Notes on Contents

1. Block Diagrams

Some of the functional blocks, circuits, or constants in the block diagram may be omitted or simplified for explanatory purposes.

2. Equivalent Circuits

The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.

3. Timing Charts

Timing charts may be simplified for explanatory purposes.

4. Application Circuits

The application circuits shown in this document are provided for reference purposes only. Thorough evaluation is required, especially at the mass-production design stage.

Toshiba does not grant any license to any industrial property rights by providing these examples of application circuits.

Test Circuits

Components in the test circuits are used only to obtain and confirm the device characteristics. These components and circuits are not guaranteed to prevent malfunction or failure from occurring in the application equipment.



10. IC Usage Considerations

10.1. Notes on handling of ICs

- (1) The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings. Exceeding the rating(s) may cause device breakdown, damage or deterioration, and may result in injury by explosion or combustion.
- (2) Use an appropriate power supply fuse to ensure that a large current does not continuously flow in the case of overcurrent and/or IC failure. The IC will fully break down when used under conditions that exceed its absolute maximum ratings, when the wiring is routed improperly or when an abnormal pulse noise occurs from the wiring or load, causing a large current to continuously flow and the breakdown can lead to smoke or ignition. To minimize the effects of the flow of a large current in the case of breakdown, appropriate settings, such as fuse capacity, fusing time and insertion circuit location, are required.
- (3) If your design includes an inductive load such as a motor coil, incorporate a protection circuit into the design to prevent device malfunction or breakdown caused by the current resulting from the inrush current at power ON or the negative current resulting from the back electromotive force at power OFF. IC breakdown may cause injury, smoke or ignition. Use a stable power supply with ICs with built-in protection functions. If the power supply is unstable, the protection function may not operate, causing IC breakdown. IC breakdown may cause injury, smoke or ignition.
- (4) Do not insert devices in the wrong orientation or incorrectly. Make sure that the positive and negative terminals of power supplies are connected properly. Otherwise, the current or power consumption may exceed the absolute maximum rating, and exceeding the rating(s) may cause device breakdown, damage or deterioration, and may result in injury by explosion or combustion.
 - In addition, do not use any device inserted in the wrong orientation or incorrectly to which current is applied even just once.
- (5) (Carefully select external components (such as inputs and negative feedback capacitors) and load components (such as speakers), for example, power amp and regulator.

 If there is a large amount of leakage current such as from input or negative feedback capacitor, the IC output DC voltage will increase. If this output voltage is connected to a speaker with low input withstand voltage, overcurrent or IC failure may cause smoke or ignition. (The overcurrent may cause smoke or ignition from the IC itself.) In particular, please pay attention when using a Bridge Tied Load (BTL) connection-type IC that inputs output DC voltage to a speaker directly.



10.2. Points to remember on handling of Ics

(1) Over-current Protection Circuit

Over-current protection circuits (referred to as current limiter circuits) do not necessarily protect ICs under all circumstances. If the Over-current protection circuits operate against the over-current, clear the over-current status immediately.

Depending on the method of use and usage conditions, such as exceeding absolute maximum ratings can cause the over-current protection circuit to not operate properly or IC breakdown before operation. In addition, depending on the method of use and usage conditions, if over-current continues to flow for a long time after operation, the IC may generate heat resulting in breakdown.

(2) Thermal Shutdown Circuit

Thermal shutdown circuits do not necessarily protect ICs under all circumstances. If the thermal shutdown circuits operate against the over temperature, clear the heat generation status immediately. Depending on the method of use and usage conditions, such as exceeding absolute maximum ratings can cause the thermal shutdown circuit to not operate properly or IC breakdown before operation.

(3) Heat Radiation Design

In using an IC with large current flow such as power amp, regulator or driver, please design the device so that heat is appropriately radiated, not to exceed the specified junction temperature (Tj) at any time and condition. These ICs generate heat even during normal use. An inadequate IC heat radiation design can lead to decrease in IC life, deterioration of IC characteristics or IC breakdown. In addition, please design the device taking into considerate the effect of IC heat radiation with peripheral components.

(4) Back-EMF

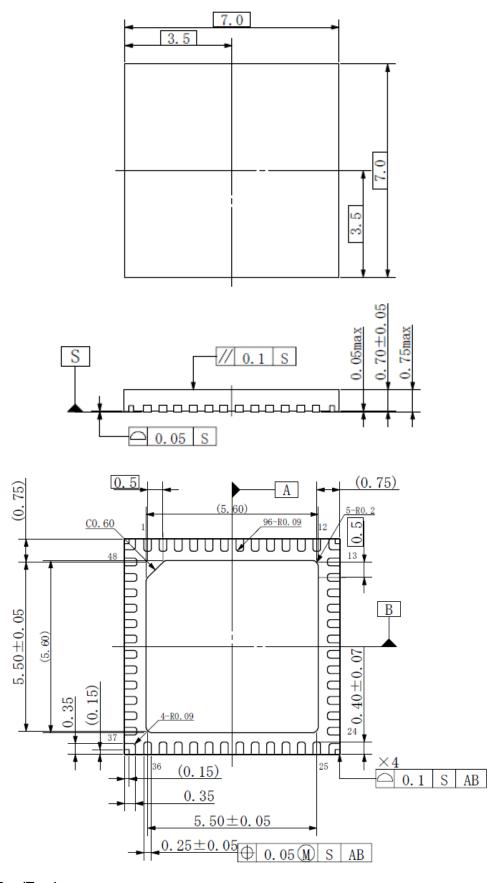
When a motor rotates in the reverse direction, stops or slows down abruptly, a current flow back to the motor's power supply due to the effect of back-EMF. If the current sink capability of the power supply is small, the device's motor power supply and output pins might be exposed to conditions beyond absolute maximum ratings. To avoid this problem, take the effect of back-EMF into consideration in system design.



11. Package Dimensions

P-WQFN48-0707-0.50-003

(unit:mm)



Weight: 0.10 g (Typ.)



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